TSG-RAN Meeting #9 Hawaii, U.S.A. , 20-22 September 2000

Title: Agreed CRs to TS 25.223

Source: TSG-RAN WG1

Agenda item: 5.1.3

No.	R1 T-doc	Spec	CR	Rev	Subject	Cat	Current	New
1	R1-000992	25.223	007	1	Gain Factors for TDD Mode	F	3.3.0	3.4.0
2	R1-000988	25.223	014	-	Synchronisation codes	F	3.3.0	3.4.0

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		25.223	CR	007r1		Current	Versio	n: <mark>3.3.0</mark>	
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Form: CR cover sheet, version 2 for 3GPP and SMG The latest version of this form is available from: ftp://ftp.3gpp.org/Information/CR-Form-v2.doc Proposed change affects: (U)SIM ME X UTRAN / Radio X Core Network (at least one should be marked with an X) (U)SIM ME X UTRAN / Radio X Core Network									
Source:	TSG RAN W	/G1				<u>C</u>	Date:	27/06/2000	
Subject:	Gain Factors	s for TDD Mode							
Work item:									
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6.5.1 Combination of physical channels in uplink

Figure 4 illustrates the principle of combination how the maximum of two different physical uplink channels are combined within one timeslot. The DPCHs to be combined belong to same CCTrCH, did undergo spreading as described in sections before and are thus represented by complex-valued sequences. First, the amplitude of all DPCHs is adjusted according to UL open loop power control as described in [10]. Each DPCH complex valued spread channel is then separately weighted by a weight factor $\gamma_{\underline{L}}$ and combined using complex addition. After combination of Physical Channels the gain factor β_{i} is applied, depending on the actual TFC as described in [10].

In case of different CCTrCH, principle shown in Figure 4 applies to each CCTrCH separately.

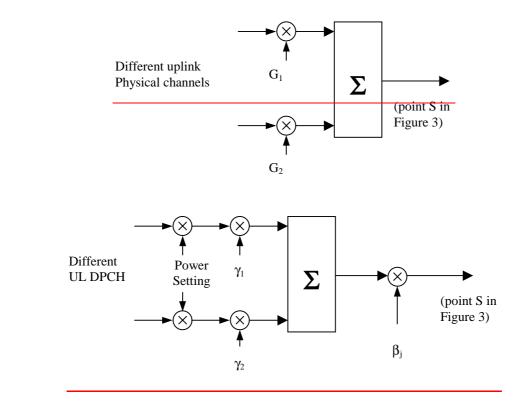


Figure 4: Combination of different physical channels in uplink

The values of weight factors γ_i are depending on the spreading factor SF of the corresponding DPCH:

<u>SF of DPCH_i</u>	$\gamma_{\underline{i}}$
<u>16</u>	1
<u>8</u>	$\sqrt{2}$
<u>4</u>	2
<u>2</u>	$2\sqrt{2}$
<u>1</u>	4

The possible values for gain factors β_j (corresponding to *j*-th TFC) are listed in table below:

Signalling value for β_j	<u>Quantized value β_j</u>
<u>15</u>	<u>16/8</u>
<u>14</u>	<u>15/8</u>
<u>13</u>	<u>14/8</u>
<u>12</u>	<u>13/8</u>
<u>11</u>	<u>12/8</u>

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<u>10</u>	<u>11/8</u>
<u>9</u>	<u>10/8</u>
<u>8</u>	<u>9/8</u>
<u>7</u>	<u>8/8</u>
<u>6</u>	<u>7/8</u>
<u>5</u>	<u>6/8</u>
<u>4</u>	<u>5/8</u>
<u>3</u>	<u>4/8</u>
2	<u>3/8</u>
<u>1</u>	<u>2/8</u>
<u>0</u>	<u>1/8</u>

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Source:	TSG RAN	WG1				Date:		
Subject:	Synchronis	ation codes						
Work item:	TS 25.223							
Category: F (only one category E shall be marked C with an X) E Reason for C change: C	Correspon Addition of Functional Editorial m	ds to a correction f feature modification of fea	ature		se	<u>Release:</u> cribing th	Phase 2 Release 96 Release 97 Release 98 Release 99 Release 00	x n of
Clauses affecte	<u>d:</u> 7.1, 7.	.2						
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7 Synchronisation codes

7.1 Code Generation

The primary synchronisation code (PSC), C_p , is constructed as a so-called generalised hierarchical Golay sequence. The PSC is furthermore chosen to have good aperiodic auto correlation properties.

Define $a = \langle x_1, x_2, x_3, ..., x_{16} \rangle = \langle 1, 1, 1, 1, 1, 1, -1, -1, 1, -1, 1, -1, 1, -1, 1 \rangle$ The PSC is generated by repeating the sequence 'a' modulated by a Golay complementary sequence and creating a complex-valued sequence with identical real and imaginary components.

and the left most index corresponds to the chip transmitted first in time.

The 12 secondary synchronization codes, { C_0 , C_1 , C_3 , C_4 , C_5 , C_6 , C_8 , C_{10} , C_{12} , C_{13} , C_{14} , C_{15} } are complex valued with identical real and imaginary components, and are constructed from the position wise multiplication of a Hadamard sequence and a sequence z, defined as

and $x_1, x_2, x_3, ..., x_{16}$ are the same as in the definition of the sequence 'a' above. The Hadamard sequences are obtained as the rows in a matrix H_8 constructed recursively by:

$$\begin{aligned} H_0 &= (1) \\ H_k &= \begin{pmatrix} H_{k-1} & H_{k-1} \\ H_{k-1} & -H_{k-1} \end{pmatrix}, \quad k \geq 1 \end{aligned}$$

The rows are numbered from the top starting with row 0 (the all ones sequence). Denote the *n*:th Hadamard sequence h_n as a row of H_8 numbered from the top, n = 0, 1, 2, ..., 255, in the sequel.

Furthermore, let $h_m(\underline{il})$ and $z(\underline{il})$ denote the $\underline{i:l}$ th symbol of the sequence h_m and z, respectively where $\underline{i-l} = 0, 1, 2, ..., 255$ and $\underline{i-l} = 0$ corresponds to the leftmost symbol. The i:th secondary SCH code word, C_i , i = 0, 1, 3, 4, 5, 6, 8, 10, 12, 13, 14, 15 is then defined as

$$C_i = (1 + j) \times \langle h_m(0) \times z(0), h_m(1) \times z(1), h_m(2) \times z(2), \dots, h_m(255) \times z(255) \rangle,$$

where $m = (16 \times i)$ and the leftmost chip in the sequence corresponds to the chip transmitted first in time.

7.2 Code Allocation

Three <u>secondary</u> SCH codes are QPSK modulated and transmitted in parallel with the primary synchronization code. The QPSK modulation carries the following information:

- the code group that the base station belongs to (<u>32 code groups:</u>5 bits; Cases 1, 2);
- the position of the frame within an interleaving period of 20 msec (<u>2 frames:</u>1 bit, Cases 1, 2);
- the position of the <u>SCH</u> slot(<u>s)</u> within the frame (<u>2 SCH slots:</u>1 bit, Case 2).

The modulated <u>secondary SCH</u> codes are also constructed such that their cyclic-shifts are unique, i.e. a non-zero cyclic shift less than 2 (Case 1) and 4 (Case 2) of any of the sequences is not equivalent to some cyclic shift of any other of the sequences. Also, a non-zero cyclic shift less than 2 (Case 1) and 4 (Case 2) of any of the sequences is not equivalent to itself with any other cyclic shift less than 8. The secondary synchronization codes are partitioned into two code sets for Case 1 and four code sets for Case 2. The set is used to provide the following information: