TSG-RAN Meeting #9 Hawaii, U.S.A. , 20-22 September 2000

Title: Agreed CRs to TS 25.212

Source: TSG-RAN WG1

Agenda item: 5.1.3

No.	R1 T-doc	Spec	CR	Rev	Subject	Cat	Current	New
1	R1-000698	25.212	079	-	Clarification of compressed mode terminology	F	3.3.0	3.4.0
2	R1-001134	25.212	085	1	Editorial corrections in Turbo code internal interleaver section	F	3.3.0	3.4.0
3	R1-000918	25.212	086	1	Clarification on DL slot format for compressed mode by SF/2	F	3.3.0	3.4.0
4	R1-001042	25.212	087	-	Corrections	F	3.3.0	3.4.0
5	R1-001101	25.212	088	1	Clarifications to TS 25.212	F	3.3.0	3.4.0
6	R1-001058	25.212	089	-	Correction regarding DSCH	F	3.3.0	3.4.0
7	R1-001059	25.212	090	-	Correction regarding CPCH	F	3.3.0	3.4.0
8	R1-001143	25.212	092	1	Bit separation and collection for rate matching	F	3.3.0	3.4.0
9	R1-001104	25.212	093	-	Puncturing Limit definition in WG1 specification	F	3.3.0	3.4.0

3GPP TSG RAN Meeting #9

Document	<i>R1-00-0698</i>
	or 3GPP use the format TP-99xxx for SMG, use the format P-99-xxx

Hawaii, USA, 20-22 September 2000

	CHANGE REQUEST Please see embedded help file at the bottom of this page for instructions on how to fill in this form correctly.					
	25.212 CR 079 Current Version: 3.3.0					
GSM (AA.BB) or 3G ((AA.BBB) specification number ↑					
For submission t list expected approval r	neeting # here for information non-strategic use only)					
For	m: CR cover sheet, version 2 for 3GPP and SMG The latest version of this form is available from: ftp://ftp.3gpp.org/Information/CR-Form-v2.doc					
Proposed change affects: (at least one should be marked with an X) (U)SIM ME X UTRAN / Radio X Core Network						
Source:	TSG RAN WG1 Date: 2000-07-03					
Subject:	Clarification of compressed mode terminology					
Work item:						
Category:FA(only one categoryBShall be markedCWith an X)D	CorrectionXRelease:Phase 2Corresponds to a correction in an earlier releaseRelease 96Release 96Addition of featureRelease 97Release 97Functional modification of featureRelease 98Release 98Editorial modificationRelease 00X					
<u>Reason for</u> change:	Clarification of "compressed mode" with respect to TTI or frame. Clarification of frame structure A or B. Editorial correction in references section.					
Clauses affected	<u>2, 3.2, 4.2.5.1, 4.2.7.1.2, 4.2.7.2, 4.2.9, 4.4</u>					
Affected:	Other 3G core specifications \rightarrow List of CRs:Other GSM core specifications \rightarrow List of CRs:MS test specifications \rightarrow List of CRs:BSS test specifications \rightarrow List of CRs:O&M specifications \rightarrow List of CRs:					
<u>Other</u> comments:						

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<----- double-click here for help and instructions on how to create a CR.

2 References

The following documents contain provisions which, through reference in this text, constitute provisions of the present document.

- References are either specific (identified by date of publication, edition number, version number, etc.) or non-specific.
- For a specific reference, subsequent revisions do not apply.
- For a non-specific reference, the latest version applies.
- [1] 3G TS 25.201: "Physical layer General Description".
- [2] 3G TS 25.211: "Physical channels and mapping of transport channels onto physical channels (FDD)".
- [3] 3G TS 25.213: "Spreading and modulation (FDD)".
- [4] 3G TS 25.214: "Physical layer procedures (FDD)".
- [5] 3G TS 25.215: "<u>Physical layer –</u> Measurements (FDD)".
- [6] 3G TS 25.221: "Physical channels and mapping of transport channels onto physical channels (TDD)".
- [7] 3G TS 25.222: "Multiplexing and channel coding (TDD)".
- [8] 3G TS 25.223: "Spreading and modulation (TDD)".
- [9] 3G TS 25.224: "Physical layer procedures (TDD)".
- [10] 3G TS 25.225: "<u>Physical layer Measurements (TDD)</u>".
- [11] 3G TS 25.302: "Services Provided by the Physical Layer".
- [12] 3G TS 25.402: "Synchronisation in UTRAN, Stage 2".

6

Symbols 3.2

For the purposes of the present document, the following symbols apply:

$\begin{bmatrix} x \\ Lx \end{bmatrix}$	round towards ∞ , i.e. integer such that $x \le \sqrt{x} < x+1$ round towards $-\infty$, i.e. integer such that $x-1 < \lfloor x \rfloor \le x$
x	absolute value of x
$\operatorname{sgn}(x)$	signum function, i.e. $sgn(x) = \begin{cases} 1; & x \ge 0\\ -1; & x < 0 \end{cases}$
N _{first} N _{last}	The first slot in the TG, located in the first compressed radio frame if the TG spans two frames.
N_{last}	The last slot in the TG, located in the second compressed radio frame if the TG spans two frames.
N _{tr}	Number of transmitted slots in a radio frame.

Unless otherwise is explicitly stated when the symbol is used, the meaning of the following symbols is:

i	TrCH number
j	TFC number
k	Bit number
l	TF number
т	Transport block number
n _i	Radio frame number of TrCH <i>i</i> .
р	PhCH number
r	Code block number
Ι	Number of TrCHs in a CCTrCH.
C_i	Number of code blocks in one TTI of TrCH <i>i</i> .
F_i	Number of radio frames in one TTI of TrCH <i>i</i> .
M_i	Number of transport blocks in one TTI of TrCH <i>i</i> .
$N_{data,j}$	Number of data bits that are available for the CCTrCH in a radio frame with TFC <i>j</i> .
$N^{\scriptscriptstyle cm}_{\scriptscriptstyle data,j}$	Number of data bits that are available for the CCTrCH in a compressed radio frame with TFC <i>j</i> .
Р	Number of PhCHs used for one CCTrCH.
PL	Puncturing Limit for the uplink. Signalled from higher layers
RM_i	Rate Matching attribute for TrCH <i>i</i> . Signalled from higher layers.

Temporary variables, i.e. variables used in several (sub)clauses with different meaning.

x, X

y, Y z, Z

-- initialisation of counter of

4.2.5 1st interleaving

In Compressed Mode by puncturing, bits marked with a fourth value on top of $\{0, 1, \delta\}$ and noted p, are introduced in the radio frames to be compressed, in positions corresponding to the first bits of the radio frames. They will be removed in a later stage of the multiplexing chain to create the actual gap. Additional puncturing has been performed in the rate matching step, over the TTI containing the compressed radio frame, to create room for these p-bits. The following subclause describes this feature.

4.2.5.1 Insertion of marked bits in the sequence to be input in first interleaver

In normal mode, compressed mode by higher layer scheduling, and compressed mode by spreading factor reduction:

 $x_{i,k} = z_{i,k}$ and $X_i = Z_i$

In case of the TTI contains a radio frame that is compressed mode by puncturing and fixed positions are used, sequence $x_{i,k}$ which will be input to first interleaver for TrCH *i* and TTI *m* within largest TTI, is built from bits $z_{i,k}$, $k=1, ..., Z_i$, plus $Np_{i,\max}^{TTI,m}$ bits marked p and $X_i = Z_i + Np_{i,\max}^{TTI,m}$, as is described thereafter.

 $Np_{i,\text{max}}^{TTI,m}$ is defined in the Rate Matching subclause 4.2.7.

 $P1_{Fi}(x)$ defines the inter column permutation function for a TTI of length $F_i \times 10$ ms, as defined in Table 3 in section 4.2.5.2. $P1_{Fi}(x)$ is the Bit Reversal function of x on $log_2(F_i)$ bits.

- NOTE 1: C[x], x= 0 to F_i -1, the number of bits p which have to be inserted in each of the F_i segments of the TTI, where x is the column number before permutation, i.e. in each column of the first interleaver. C[P1_{*Fi*}(x)] is equal to $Np_{i,\max}^{m \times F_i + x}$ for x equal 0 to F_i -1 for fixed positions. It is noted $Np_i^{m \times F_i + x}$ in the following initialisation step.
- NOTE 2: cbi[x], x=0 to F_i 1, the counter of the number of bits p inserted in each of the F_i segments of the TTI, i.e. in each column of the first interleaver x is the column number before permutation.

col = 0

while $col < F_i$ do -- here col is the column number after column permutation

 $C[P1_{Fi}(col)] = Np_i^{m \times F_i + col}$ -- initialisation of number of bits p to be inserted in each of the F_i segments of the TTI number m

 $\operatorname{cbi}[\operatorname{P1}_{Fi}(\operatorname{col})] = 0$

number of bits p inserted in each of the F_i segments of the TTI

col = col + 1

end do

n = 0, m = 0

while $n < X_i$ do -- from here col is the column number before column permutation

 $col = n \mod F_i$

if cbi[col] < C[col] **do**

 $x_{i,n} = p$

cbi[col] = cbi[col]+1 -- update counter of number of bits p inserted

-- insert one p bit

else

 $x_{i,n} = z_{i,m}$

m = m+1

-- no more p bit to insert in this segment

n = n + 1

end do

21

4.2.7.1.2 Determination of parameters needed for calculating the rate matching pattern

The number of bits to be repeated or punctured, $\Delta N_{i,j}$, within one radio frame for each TrCH *i* is calculated with equation 1 for all possible transport format combinations *j* and selected every radio frame. $N_{data,j}$ is given from subclause 4.2.7.1.1.

In <u>a</u> compressed <u>mode-radio frame</u>, $N_{data,j}$ is replaced by $N_{data,j}^{cm}$ in Equation 1. $N_{data,j}^{cm}$ is given as follows:

In <u>a radio frame</u> compressed mode by higher layer scheduling, $N_{data,i}^{cm}$ is obtained by executing the algorithm in

subclause 4.2.7.1.1 but with the number of bits in one radio frame of one PhCH reduced to $\frac{N_{tr}}{15}$ of the value in normal mode.

 N_{tr} is the number of transmitted slots in a compressed radio frame and is defined by the following relation:

$$N_{tr} = \begin{cases} 15 - TGL, \text{ if } N_{first} + TGL \le 15\\ N_{first}, \text{ in first frame if } N_{first} + TGL > 15\\ 30 - TGL - N_{first}, \text{ in second frame if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

In <u>a radio frame</u> compressed mode by spreading factor reduction, $N_{data,j}^{cm} = 2 \times (N_{data,j} - N_{TGL})$, where

$$N_{TGL} = \frac{15 - N_{tr}}{15} \times N_{data,j}$$

If $\Delta N_{i,j} = 0$ then the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed.

If $\Delta N_{i,j} \neq 0$ the parameters listed in subclauses 4.2.7.1.2.1 and 4.2.7.1.2.2 shall be used for determining e_{ini} , e_{plus} , and e_{minus} (regardless if the radio frame is compressed or not).

4.2.7.2 Determination of rate matching parameters in downlink

For downlink $N_{data,j}$ does not depend on the transport format combination *j*. $N_{data,*}$ is given by the channelization code(s) assigned by higher layers. Denote the number of physical channels used for the CCTrCH by *P*. $N_{data,*}$ is the number of bits available to the CCTrCH in one radio frame and defined as $N_{data,*}=P\times15\times(N_{data1}+N_{data2})$, where N_{data1} and N_{data2} are defined in [2]. Note that contrary to the uplink, the same rate matching patterns are used in normal TTIs containing no comressed radio frames and in TTIs containing radio frames compressed mode by spreading factor reduction or higher layer scheduling.

In the following, the total amount of puncturing or repetition for the TTI is calculated.

Additional calculations for <u>TTIs containing radio frames</u> compressed <u>mode</u> by puncturing in case of fixed positions <u>are</u> <u>used</u>, are performed to determine this total amount of rate matching needed.

For compressed mode by puncturing, in TTIs where some compressed radio frames occur, the puncturing is increased or the repetition is decreased compared to what is calculated according to the rate matching parameters provided by higher layers. This allows to create room for later insertion of marked bits, noted p-bits, which will identify the positions of the gaps in the compressed radio frames.

The amount of additional puncturing corresponds to the number of bits to create the gap in the TTI for TrCH *i*. In case of fixed positions, it is calculated in addition to the amount of rate matching indicated by higher layers. It is noted $Np_{i,\max}^{TTI,m}$.

In fixed positions case, to obtain the total rate matching $\Delta N_{i,\max}^{TTI,cm,m}$ to be performed on the TTI *m*, $Np_{i,\max}^{TTI,m}$ is subtracted from $\Delta N_{i,\max}^{TTI,m}$ (calculated based on higher layers RM parameters as for normal rate matching). This allows to create room for the $Np_{i,\max}^{TTI,m}$ bits p to be inserted later. If the result is null, i.e. the amount of repetition matches exactly the amount of additional puncturing needed, then no rate matching is necessary.

In case of compressed mode by puncturing and fixed positions, for some calculations, $N'_{data,*}$ is used for radio frames with gap instead of $N_{data,*}$, where $N'_{data,*} = P \times 15 \times (N'_{data1} + N'_{data2})$. N'_{data1} and N'_{data2} are the number of bits in the data fields of the slot format used for the current compressed mode, i.e. slot format A or B as defined in [2] corresponding to the Spreading Factor and the number of transmitted slots in use.

The number of bits corresponding to the gap for TrCH *i*, in each radio frame of its TTI is calculated using the number of bits to remove on all Physical Channels $N_{TGL}[k]$, where k is the radio frame number in the largest TTI.

For each radio frame k of the largest TTI that is overlapping with a transmission gap, N_{TGL}[k] is given by the relation:

$$N_{TGL} = \begin{cases} \frac{TGL}{15} \times N'_{data,*}, \text{ if } N_{first} + TGL \leq 15\\ \frac{15 - N_{first}}{15} \times N'_{data,*}, \text{ in first radio frame of the gap if } N_{first} + TGL > 15\\ \frac{TGL - (15 - N_{first})}{15} \times N'_{data,*}, \text{ in second radio frame of the gap if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

Note that N $_{TGL}[k] = 0$ if radio frame k is not overlapping with a transmission gap.

4.2.9 Insertion of discontinuous transmission (DTX) indication bits

In the downlink, DTX is used to fill up the radio frame with bits. The insertion point of DTX indication bits depends on whether fixed or flexible positions of the TrCHs in the radio frame are used. It is up to the UTRAN to decide for each CCTrCH whether fixed or flexible positions are used during the connection. DTX indication bits only indicate when the transmission should be turned off, they are not transmitted.

4.2.9.1 1st insertion of DTX indication bits

This step of inserting DTX indication bits is used only if the positions of the TrCHs in the radio frame are fixed. With fixed position scheme a fixed number of bits is reserved for each TrCH in the radio frame.

The bits from rate matching are denoted by $g_{i1}, g_{i2}, g_{i3}, \dots, g_{iG_i}$, where G_i is the number of bits in one TTI of TrCH *i*. Denote the number of bits in one radio frame of TrCH *i* by H_i . Denote D_i the number of bits output of the first DTX insertion block.

In <u>TTIs containing no compressed frames or frames normal or compressed mode</u> by spreading factor reduction, H_i is constant and corresponds to the maximum number of bits from TrCH *i* in one radio frame for any transport format of TrCH *i*- and $D_i = F_i \succeq^{\pm} H_i$.

In <u>TTIs containing frames</u> compressed <u>mode</u> by puncturing, additional puncturing is performed in the rate matching block. The empty positions resulting from the additional puncturing are used to insert p-bits in the first interleaving block, the DTX insertion is therefore limited to allow for later insertion of p-bits. Thus DTX bits are inserted until the total number of bits is D_i where $D_i = F_i \ge \# H_{i,*} - Np^{TTI, m}_{i,max}$ and $H_i = N_{i,*} + \Delta N_{i,*}$.

The bits output from the DTX insertion are denoted by h_{il} , h_{i2} , h_{i3} , ..., h_{iDi2} Note that these bits are three valued. They are defined by the following relations:

$$h_{ik} = g_{ik} \ k = 1, 2, 3, ..., G_i$$

$$h_{ik} = \delta$$
 $k = G_i + 1, G_i + 2, G_i + 3, ..., D_i$

where DTX indication bits are denoted by δ . Here $g_{ik} \in \{0, 1\}$ and $\delta \notin \{0, 1\}$.

4.2.9.2 2nd insertion of DTX indication bits

The DTX indication bits inserted in this step shall be placed at the end of the radio frame. Note that the DTX will be distributed over all slots after 2nd interleaving.

The bits input to the DTX insertion block are denoted by $s_1, s_2, s_3, \ldots, s_s$, where S is the number of bits from TrCH multiplexing. The number of PhCHs is denoted by P and the number of bits in one radio frame, including DTX indication bits, for each PhCH by R..

In normal modenon-compressed frames,
$$R = \frac{N_{data,*}}{P} = 15 \times (N_{data1} + N_{data2})$$
, where N_{data1} and N_{data2} are defined in

[2].

For compressed <u>modeframes</u>, $N'_{data,*}$ is defined as $N'_{data,*} = P \times 15 \times (N'_{data1} + N'_{data2})$. N'_{data1} and N'_{data2} are the number of bits in the data fields of the slot format used for the current compressed <u>modeframe</u>, i.e. slot format A or B as defined in [2] corresponding to the <u>sSpreading fFactor</u> and the number of transmitted slots in use.

In <u>case of frames</u> compressed <u>mode</u> by puncturing and <u>when</u> fixed positions <u>are used</u>, DTX shall be inserted until $N'_{data,*}$ bits, since the exact room for the gap is already reserved thanks to the earlier insertion of the p-bits. Therefore R is defined as $R = N'_{data,*}/P$.

In <u>frames</u> compressed <u>mode</u> by higher layer scheduling, additional DTX with respect to normal mode shall be inserted if the transmission time reduction by higher layer scheduling does not exactly create a transmission gap of the desired *TGL*.

The number of bits available to the CCTrCH in one radio frame in compressed mode-by SF spreading factor reduction and <u>or</u> by higher layer scheduling is denoted by $N_{data,*}^{cm}$ and $R = \frac{N_{data,*}^{cm}}{P}$.

For frames compressed by the transmission time spreading factor reduction by SF/2 method in compressed mode N'_{data} *

$$N_{data,*}^{cm} = \frac{1 \sqrt{data,*}}{2}.$$

For <u>frames</u> compressed <u>mode</u> by higher layer scheduling the exact value of $N_{data,*}^{cm}$ is dependent on the *TGL* which is signalled from higher layers. It can be calculated as $N_{data,*}^{cm} = N_{data,*}^{'} - N_{TGL}$.

 N_{TGL} is the number of bits that are located within the transmission gap and defined as:

$$N_{TGL} = \begin{cases} \frac{TGL}{15} \times N'_{data,*}, \text{ if } N_{first} + TGL \le 15 \\ \frac{15 - N_{first}}{15} \times N'_{data,*}, \text{ in first frame if } N_{first} + TGL > 15 \\ \frac{TGL - (15 - N_{first})}{15} \times N'_{data,*}, \text{ in second frame if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

The bits output from the DTX insertion block are denoted by $w_1, w_2, w_3, \ldots, w_{(PR)}$. Note that these bits are four valued in case of compressed mode by puncturing, and three valued otherwise. They are defined by the following relations:

$$w_k = s_k$$
 k = 1, 2, 3, ..., S
 $w_k = \delta$ k = S+1, S+2, S+3, ..., *P*·*R*

where DTX indication bits are denoted by δ . Here $S_k \in \{0,1, p\}$ and $\delta \notin \{0,1\}$.

4.4 Compressed mode

In compressed mode<u>frames</u>, TGL slots from N_{first} to N_{last} are not used for transmission of data. As illustrated in figure 11, the instantaneous transmit power is increased in the compressed frame in order to keep the quality (BER, FER, etc.) unaffected by the reduced processing gain. The amount of power increase depends on the transmission time reduction method (see subclause 4.4.3). What frames are compressed, are decided by the network. When in compressed mode, compressed frames can occur periodically, as illustrated in figure 11, or requested on demand. The rate and type of compressed frames is variable and depends on the environment and the measurement requirements.

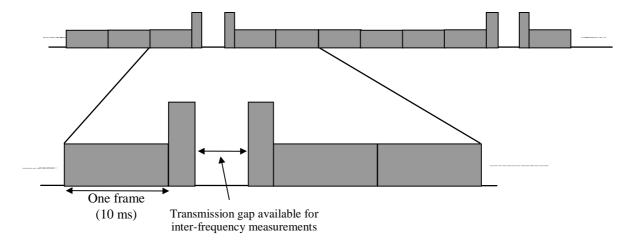


Figure 11: Compressed mode transmission

4.4.1 Frame structure in the uplink

The frame structure for uplink compressed mode frames is illustrated in figure 12.

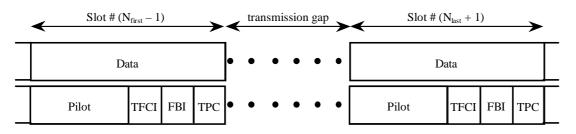


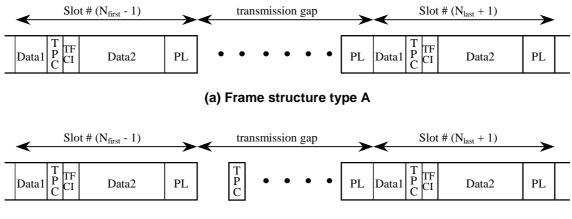
Figure 12: Frame structure in uplink compressed transmission

4.4.2 Frame structure types in the downlink

There are two different types of frame structures defined for downlink compressed <u>modeframes</u>. Type A maximises the transmission gap length and type B is optimised for power control. <u>The frame structure type A or B is set by higher</u> layers independent from the downlink slot format type A or B.

- With frame structure of type A, the pilot field of the last slot in the transmission gap is transmitted. Transmission is turned off during the rest of the transmission gap (figure 13(a)).
- With frame structure of type B, the TPC field of the first slot in the transmission gap and the pilot field of the last slot in the transmission gap is transmitted. Transmission is turned off during the rest of the transmission gap (figure 13(b)).

51



(b) Frame structure type B

Figure 13: Frame structure types in downlink compressed transmission

4.4.3 Transmission time reduction method

When in compressed mode, the information normally transmitted during a 10 ms frame is compressed in time. The mechanisms provided for achieving this are puncturing, reduction of the spreading factor by a factor of two-, and higher layer scheduling. In the downlink, all methods are supported while compressed mode by puncturing is not used in the uplink. The maximum idle length is defined to be 7 slots per one 10 ms frame. The slot formats that are used in compressed mode <u>frames</u> are listed in [2].

4.4.3.1 Compressed mode by puncturing

During compressed mode, rRate matching (puncturing) is applied for -creating <u>a</u> transmission gap in one <u>or two</u> frames. The algorithm for rate matching (puncturing) as described in subclause 4.2.7 is used.

4.4.3.2 Compressed mode by reducing the spreading factor by 2

During compressed mode, t_{T} he spreading factor (SF) can be reduced by 2 during one <u>compressed</u> radio frame to enable the transmission of the information bits in the remaining time slots of <u>a the</u> compressed frame.

On the downlink, UTRAN can also order the UE to use a different scrambling code in <u>a</u> compressed <u>mode_frame</u> than in <u>normal a non-compressed mode_frame</u>. If the UE is ordered to use a different scrambling code in <u>a</u> compressed <u>mode_frame</u>, then there is a one-to-one mapping between the scrambling code used in <u>normal mode_the non-compressed</u> <u>frame</u> and the one used in <u>the compressed mode_frame</u>, as described in <u>TS 25.213</u>[3] subclause 5.2.1.

4.4.3.3 Compressed mode by higher layer scheduling

Compressed <u>mode-frames</u> can be obtained by higher layer scheduling. Higher layers then set restrictions so that only a subset of the allowed TFCs are used in <u>a</u> compressed <u>modeframe</u>. The maximum number of bits that will be delivered to the physical layer during the compressed radio frame is then known and a transmission gap can be generated. Note that in the downlink, the TFCI field is expanded on the expense of the data fields and this shall be taken into account by higher layers when setting the restrictions on the TFCs. Compressed mode by higher layer scheduling shall not be used with fixed starting positions of the TrCHs in the radio frame.

52

3GPP TSG RAN Meeting #9

Document	<i>R1-00-1134</i>

Hawaii, USA,	20-22 Septe	mber 2000		•	_		3GPP use the format T SMG, use the format	
		CHANGE F	REQI	JEST		see embedded help i or instructions on how		
		25.212	CR	085r1		Current Versi	on: <u>3.3.0</u>	
GSM (AA.BB) or 30	G (AA.BBB) specific	ation number ↑		<i>↑ CR</i>	number a	as allocated by MCC	support team	
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Proposed chan (at least one should be		(U)SIM	ME	X U	TRAN	/ Radio X	Core Networl	k 📃
Source:	TSG RAN	WG1				Date:	24-August-2	000
Subject:	Editorial co	prrections in Turbo	code in	ternal inter	leaver	section		
Work item:								
(only one category E shall be marked	CorresponAddition of	ds to a correction i feature modification of fea		rlier releas	e	Release:	Phase 2 Release 96 Release 97 Release 98 Release 99 Release 00	X
<u>Reason for</u> change:		its padding and pro athematical notatio					525.201 Anne:	x A.
Clauses affecte	ed: 4.2.3.	2.3 of TS25.212						
<u>Other specs</u> affected:	Other 3G co Other GSM o specifica MS test spec BSS test specific O&M specific	tions cifications ecifications	-	$\begin{array}{l} \rightarrow \text{ List of C} \\ \rightarrow \text{ List of C} \end{array}$	CRs: CRs: CRs:			
<u>Other</u> comments:								

The initial value of the shift registers of the 8-state constituent encoders shall be all zeros when starting to encode the input bits.

Output from the Turbo coder is

$$x_1, z_1, z'_1, x_2, z_2, z'_2, \ldots, x_K, z_K, z'_K,$$

where $x_1, x_2, ..., x_K$ are the bits input to the Turbo coder i.e. both first 8-state constituent encoder and Turbo code internal interleaver, and *K* is the number of bits, and $z_1, z_2, ..., z_K$ and $z'_1, z'_2, ..., z'_K$ are the bits output from first and second 8-state constituent encoders, respectively.

The bits output from Turbo code internal interleaver are denoted by $x'_1, x'_2, ..., x'_K$, and these bits are to be input to the second 8-state constituent encoder.

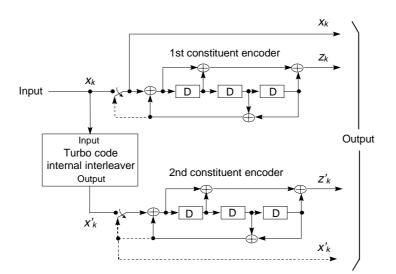


Figure 4: Structure of rate 1/3 Turbo coder (dotted lines apply for trellis termination only)

4.2.3.2.2 Trellis termination for Turbo coder

Trellis termination is performed by taking the tail bits from the shift register feedback after all information bits are encoded. Tail bits are padded after the encoding of information bits.

The first three tail bits shall be used to terminate the first constituent encoder (upper switch of figure 4 in lower position) while the second constituent encoder is disabled. The last three tail bits shall be used to terminate the second constituent encoder (lower switch of figure 4 in lower position) while the first constituent encoder is disabled.

The transmitted bits for trellis termination shall then be:

$$x_{K+1}, z_{K+1}, x_{K+2}, z_{K+2}, x_{K+3}, z_{K+3}, x'_{K+1}, z'_{K+1}, x'_{K+2}, z'_{K+2}, x'_{K+3}, z'_{K+3}$$

4.2.3.2.3 Turbo code internal interleaver

The Turbo code internal interleaver consists of bits-input to a rectangular matrix with padding, intra-row and inter-row permutations of the rectangular matrix, and bits-output from the rectangular matrix with pruning. The bits input to the Turbo code internal interleaver are denoted by $x_1, x_2, x_3, \ldots, x_K$, where *K* is the integer number of the bits and takes one value of $40 \le K \le 5114$. The relation between the bits input to the Turbo code internal interleaver and the bits input to the channel coding is defined by $x_k = o_{irk}$ and $K = K_i$.

The following subclause specific symbols are used in subclauses 4.2.3.2.3.1 to 4.2.3.2.3.3:

 $\frac{KK}{R}$ Number of bits input to Turbo code internal interleaver RR Number of rows of rectangular matrix

€ <u>C</u> ₱₽ ₩v	Number of columns of rectangular matrix Prime number Primitive root
	$))_{j \in \{0,1,\dots,p-2\}}$ Base sequence for intra-row permutation
$\mathbf{q}_{\mathbf{j}} \underline{q}_{\mathbf{j}}$	Minimum prime integers
₽ <u>;</u>	Permuted prime integers
	(i) Inter-row permutation pattern
$\frac{U_{j}(i)}{U}$	$i_i(j)_{j \in \{0,1,\dots,C-1\}}$ Intra-row permutation pattern <u>of <i>i</i>-th row</u>
<u>ii</u>	Index of row number of rectangular matrix
ii ji	Index of <u>column number of rectangular</u> matrix
<u>k</u>	Index of bit sequence
_	-

4.2.3.2.3.1 Bits-input to rectangular matrix with padding

The bit sequence $x_1, x_2, x_3, \dots, x_{K_{-}}$ input to the Turbo code internal interleaver x_k -is written into the rectangular matrix as follows:

(1) Determine the number of rows R-of the rectangular matrix, R, such that:

 $R = \begin{cases} 5, \text{ if } (40 \le K \le 159) \\ 10, \text{ if } ((160 \le K \le 200) \text{ or } (481 \le K \le 530)) \\ 20, \text{ if } (K = \text{ any other value}) \end{cases}$

where tThe rows of rectangular matrix are numbered 0, 1, $\frac{2}{2}$, ..., R - 1 from top to bottom.

(2) Determine the prime number to be used in the intra-permutation, p, and the number of columns C-of rectangular matrix, C, such that:

if $(481 \le K \le 530)$ then

$$p = 53$$
 and $C = p$.

else

Find minimum prime <u>number p from table 2</u> such that

 $(p+1) \quad K/R \geq 0 \quad K \leq R \times (p+1),$

and determine C such that

$$C = \begin{cases} p - 1 & \text{if } K \le R \times (p - 1) \\ p & \text{if } R \times (p - 1) < K \le R \times p \\ p + 1 & \text{if } R \times p < K \end{cases}$$

```
if (p - K/R \ge 0) then
```

```
if (p - 1 - K/R \ge 0) then
```

C = p - 1.

else

C = I

end if

else

C = p + 1

18

end if

end if

where t<u>T</u>he columns of rectangular matrix are numbered 0, $1, \frac{2}{2}, ..., C$ - 1 from left to right.

<u>p</u>	<u>v</u>	<u>p</u>	<u>v</u>	р	<u>v</u>	P	<u>v</u>	<u>p</u>	<u>v</u>
<u>7</u>	<u>3</u>	<u>47</u>	<u>5</u>	<u>101</u>	<u>2</u>	<u>157</u>	<u>5</u>	<u>223</u>	<u>3</u>
<u>11</u>	2	<u>53</u>	2	<u>103</u>	<u>5</u>	<u>163</u>	2	<u>227</u>	2
<u>13</u>	<u>2</u>	<u>59</u>	2	<u>107</u>	<u>2</u>	<u>167</u>	<u>5</u>	<u>229</u>	<u>6</u>
<u>17</u>	<u>3</u>	<u>61</u>	<u>2</u>	<u>109</u>	<u>6</u>	<u>173</u>	<u>2</u>	<u>233</u>	<u>3</u>
<u>19</u>	<u>2</u>	<u>67</u>	2	<u>113</u>	<u>3</u>	<u>179</u>	2	<u>239</u>	<u>7</u>
<u>23</u>	<u>5</u>	<u>71</u>	<u>7</u>	<u>127</u>	<u>3</u>	<u>181</u>	<u>2</u>	<u>241</u>	<u>7</u>
<u>29</u>	<u>2</u>	<u>73</u>	<u>5</u>	<u>131</u>	<u>2</u>	<u>191</u>	<u>19</u>	<u>251</u>	<u>6</u>
<u>31</u>	<u>3</u>	<u>79</u>	3	<u>137</u>	<u>3</u>	<u>193</u>	<u>5</u>	<u>257</u>	<u>3</u>
<u>37</u>	<u>2</u>	<u>83</u>	<u>2</u>	<u>139</u>	<u>2</u>	<u>197</u>	<u>2</u>		
<u>41</u>	<u>6</u>	<u>89</u>	<u>3</u>	<u>149</u>	<u>2</u>	<u>199</u>	<u>3</u>		
<u>43</u>	<u>3</u>	<u>97</u>	<u>5</u>	<u>151</u>	<u>6</u>	<u>211</u>	<u>2</u>		

Table 2: List of prime number p and associated primitive root v

(3) Write the input bit sequence $x_k x_1, x_2, x_3, \dots, x_K$ into the $R \times C$ rectangular matrix row by row starting with bit xy_1 in column 0 of row 0:

	x_1	<i>x</i> ₂	<i>x</i> ₃	$\ldots x_C$	y ₁	<i>y</i> ₂	<i>y</i> ₃	$\cdots y_C$
	<i>x</i> _(C+1)	$x_{(C+2)}$	$x_{(C+3)}$	$\dots x_{2C}$	<i>y</i> _(C+1)	$y_{(C+2)}$	$y_{(C+3)}$	y _{2C}
1	:	÷	:	:		÷	÷	:
	$x_{((R-1)C+1)}$	$x_{((R-1)C+2)}$	$x_{((R-1)C+3)}$	$\dots x_{RC}$	$y_{((R-1)C+1)}$	$y_{((R-1)C+2)}$	$y_{((R-1)C+3)}$	$\cdots y_{R \times C}$

where $y_k = x_k$ for k = 1, 2, ..., K and if $R \times C > K$, the dummy bits are padded such that $y_k = 0 or1$ for $k = K + 1, K + 2, ..., R \times C$. These dummy bits are pruned away from the output of the rectangular matrix after intra-row and interrow permutations.

4.2.3.2.3.2 Intra-row and inter-row permutations

After the bits-input to the $R \times C$ rectangular matrix, the intra-row and inter-row permutations for the $R \times C$ rectangular matrix are performed <u>stepwise</u> by using the following algorithm with steps (1) – (6):-

(1) Select a primitive root v from table 2 in section 4.2.3.2.3.1, which is indicated on the right side of the prime number \underline{p} .

(2) Construct the base sequence $\frac{s(i)}{s(s(j))}_{j \in \{0,1,\dots,p-2\}}$ for intra-row permutation as:

 $\frac{s(i) = [v \times s(i-1)] \mod p}{s(j) = (v \times s(j-1)) \mod p}, \frac{ij}{i} = 1, 2, \dots, (p-2), \text{ and } s(0) = 1.$

(3) Let<u>Assign</u> $q_0 = 1$ to be the first prime integer in $\{q_j\}$ the sequence $\langle q_i \rangle_{i \in \{0,1,\dots,R-1\}}$, and selectdetermine the consecutive minimum prime integers $\{q_{j\bar{\ell}}\}$ in the sequence $\langle q_i \rangle_{i \in \{0,1,\dots,R-1\}}$. $(j = 1, 2, \dots, R-1)$ to be a least prime integer such that:

g.c.d $(q_{jj}, p - 1) = 1, q_{ji} > 6$, and $q_{jj} > q_{(jj-1)}$, for each i = 1, 2, ..., R - 1.

Here where g.c.d. is greatest common divisor.

(4) Permute $\frac{\{q_i\}}{\text{the sequence}} \langle q_i \rangle_{i \in \{0,1,\dots,R-1\}}$ to make $\frac{\{r_i\}}{\text{the sequence}} \langle r_i \rangle_{i \in \{0,1,\dots,R-1\}}$ such that

 $r_{T(\underline{j}\underline{i})} = q_{\underline{j}\underline{i}}, \ \underline{j}\underline{i} = 0, 1, \ \dots, R - 1,$

where T(j) $(j = 0, 1, 2, ..., R-1) \langle T(i) \rangle_{i \in \{0, 1, \dots, R-1\}}$ is the inter-row permutation pattern defined as the one of the

following four kind of patterns, which are shown in table 3,: $Pat_1, Pat_2, Pat_3, and Pat_4$ depending on the number of input bits *K*.

$$\left\{ \begin{aligned} &Pat_4 & \text{if} (40 \le K \le 159) \\ &Pat_3 & \text{if} (160 \le K \le 200) \\ &Pat_1 & \text{if} (201 \le K \le 480) \\ &Pat_3 & \text{if} (481 \le K \le 530) \\ &Pat_1 & \text{if} (531 \le K \le 2280) \\ &Pat_2 & \text{if} (2281 \le K \le 2480) \\ &Pat_1 & \text{if} (2481 \le K \le 3160) \\ &Pat_2 & \text{if} (3161 \le K \le 3210) \\ &Pat_1 & \text{if} (3211 \le K \le 5114) \end{aligned} \right.$$

where Pat₁. Pat₂. Pat₃ and Pat₄ have the following patterns respectively.

*Pat*₁: {19, 9, 14, 4, 0, 2, 5, 7, 12, 18, 10, 8, 13, 17, 3, 1, 16, 6, 15, 11}

 $Pat_{\underline{2}}: \{19, 9, 14, 4, 0, 2, 5, 7, 12, 18, 16, 13, 17, 15, 3, 1, 6, 11, 8, 10\}$

Pat₃: {9, 8, 7, 6, 5, 4, 3, 2, 1, 0}

 $Pat_4: \{4, 3, 2, 1, 0\}$

Table 3: Inter-row permutation patterns for Turbo code internal interleaver

Number of input bits <u>K</u>	<u>Number</u> of rows <i>R</i>	$\frac{\text{Inter-row permutation patterns}}{\leq T(0), T(1),, T(R-1) >}$
<u>(40</u> ≤ <u>K</u> ≤ <u>159)</u>	<u>5</u>	<u><4, 3, 2, 1, 0></u>
<u>(160≤K≤200) or (481≤K≤530)</u>	<u>10</u>	<u><9, 8, 7, 6, 5, 4, 3, 2, 1, 0></u>
$(2281 \le K \le 2480)$ or $(3161 \le K \le 3210)$	<u>20</u>	<u><19, 9, 14, 4, 0, 2, 5, 7, 12, 18, 16, 13, 17, 15, 3, 1, 6, 11, 8, 10></u>
<u>K = any other value</u>	20	<u><19, 9, 14, 4, 0, 2, 5, 7, 12, 18, 10, 8, 13, 17, 3, 1, 16, 6, 15, 11></u>

(5) Perform the \underline{ji} -th ($\underline{ji} = 0, 1, \frac{2}{2}, \dots, R - 1$) intra-row permutation as:

if
$$(C = p)$$
 then

 $\frac{U_{j}(i) = s([i \times r_{j}] \mod (p-1))}{U_{i}(j)} U_{i}(j) = s((j \times r_{i}) \mod (p-1)), \quad ij = 0, 1, \frac{2}{2}, \dots, (p-2), \frac{1}{2}, \dots, (p-2), \dots, (p-2), \frac{1}{2}, \dots, (p-2), \dots, (p-2), \dots, (p-2),$

where $U_{ij}(ij)$ is the inputoriginal bit position of ij-th output after the permutation permuted bit of ji-th row.

end if

if (C = p + 1) then

$$U_{j(i)} = s([i \times r_j] \mod (p-1)) U_i(j) = s((j \times r_i) \mod (p-1)), \quad ij = 0, 1, 2, \dots, (p-2), U_{j(p-1)} = 0, \text{ and } U_{j(p)} = p, i = 0, 1, 2, \dots, (p-2), U_{j(p-1)} = 0, i = 0, \dots, (p-2), U_{j(p-1)} = 0, \dots, (p-2), \dots, (p-$$

where $U_{ii}(ij)$ is the input original bit position of ij-th output after the permutation permuted bit of jj-th row, and

if $(K = \frac{C \times R \times C}{C})$ then

Exchange $U_{R-1}(p)$ with $U_{R-1}(0)$.

end if

end if

if
$$(C = p - 1)$$
 then

 $\frac{U_{i}(i) = s([i \times r_{i}] \mod (p-1)) - 1}{U_{i}(j)} = s((j \times r_{i}) \mod (p-1)) - 1, \quad \underline{ij} = 0, 1, \underline{-2}, \dots, (p-2),$

where $U_{ij}(ij)$ is the input-original bit position of ij-th output after the permutation permuted bit of ji-th row.

end if

(6) Perform the inter-row permutation for the rectangular matrix based on the pattern T(j) (j = 0, 1, 2, ..., R-1) $T(i)_{i \in \{0, 1, \dots, R-1\}}$,

where $T(\underline{j}\underline{i})$ is the original row position of the $\underline{j}\underline{i}$ -th permuted row.

₽	¥	þ	¥	Ð	¥	₽	¥	Ð	¥
7	3	47	5	101	2	157	5	223	3
11	2	53	2	103	5	163	2	227	2
13	2	59	2	107	2	167	5	229	6
17	3	61	2	109	¢	173	2	233	ሳ
19	2	67	2	113	4	179	2	239	7
23	5	71	7	127	З	181	2	241	7
29	2	73	5	131	2	191	19	251	ф
31	3	79	З	137	З	193	5	257	З
37	2	83	2	139	2	197	2		
41	6	89	ф	149	2	199	ማ		
43	3	97	5	151	6	211	2		

Table 2: Table of prime p and associated primitive root v

4.2.3.2.3.3 Bits-output from rectangular matrix with pruning

After intra-row and inter-row permutations, the bits of the permuted rectangular matrix are denoted by y'k:

$$\begin{bmatrix} y'_1 & y'_{(R+1)} & y'_{(2R+1)} & \cdots & y'_{((C-1)R+1)} \\ y'_2 & y'_{(R+2)} & y'_{(2R+2)} & \cdots & y'_{((C-1)R+2)} \\ \vdots & \vdots & \vdots & & \vdots \\ y'_R & y'_{2R} & y'_{3R} & \cdots & y'_{CR} \end{bmatrix} \begin{bmatrix} y'_1 & y'_{(R+1)} & y'_{(2R+1)} & \cdots & y'_{((C-1)R+1)} \\ y'_2 & y'_{(R+2)} & y'_{(2R+2)} & \cdots & y'_{((C-1)R+2)} \\ \vdots & \vdots & \vdots & & \vdots \\ y'_R & y'_{2R} & y'_{3R} & \cdots & y'_{CXR} \end{bmatrix}$$

The output of the Turbo code internal interleaver is the bit sequence read out column by column from the intra-row and inter-row permuted $R \times C$ rectangular matrix starting with bit y'_1 in row 0 of column 0 and ending with bit y'_{CR} in row R - 1 of column C - 1. The output is pruned by deleting <u>dummy</u> bits that were not presentpadded into the input bit sequence of the rectangular matrix before intra-row and inter row permutations, i.e. bits y'_k that corresponds to bits xy_k with k > K are removed from the output. The bits output from Turbo code internal interleaver are denoted by $x'_1, x'_2, ..., x'_K$, where x'_1 corresponds to the bit y'_k with smallest index k after pruning, x'_2 to the bit y'_k with second smallest index k after pruning, and so on. The number of bits output from Turbo code internal interleaver is K and the total number of pruned bits is:

$$R \times C - K$$
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4.2.12 Physical channel mapping

The PhCH for both uplink and downlink is defined in [2]. The bits input to the physical channel mapping are denoted by $v_{p1}, v_{p2}, \dots, v_{pU}$, where *p* is the PhCH number and *U* is the number of bits in one radio frame for one PhCH. The bits v_{pk} are mapped to the PhCHs so that the bits for each PhCH are transmitted over the air in ascending order with respect to *k*.

In compressed mode, no bits are mapped to certain slots of the PhCH(s). If $N_{first} + TGL \le 15$, no bits are mapped to slots N_{first} to N_{last} . If $N_{first} + TGL > 15$, i.e. the transmission gap spans two consecutive radio frames, the mapping is as follows:

- In the first radio frame, no bits are mapped to slots N_{first} , $N_{first}+1$, $N_{first}+2$, ..., 14.
- In the second radio frame, no bits are mapped to the slots 0, 1, 2, ..., N_{last} .

TGL, N_{first} , and N_{last} are defined in subclause 4.4.

4.2.12.1 Uplink

In uplink, the PhCHs used during a radio frame are either completely filled with bits that are transmitted over the air or not used at all. The only exception is when the UE is in compressed mode. The transmission can then be turned off during consecutive slots of the radio frame.

4.2.12.2 Downlink

In downlink, the PhCHs do not need to be completely filled with bits that are transmitted over the air. Bits $v_{pk} \notin \{0, 1\}$ are not transmitted.

During compressed mode by reducing the spreading factor by 2, <u>the data bits are always mapped into 7.5 slots within a</u> <u>compressed frame.</u> **n**<u>N</u>o bits are mapped to the DPDCH field as follows:</u>

If $N_{first} + TGL \le 15$, i.e. the transmission gap spans one radio frame,

if
$$N_{first} + 7 \le 14$$

no bits are mapped to slots $N_{first}, N_{first} + 1, N_{first} + 2, \dots, N_{first} + 6N_{hast} + (7 - TGL)$

no bits are mapped to the first $(N_{Data1} + N_{Data2})/2$ bit positions of slot $N_{first} + 7N_{last} + (8 - TGL)$

else

no bits are mapped to slots N_{first} , N_{first} + 1, N_{first} + 2,..., 14

no bits are mapped to slots N_{first} - 1, N_{first} - 2, N_{first} - 3, ..., <u>8</u> N_{first} (7 TGL (14 N_{last}))

no bits are mapped to the last $(N_{Data1} + N_{Data2})/2$ bit positions of slot $\frac{7N_{first}}{(8 TGL (14 N_{last}))}$

end if

If $N_{first} + TGL > 15$, i.e. the transmission gap spans two consecutive radio frames,

In the first radio frame, no bits are mapped to last $(N_{Data1} + N_{Data2})/2$ bit positions in slot 7 as well as to slots 8, 9, 10, ..., 14.

In the second radio frame, no bits are mapped to slots 0, 1, 2, ..., 6 as well as to first $(N_{Data1} + N_{Data2})/2$ bit positions in slot 7.

N_{Data1}and N_{Data2} are defined in [2].

4.2.13 Restrictions on different types of CCTrCHs

Restrictions on the different types of CCTrCHs are described in general terms in TS 25.302[11]. In this subclause those restrictions are given with layer 1 notation.

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4.2.14 Multiplexing of different transport channels into one CCTrCH, and mapping of one CCTrCH onto physical channels

The following rules shall apply to the different transport channels which are part of the same CCTrCH:

 Transport channels multiplexed into one CCTrCh shall have co-ordinated timings. When the TFCS of a CCTrCH is changed because one or more transport channels are added to the CCTrCH or reconfigured within the CCTrCH, or removed from the CCTrCH, the change may only be made at the start of a radio frame with CFN fulfilling the relation

CFN mod $F_{max} = 0$,

where F_{max} denotes the maximum number of radio frames within the transmission time intervals of all transport channels which are multiplexed into the same CCTrCH, including any transport channels *i* which are added, reconfigured or have been removed, and CFN denotes the connection frame number of the first radio frame of the changed CCTrCH.

After addition or reconfiguration of a transport channel *i* within a CCTrCH, the TTI of transport channel *i* may only start in radio frames with CFN fulfilling the relation:

 $CFN_i \mod F_i = 0.$

- 2) Only transport channels with the same active set can be mapped onto the same CCTrCH.
- 3) Different CCTrCHs cannot be mapped onto the same PhCH.
- 4) One CCTrCH shall be mapped onto one or several PhCHs. These physical channels shall all have the same SF.
- 5) Dedicated Transport channels and common transport channels cannot be multiplexed into the same CCTrCH.
- 6) For the common transport channels, only the FACH and PCH may belong to the same CCTrCH.

There are hence two types of CCTrCH:

- 1) CCTrCH of dedicated type, corresponding to the result of coding and multiplexing of one or several DCHs.
- 2) CCTrCH of common type, corresponding to the result of the coding and multiplexing of a common channel, RACH in the uplink, DSCH ,BCH, or FACH/PCH for the downlink.

4.2.14.1 Allowed CCTrCH combinations for one UE

4.2.14.1.1 Allowed CCTrCH combinations on the uplink

A maximum of one CCTrCH is allowed for one UE on the uplink. It can be either:

- 1) one CCTrCH of dedicated type;
- 2) one CCTrCH of common type.

4.2.14.1.2 Allowed CCTrCH combinations on the downlink

The following CCTrCH combinations for one UE are allowed:

- x CCTrCH of dedicated type + y CCTrCH of common typeThe allowed combination of CCTrCHs of dedicated and common type are given from UE radio access capabilities. There can be a maximum on one CCTrCH of common type for DSCH and a maximum of one CCTrCH of common type for FACH. With one CCTrCH of common type for DSCH, there shall be at least one CCTrCH of dedicated type.
- NOTE 1: There is only one DPCCH in the uplink, hence one TPC bits flow on the uplink to control possibly the different DPDCHs on the downlink, part of the same or several CCTrCHs.

4.3.3 Coding of Transport-Format-Combination Indicator (TFCI)

The TFCI is encoded using a (32, 10) sub-code of the second order Reed-Muller code. The coding procedure is as shown in figure 9.

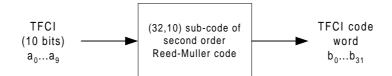


Figure 9: Channel coding of TFCI information bits

If the TFCI consist of less than 10 bits, it is padded with zeros to 10 bits, by setting the most significant bits to zero. The length of the TFCI code word is 32 bits.

The code words of the (32,10) sub-code of second order Reed-Muller code are linear combination of 10 basis sequences. The basis sequences are as in the following table 7.

i	M i,0	M i,1	M i,2	M _{i,3}	M i,4	M i,5	M i,6	M i,7	M i,8	M _{i,9}
0	1	0	0	0	0	1	0	0	0	0
1	0	1	0	0	0	1	1	0	0	0
2	1	1	0	0	0	1	0	0	0	1
3	0	0	1	0	0	1	1	0	1	1
4	1	0	1	0	0	1	0	0	0	1
5	0	1	1	0	0	1	0	0	1	0
6	1	1	1	0	0	1	0	1	0	0
7	0	0	0	1	0	1	0	1	1	0
8	1	0	0	1	0	1	1	1	1	0
9	0	1	0	1	0	1	1	0	1	1
10	1	1	0	1	0	1	0	0	1	1
11	0	0	1	1	0	1	0	1	1	0
12	1	0	1	1	0	1	0	1	0	1
13	0	1	1	1	0	1	1	0	0	1
14	1	1	1	1	0	1	1	1	1	1
15	1	0	0	0	1	1	1	1	0	0
16	0	1	0	0	1	1	1	1	0	1
17	1	1	0	0	1	1	1	0	1	0
18	0	0	1	0	1	1	0	1	1	1
19	1	0	1	0	1	1	0	1	0	1
20	0	1	1	0	1	1	0	0	1	1
21	1	1	1	0	1	1	0	1	1	1
22	0	0	0	1	1	1	0	1	0	0
23	1	0	0	1	1	1	1	1	0	1
24	0	1	0	1	1	1	1	0	1	0
25	1	1	0	1	1	1	1	0	0	1
26	0	0	1	1	1	1	0	0	1	0
27	1	0	1	1	1	1	1	1	0	0
28	0	1	1	1	1	1	1	1	1	0
29	1	1	1	1	1	1	1	1	1	1
30	0	0	0	0	0	1	0	0	0	0
31	0	0	0	0	1	1	1	0	0	0

Table 7: Basis sequences for (32,10) TFCI code

Let's define t<u>T</u>he TFCI information bits as a_0 , a_1 , a_2 , a_3 , a_4 , a_5 , a_6 , a_7 , a_8 , a_9 (where a_0 is LSB and a_9 is MSB). The TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the CCTrCH in the associated DPCH radio frame.

The output code word bits b_i are given by:

$$b_i = \sum_{n=0}^{9} (a_n \times M_{i,n}) \mod 2$$

where i = 0, ..., 31.

The output bits are denoted by b_k , k = 0, 1, 2, ..., 31.

In downlink, when the SF < 128 the encoded TFCI code words are repeated yielding 8 encoded TFCI bits per slot in normal mode and 16 encoded TFCI bits per slot in compressed mode. Mapping of repeated bits to slots is explained in subclause 4.3.5.

4.3.4 Operation of Transport-Format-Combination Indicator (TFCI) in Split Mode

If one of the DCH is associated with a DSCH, the TFCI code word may be split in such a way that the code word relevant for TFCI activity indication is not transmitted from every cell. The use of such a functionality shall be indicated by higher layer signalling.

The TFCI is encoded using a (16, 5) bi-orthogonal (or first order Reed-Muller) code. The coding procedure is as shown in figure 10.

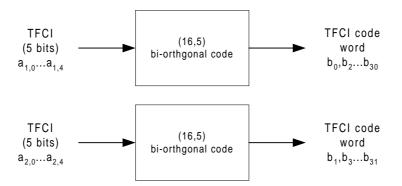


Figure 10: Channel coding of split mode TFCI information bits

The code words of the (16,5) bi-orthogonal code are linear combinations of 5 basis sequences as defined in table 8.

i	M i,0	M i,1	M i,2	M i,3	M i,4
0	1	0	0	0	1
1	0	1	0	0	1
2	1	1	0	0	1
3	0	0	1	0	1
4	1	0	1	0	1
5	0	1	1	0	1
6	1	1	1	0	1
7	0	0	0	1	1
8	1	0	0	1	1
9	0	1	0	1	1
10	1	1	0	1	1
11	0	0	1	1	1
12	1	0	1	1	1
13	0	1	1	1	1
14	1	1	1	1	1
15	0	0	0	0	1

Table 8: Basis sequences for (16,5) TFCI code

Let's define a <u>The</u> first set of TFCI information bits as $(a_{1,0}, a_{1,1}, a_{1,2}, a_{1,3}, a_{1,4}, (where a_{1,0} is LSB and a_{1,4} is MSB). This set of TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the DCH CCTrCH in the associated DPCH radio frame.$

Let's define a<u>The</u> second set of TFCI information bits as $(a_{2,0}, a_{2,1}, a_{2,2}, a_{2,3}, a_{2,4}, (where a_{2,0} is LSB and a_{2,4} is MSB). This set of TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the associated DSCH CCTrCH in the corresponding PDSCH radio frame.$

The output code word bits b_k are given by:

$$b_{2i} = \sum_{n=0}^{4} (a_{1,n} \times M_{i,n}) \mod 2;$$
 $b_{2i+1} = \sum_{n=0}^{4} (a_{2,n} \times M_{i,n}) \mod 2$

where i = 0, ..., 15.

The output bits are denoted by b_k , k = 0, 1, 2, ..., 31.

4.3.5 Mapping of TFCI words

4.3.5.1 Mapping of TFCI word in normal mode

The bits of the code word are directly mapped to the slots of the radio frame. Within a slot the bit with lower index is transmitted before the bit with higher index. The coded bits b_k , are mapped to the transmitted TFCI bits d_k , according to the following formula:

 $d_k = b_{k \bmod 32}$

For uplink physical channels regardless of the SF and downlink physical channels, if SF \ge 128, k = 0, 1, 2, ..., 29. Note that this means that bits b_{30} and b_{31} are not transmitted.

For downlink physical channels whose SF < 128, k = 0, 1, 2, ..., 119. Note that this means that bits b_0 to b_{23} are transmitted four times and bits b_{24} to b_{31} are transmitted three times.

4.3.5.2 Mapping of TFCI word in compressed mode

The mapping of the TFCI bits in compressed mode is different for uplink, downlink with $SF \ge 128$ and downlink with SF < 128.

4.3.5.2.1 Uplink compressed mode

For uplink compressed mode, the slot format is changed so that no TFCI coded bits are lost. The different slot formats in compressed mode do not match the exact number of TFCI coded bits for all possible TGLs. Repetition of the TFCI bits is therefore used.

Denote the number of bits available in the TFCI fields of one compressed radio frame by D and the number of bits in the TFCI field in a slot by N_{TFCI}. The parameter E is used to determine the number of the first TFCI bit to be repeated.

 $E = N_{\text{first}} N_{\text{TFCI}}$, if the start of the transmission gap is allocated to the current frame. E = 0, if the start of the transmission gap is allocated to the previous frame and the end of the transmission gap is allocated to the current frame.

The TFCI coded bits b_k are mapped to the bits in the TFCI fields d_k . The following relations define the mapping for each compressed frame.

 $d_k = b_k$

where k = 0, 1, 2, ..., min (31, D-1).

If D > 32, the remaining positions are filled by repetition (in reversed order):

 $d_{D-k-1} = b_{(E+k) \mod 32}$

where k = 0, ..., D-33.

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<u>Reason for</u> change:	Correct some smaller ambiguities in TS 25.212.							
Clauses affected	L: 4.2.7, 4.2.7.2.1.1, 4.2.9.2, 4.2.10, 4.3.3, 4.3.4							
Affected:	Other 3G core specifications \rightarrow List of CRs:Other GSM core specifications \rightarrow List of CRs:MS test specifications \rightarrow List of CRs:BSS test specifications \rightarrow List of CRs:O&M specifications \rightarrow List of CRs:							
<u>Other</u> comments:								

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4.2.7 Rate matching

Rate matching means that bits on a transport channel are repeated or punctured. Higher layers assign a rate-matching attribute for each transport channel. This attribute is semi-static and can only be changed through higher layer signalling. The rate-matching attribute is used when the number of bits to be repeated or punctured is calculated.

The number of bits on a transport channel can vary between different transmission time intervals. In the downlink the transmission is interrupted if the number of bits is lower than maximum. When the number of bits between different transmission time intervals in uplink is changed, bits are repeated or punctured to ensure that the total bit rate after TrCH multiplexing is identical to the total channel bit rate of the allocated dedicated physical channels.

If no bits are input to the rate matching for all TrCHs within a CCTrCH, the rate matching shall output no bits for all TrCHs within the CCTrCH and no uplink DPDCH will be selected in the case of uplink rate matching.

Notation used in subcaluse 4.2.7 and subclauses:

 $N_{i,j}$: For uplink: Number of bits in a radio frame before rate matching on TrCH *i* with transport format combination *j*.

For downlink: An intermediate calculation variable (not an integer but a multiple of 1/8).

- $N_{i,l}^{TTI}$: Number of bits in a transmission time interval before rate matching on TrCH *i* with transport format *l*. Used in downlink only.
- $\Delta N_{i,j}$: For uplink: If positive number of bits that should be repeated in each radio frame on TrCH *i* with transport format combination *j*.

If negative - number of bits that should be punctured in each radio frame on TrCH i with transport format combination j.

For downlink : An intermediate calculation variable (not an integer but a multiple of 1/8).

 $\Delta N_{i,l}^{TTI}$: If positive - number of bits to be repeated in each transmission time interval on TrCH *i* with transport format *l*.

If negative - number of bits to be punctured in each transmission time interval on TrCH i with transport format l.

Used in downlink only.

 $Np_{i,l}^{TTI,m}$, m=0 to $(F_{max}/F_i) - 1$:Positive or null: number of bits to be removed in TTI number m within the largest TTI, to create the required gaps in the compressed radio frames of this TTI, in case of compressed mode by puncturing, for TrCH i with transport format l. In case of fixed positions and compressed mode by puncturing, this value is noted $Np_{i,max}^{TTI,m}$ since it is calculated for all TrCH with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $Np_{i,l}^n$ n=0 to F_{max} -1:Positive or null: number of bits, in radio frame number n within the largest TTI, corresponding to the gap for compressed mode in this radio frame, for TrCH i with transport format l. The value will be null for the radio frames not overlapping with a transmission gap. In case of fixed positions and compressed mode by puncturing, this value is noted $Np_{i,max}^n$ since it is calculated for all TrCHs with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $N_{TGL}[k]$, k=0 to F_{max} -1 : Positive or null: number of bits in each radio frame corresponding to the gap for compressed mode for the CCTrCH.

- *RM_i*: Semi-static rate matching attribute for transport channel *i*. *RM_i* is provided by higher layers or takes a value as indicated in section 4.2.13.
- *PL:* Puncturing limit for uplink. This value limits the amount of puncturing that can be applied in order to avoid multicode or to enable the use of a higher spreading factor. Signalled from higher layers.
- $N_{data,j}$: Total number of bits that are available for the CCTrCH in a radio frame with transport format combination *j*.
- *I:* Number of TrCHs in the CCTrCH.
- $Z_{i,i}$: Intermediate calculation variable.
- F_i : Number of radio frames in the transmission time interval of TrCH *i*.
- F_{max} Maximum number of radio frames in a transmission time interval used in the CCTrCH :

$$F_{\max} = \max_{1 \le i \le I} F_i$$

- n_i : Radio frame number in the transmission time interval of TrCH i ($0 \le n_i < F_i$).
- *q:* Average puncturing or repetition distance (normalised to only show the remaining rate matching on top of an integer number of repetitions). Used in uplink only.
- P1_{*F*}(n_i): The column permutation function of the 1st interleaver, P1_{*F*}(x) is the original position of column with number x after permutation. P1 is defined on table 3 of section 4.2.5.2 (note that the P1_{*F*} is self-inverse). Used for rate matching in uplink only.
- *S*[*n*]: The shift of the puncturing or repetition pattern for radio frame n_i when $n = P1_{F_i}(n_i)$. Used in uplink only.
- $TF_i(j)$: Transport format of TrCH *i* for the transport format combination *j*.
- TFS(i) The set of transport format indexes *l* for TrCH *i*.
- *TFCS* The set of transport format combination indexes *j*.
- e_{ini} Initial value of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- e_{plus} Increment of variable *e* in the rate matching pattern determination algorithm of subclause4.2.7.5.
- e_{minus} Decrement of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- *b:* Indicates systematic and parity bits

b=1: Systematic bit. x_k in subclause 4.2.3.2.1.

b=2: 1^{st} parity bit (from the upper Turbo constituent encoder). z_k in subcaluse 4.2.3.2.1.

b=3: 2^{nd} parity bit (from the lower Turbo constituent encoder). z'_k in subclause 4.2.3.2.1.

The * (star) notation is used to replace an index x when the indexed variable X_x does not depend on the index x. In the left wing of an assignment the meaning is that " $X_* = Y$ " is equivalent to "**for all** <u>x</u> **do** $X_x = Y$ ". In the right wing of an assignment, the meaning is that " $Y = X_*$ " is equivalent to "**take any** <u>x</u> **and do** $Y = X_x$ ".

The following relations, defined for all TFC *j*, are used when calculating the rate matching parameters:

$$Z_{0,j} = 0$$

$$Z_{i,j} = \left[\frac{\left(\left(\sum_{m=1}^{i} RM_{m} \times N_{m,j} \right) \times N_{data,j} \right)}{\sum_{m=1}^{I} RM_{m} \times N_{m,j}} \right] \text{ for all } i = 1 \dots I$$
(1)

 $\Delta N_{i,j} = Z_{i,j} - Z_{i-1,j} - N_{i,j} \text{ for all } i = 1 \ \dots \ I$

30

4.2.7.2.1.1 Calculation of $\Delta N_{i,max}$ for normal mode and compressed mode by higher layer scheduling and spreading factor reduction

First an intermediate calculation variable $N_{i,*}$ is calculated for all transport channels *i* by the following formula:

$$N_{i,*} = \frac{1}{F_i} \times \left(\max_{l \in TFS(i)} N_{i,l}^{TTI} \right)$$

In order to compute the $\Delta N_{i,l}^{TTI}$ parameters for all TrCH *i* and all TF *l*, we first compute an intermediate parameter $\Delta N_{i,\text{max}}$ by the following formula, where $\Delta N_{i,*}$ is derived from $N_{i,*}$ by the formula given at subclause 4.2.7:

 $\Delta N_{i,max} = F_i \times \Delta N_{i,*}$

If $\Delta N_{i,max} = 0$ then, for TrCH *i*, the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed. In this case we have :

$$\forall l \in TFS(i) \Delta N_{il}^{TTI} = 0$$

If $\Delta N_{i,max} \neq 0$ the parameters listed in subclauses 4.2.7.2.1.3 and 4.2.7.2.1.4 shall be used for determining e_{ini} , e_{plus} , and e_{minus} and $\Delta N_{i,l}^{TTI}$.

4.2.9.2 2nd insertion of DTX indication bits

The DTX indication bits inserted in this step shall be placed at the end of the radio frame. Note that the DTX will be distributed over all slots after 2^{nd} interleaving.

The bits input to the DTX insertion block are denoted by $s_1, s_2, s_3, \ldots, s_s$, where *S* is the number of bits from TrCH multiplexing. The number of PhCHs is denoted by *P* and the number of bits in one radio frame, including DTX indication bits, for each PhCH by *R*.

In normal mode
$$R = \frac{N_{data,*}}{P} = 15 \times (N_{data1} + N_{data2})$$
, where N_{data1} and N_{data2} are defined in [2].

For compressed mode, $N'_{data,*}$ is defined as $N'_{data,*} = P \times 15 \times (N'_{data1} + N'_{data2})$. N'_{data1} and N'_{data2} are the number of bits in the data fields of the slot format used for the current compressed mode, i.e. slot format A or B as defined in [2] corresponding to the Spreading Factor and the number of transmitted slots in use.

In <u>case of frames</u> compressed <u>mode</u> by puncturing and <u>when</u> fixed positions <u>are used</u>, <u>no</u> DTX shall be inserted-<u>until</u> $\frac{N'_{deta,*}}{N'_{deta,*}}$, since the exact room for the gap is already reserved thanks to the earlier insertion of the p-bits. Therefore R is defined as $R = N'_{deta,*}/P$.

In <u>frames</u> compressed <u>mode</u> by higher layer scheduling, additional DTX with respect to normal mode shall be inserted if the transmission time reduction by higher layer scheduling does not exactly create a transmission gap of the desired *TGL*. The number of bits available to the CCTrCH in one radio frame in compressed mode by SF reduction and by

higher layer scheduling is denoted by $N_{data,*}^{cm}$ and $R = \frac{N_{data,*}^{cm}}{P}$.

For the transmission time reduction by SF/2 method in compressed mode $N_{data,*}^{cm} = \frac{N_{data,*}^{*}}{2}$

For compressed mode by higher layer scheduling the exact value of $N_{data,*}^{cm}$ is dependent on the *TGL* which is signalled from higher layers. It can be calculated as $N_{data,*}^{cm} = N_{data,*}^{'} - N_{TGL}$.

 N_{TGL} is the number of bits that are located within the transmission gap and defined as:

$$N_{TGL} = \begin{cases} \frac{TGL}{15} \times N'_{data,*}, \text{ if } N_{first} + TGL \le 15\\ \frac{15 - N_{first}}{15} \times N'_{data,*}, \text{ in first frame if } N_{first} + TGL > 15\\ \frac{TGL - (15 - N_{first})}{15} \times N'_{data,*}, \text{ in second frame if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

The bits output from the DTX insertion block are denoted by $W_1, W_2, W_3, \dots, W_{(PR)}$. Note that these bits are four valued in case of compressed mode by puncturing, and three valued otherwise. They are defined by the following relations:

$$w_k = s_k$$
 k = 1, 2, 3, ..., S
 $w_k = \delta$ k = S+1, S+2, S+3, ..., *P*·*R*

where DTX indication bits are denoted by δ . Here $S_k \in \{0, 1, p\}$ and $\delta \notin \{0, 1\}$.

4.2.10 Physical channel segmentation

When more than one PhCH is used, physical channel segmentation divides the bits among the different PhCHs. The bits input to the physical channel segmentation are denoted by $x_1, x_2, x_3, \ldots, x_X$, where X is the number of bits input to the physical channel segmentation block. The number of PhCHs is denoted by P.

The bits after physical channel segmentation are denoted $u_{p1}, u_{p2}, u_{p3}, \dots, u_{pU}$, where *p* is PhCH number and *U* is the number of bits in one radio frame for each PhCH, i.e. $U = (X - N_{TGL}) / P$ for compressed mode by puncturing, and

$$U = \frac{X}{P}$$
 otherwise. The relation between x_k and u_{pk} is given below.

For all modes, some bits of the input flow are mapped to each code until the number of bits on the code is *U*. For modes other than compressed mode by puncturing, all bits of the input flow are taken to be mapped to the codes. For compressed mode by puncturing, only the bits of the input flow not corresponding to bits p are taken to be mapped to the codes, each bit p is removed to ensure creation the gap required by the compressed mode, as described below.

Bits on first PhCH after physical channel segmentation:

 $u_{l, k} = x_{i, f(k)} k = 1, 2, ..., U$

Bits on second PhCH after physical channel segmentation:

 $u_{2, k} = x_{i, f(k+U)} k = 1, 2, ..., U$

...

Bits on the *P*th PhCH after physical channel segmentation:

 $u_{P, k} = x_{i, f(k+(P-1) \times U)}$ k = 1, 2, ..., U

where *f* is such that :

- for modes other than compressed mode by puncturing, $x_{i,f(k)} = x_{i,k}$, i.e. f(k) = k, for all k.
- for compressed mode by puncturing, bit $u_{1,1}$ corresponds to the bit $x_{i,k}$ with smallest index k when the bits p are not counted, bit $u_{1,2}$ corresponds to the bit $x_{i,k}$ with second smallest index k when the bits p are not counted, and so on for bits $u_{1,3}$... $u_{1, U}$, $u_{2, 1}$, $u_{2, 2}$, ... $u_{2, U}$, ... $u_{P,U}$, $u_{P,U}$, $v_{2, 2}$.

4.3.3 Coding of Transport-Format-Combination Indicator (TFCI)

The TFCI is encoded using a (32, 10) sub-code of the second order Reed-Muller code. The coding procedure is as shown in figure 9.

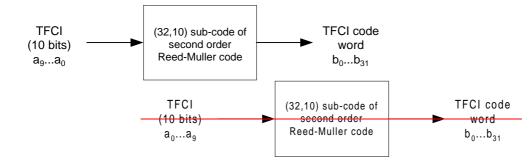


Figure 9: Channel coding of TFCI information bits

If the TFCI consist of less than 10 bits, it is padded with zeros to 10 bits, by setting the most significant bits to zero. The length of the TFCI code word is 32 bits.

The code words of the (32,10) sub-code of second order Reed-Muller code are linear combination of 10 basis sequences. The basis sequences are as in the following table 7.

i	M i,0	M i,1	M i,2	M i,3	M i,4	M i,5	M i,6	M i,7	M i,8	M i,9
0	1	0	0	0	0	1	0	0	0	0
1	0	1	0	0	0	1	1	0	0	0
2	1	1	0	0	0	1	0	0	0	1
3	0	0	1	0	0	1	1	0	1	1
4	1	0	1	0	0	1	0	0	0	1
5	0	1	1	0	0	1	0	0	1	0
6	1	1	1	0	0	1	0	1	0	0
7	0	0	0	1	0	1	0	1	1	0
8	1	0	0	1	0	1	1	1	1	0
9	0	1	0	1	0	1	1	0	1	1
10	1	1	0	1	0	1	0	0	1	1
11	0	0	1	1	0	1	0	1	1	0
12	1	0	1	1	0	1	0	1	0	1
13	0	1	1	1	0	1	1	0	0	1
14	1	1	1	1	0	1	1	1	1	1
15	1	0	0	0	1	1	1	1	0	0
16	0	1	0	0	1	1	1	1	0	1
17	1	1	0	0	1	1	1	0	1	0
18	0	0	1	0	1	1	0	1	1	1
19	1	0	1	0	1	1	0	1	0	1
20	0	1	1	0	1	1	0	0	1	1
21	1	1	1	0	1	1	0	1	1	1
22	0	0	0	1	1	1	0	1	0	0
23	1	0	0	1	1	1	1	1	0	1
24	0	1	0	1	1	1	1	0	1	0
25	1	1	0	1	1	1	1	0	0	1
26	0	0	1	1	1	1	0	0	1	0
27	1	0	1	1	1	1	1	1	0	0
28	0	1	1	1	1	1	1	1	1	0
29	1	1	1	1	1	1	1	1	1	1
30	0	0	0	0	0	1	0	0	0	0
31	0	0	0	0	1	1	1	0	0	0

Table 7: Basis sequences for (32,10) TFCI code

Let's define the TFCI information bits as a_0 , a_1 , a_2 , a_3 , a_4 , a_5 , a_6 , a_7 , a_8 , a_9 (a_0 is LSB and a_9 is MSB). The TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the CCTrCH in the associated DPCH radio frame.

49

The output code word bits b_i are given by:

$$b_i = \sum_{n=0}^{9} (a_n \times M_{i,n}) \mod 2$$

where i = 0, ..., 31.

The output bits are denoted by b_k , k = 0, 1, 2, ..., 31.

In downlink, when the SF < 128 the encoded TFCI code words are repeated yielding 8 encoded TFCI bits per slot in normal mode and 16 encoded TFCI bits per slot in compressed mode. Mapping of repeated bits to slots is explained in subclause 4.3.5.

4.3.4 Operation of Transport-Format-Combination Indicator (TFCI) in Split Mode

If one of the DCH is associated with a DSCH, the TFCI code word may be split in such a way that the code word relevant for TFCI activity indication is not transmitted from every cell. The use of such a functionality shall be indicated by higher layer signalling.

The TFCI is encoded using a (16, 5) bi-orthogonal (or first order Reed-Muller) code. The coding procedure is as shown in figure 10.

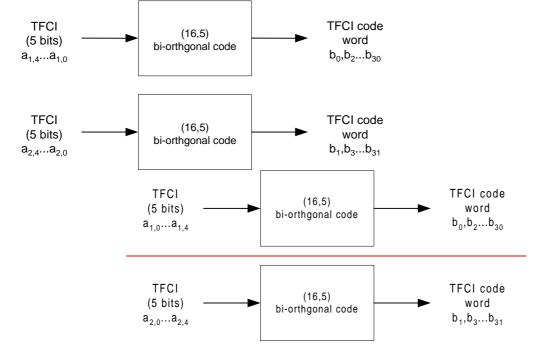


Figure 10: Channel coding of split mode TFCI information bits

The code words of the (16,5) bi-orthogonal code are linear combinations of 5 basis sequences as defined in table 8.

i	M i,0	M i,1	M i,2	M i,3	M i,4
0	1	0	0	0	1
1	0	1	0	0	1
2	1	1	0	0	1
3	0	0	1	0	1
4	1	0	1	0	1
5	0	1	1	0	1
6	1	1	1	0	1
7	0	0	0	1	1
8	1	0	0	1	1
9	0	1	0	1	1
10	1	1	0	1	1
11	0	0	1	1	1
12	1	0	1	1	1
13	0	1	1	1	1
14	1	1	1	1	1
15	0	0	0	0	1

Table 8: Basis sequences for (16,5) TFCI code

Let's define a first set of TFCI information bits as $a_{1,0}$, $a_{1,1}$, $a_{1,2}$, $a_{1,3}$, $a_{1,4}$ ($a_{1,0}$ is LSB and $a_{1,4}$ is MSB). This set of TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the DCH CCTrCH in the associated DPCH radio frame.

Let's define a second set of TFCI information bits as $a_{2,0}$, $a_{2,1}$, $a_{2,2}$, $a_{2,3}$, $a_{2,4}$ ($a_{2,0}$ is LSB and $a_{2,4}$ is MSB). This set of TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the associated DSCH CCTrCH in the corresponding PDSCH radio frame.

The output code word bits b_k are given by:

$$b_{2i} = \sum_{n=0}^{4} (a_{1,n} \times M_{i,n}) \mod 2;$$
 $b_{2i+1} = \sum_{n=0}^{4} (a_{2,n} \times M_{i,n}) \mod 2$

where i = 0, ..., 15.

The output bits are denoted by b_k , k = 0, 1, 2, ..., 31.

3GPP TSG RAN Meeting #9

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4.2.14.1.2 Allowed CCTrCH combinations on the downlink

The following CCTrCH combinations for one UE are allowed:

- x CCTrCH of dedicated type + y CCTrCH of common type. The allowed combination of CCTrCHs of dedicated and common type are given from UE radio access capabilities. There can be a maximum on one CCTrCH of common type for DSCH and a maximum of one CCTrCH of common type for FACH. With one CCTrCH of common type for DSCH, there shall be <u>at leastonly</u> one CCTrCH of dedicated type.
- NOTE 1: There is only one DPCCH in the uplink, hence one TPC bits flow on the uplink to control possibly the different DPDCHs on the downlink, part of the same or several CCTrCHs.
- NOTE 2: There is only one DPCCH in the downlink, even with multiple CCTrCHs. With multiple CCTrCHs, the DPCCH is transmitted on one of the physical channels of that CCTrCH which has the smallest SF among the multiple CCTrCHs. Thus there is only one TPC command flow and only one TFCI word in downlink even with multiple CCTrCHs.

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4.2.13.3 Common Packet Channel (CPCH)

- The maximum value of the number of TrCHs *I* in a CCTrCH, the maximum value of the number of transport blocks *M*, on each transport channel, and the maximum value of the number of DPDCHs *P* are given from the UE capability class.
- NOTE: Only the data part of the CPCH can be mapped on multiple physical channels (this note is taken from TS 25.302).
- There can only be one TrCH in each CPCH CCTrCH, i.e. I=1, $s_k = f_{1k}$ and $S = V_1$.
- The maximum value of the number of transport blocks M₁ on the transport channel is given from the UE capability class.
- Only one PCPCH is used, i.e. P=1, $u_{1k} = s_k$, and U = S.

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4.2.7.3 Bit separation and collection in uplink

The systematic bits of turbo encoded TrCHs shall not be punctured. the other bits , however, systematic bits for trellis termination may be punctured. The systematic bits, first parity bits, and second parity bits in the bit sequence input to the rate matching block are therefore separated into three sequences.

The first sequence contains:

- All of the systematic bits that are from turbo encoded TrCHs.

- From 0 to 2 first and/or second parity bits that are from turbo encoded TrCHs. These bits come into the first sequence when the total number of bits in a block after radio frame segmentation is not a multiple of three.
- Some of the systematic, first parity and second parity bits that are for trellis termination.

The second sequence contains:

- All of the first parity bits that are from turbo encoded TrCHs, except those that go into the first sequence when the total number of bits is not a multiple of three.
- Some of the systematic, first parity and second parity bits that are for trellis termination.

The third sequence contains:

- All of the second parity bits that are from turbo encoded TrCHs, except those that go into the first sequence when the total number of bits is not a multiple of three.
- Some of the systematic, first parity and second parity bits that are for trellis termination.

, one sequence containing all of the systematic bits and some systematic, first and second parity trellis termination bits; the second sequence containing all of the first parity bits and some systematic, first and second parity trellis termination bits and the third sequence containing all of the second parity bits and some systematic, first and second parity trellis termination bits. The second and third sequences shall be of equal length, whereas the first sequence can contain from 0 to 2 more bits. Puncturing is only applied only to the second and third sequences.

The bit separation function is transparent for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition. The bit separation and bit collection are illustrated in figures 5 and 6.

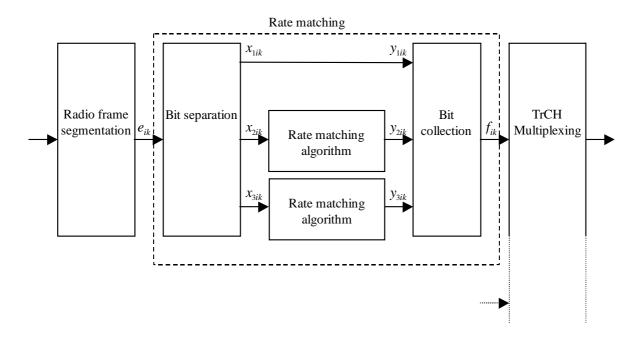


Figure 5: Puncturing of turbo encoded TrCHs in uplink

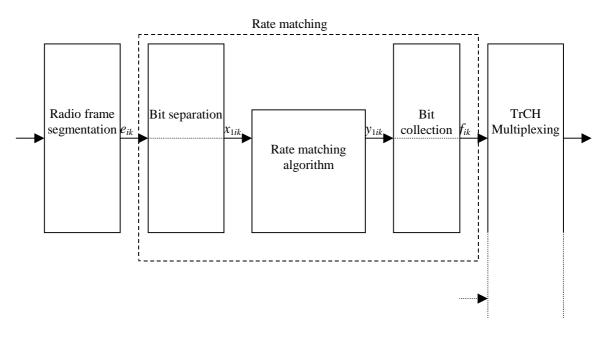


Figure 6: Rate matching for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition in uplink

The bit separation is dependent on the 1st interleaving and offsets are used to define the separation for different TTIs. <u>b</u> indicates the three sequences defined in this section, with <u>b=1</u> indicating the first sequence, b = 2 the second one, and b = 3 the third one. The sequence denoted as b=1 contains all of the systematic bits and some systematic, first and second parity trellis termination bits; the sequence denoted as b=2 contains all of the first parity bits and some systematic, first and second parity trellis termination bits; the

sequence denoted as b=3 contains all of the second parity bits and some systematic, first and second parity trellis termination bits. The offsets α_b for these sequences are listed in table 4.

TTI (ms)	α 1	<i>0</i> (2	0 ⁄3
10, 40	0	1	2
20, 80	0	2	1

Table 4: TTI dependent offset needed for bit separation

The bit separation is different for different radio frames in the TTI. A second offset is therefore needed. The radio frame number for TrCH *i* is denoted by n_i . and the offset by β_{n_i} .

Table 5: Radio frame dependent offset needed for bit separation

TTI (ms)	β₀	β 1	β 2	β 3	β4	β 5	β_6	β 7
10	0	NA	NA	NA	NA	NA	NA	NA
20	0	1	NA	NA	NA	NA	NA	NA
40	0	1	2	0	NA	NA	NA	NA
80	0	1	2	0	1	2	0	1

4.2.7.3.1 Bit separation

The bits input to the rate matching are denoted by $e_{i1}, e_{i2}, e_{i3}, \dots, e_{iN_i}$, where *i* is the TrCH number and N_i is the number of bits input to the rate matching block. Note that the transport format combination number *j* for simplicity has been left out in the bit numbering, i.e. $N_i = N_{ij}$. The bits after separation are denoted by $x_{bi1}, x_{bi2}, x_{bi3}, \dots, x_{biX_i}$. For turbo encoded TrCHs with puncturing, *b* indicates the three sequences defined in section 4.2.7.3 <u>, with *b*=1 indicating the first sequence, and so forth</u>. The sequence denoted as *b*=1 contains all of the systematic bits and some systematic, first and second parity trellis termination bits; the sequence denoted as *b*=3 contains all of the second parity bits and some systematic, first and second parity trellis termination bits. For all other cases *b* is defined to be 1. X_i is the number of bits in each separated bit sequence. The relation between e_{ik} and x_{bik} is given below.

$$\begin{array}{ll} x_{1,i,k} = e_{i,3(k-1)+1+(\alpha_1+\beta_{n_i}) \bmod 3} & k = 1, 2, 3, ..., X_i & X_i = \\ \lfloor N_i / 3 \rfloor & \\ x_{1,i,\lfloor N_i / 3 \rfloor + k} = e_{i,3\lfloor N_i / 3 \rfloor + k} & k = 1, ..., N_i \bmod 3 & \text{Note: When } (N_i \bmod 3) = 0 \text{ this row is not needed.} & \\ x_{2,i,k} = e_{i,3(k-1)+1+(\alpha_2+\beta_{n_i}) \bmod 3} & k = 1, 2, 3, ..., X_i & X_i = \lfloor N_i / 3 \rfloor \\ x_{3,i,k} = e_{i,3(k-1)+1+(\alpha_3+\beta_{n_i}) \bmod 3} & k = 1, 2, 3, ..., X_i & X_i = \lfloor N_i / 3 \rfloor \end{array}$$

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

$$x_{1,i,k} = e_{i,k}$$
 $k = 1, 2, 3, ..., X_i$ $X_i = N_i$

4.2.7.3.2 Bit collection

The bits x_{bik} are input to the rate matching algorithm described in subclause 4.2.7.5. The bits output from the rate matching algorithm are denoted $y_{bi1}, y_{bi2}, y_{bi3}, \dots, y_{biY_i}$.

Bit collection is the inverse function of the separation. The bits after collection are denoted by $z_{bi1}, z_{bi2}, z_{bi3}, ..., z_{biY_i}$. After bit collection, the bits indicated as punctured are removed and the bits are then denoted by $f_{i1}, f_{i2}, f_{i3}, ..., f_{iV_i}$, where *i* is the TrCH number and $V_i = N_{ij} + \Delta N_{ij}$. The relations between y_{bik}, z_{bik} , and f_{ik} are given below. For turbo encoded TrCHs with puncturing $(Y_i = X_i)$:

i of theoret include include pu		
$z_{i,3(k-1)+1+(\alpha_1+\beta_{n_i}) \mod 3} = y_{1,i,k}$	$k = 1, 2, 3,, Y_i$	
$z_{i,3\lfloor N_i/3\rfloor+k} = y_{1,i,\lfloor N_i/3\rfloor+k}$	$k = 1, \ldots, N_i \mod 3$	Note: When $(N_i \mod$
(3) = 0 this row is not needed.		
$z_{i,3(k-1)+1+(\alpha_2+\beta_{n_i}) \mod 3} = y_{2,i,k}$	$k = 1, 2, 3,, Y_i$	
$z_{i,3(k-1)+1+(\alpha_3+\beta_{n_i}) \mod 3} = y_{3,i,k}$	$k = 1, 2, 3,, Y_i$	

After the bit collection, bits $z_{i,k}$ with value δ , where $\delta \not\in \{0, 1\}$, are removed from the bit sequence. Bit $f_{i,1}$ corresponds to the bit $z_{i,k}$ with smallest index k after puncturing, bit $f_{i,2}$ corresponds to the bit $z_{i,k}$ with second smallest index k after puncturing, and so on. For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

$$z_{i,k} = y_{1,i,k}$$
 $k = 1, 2, 3, ..., Y_i$

When repetition is used, $f_{i,k}=z_{i,k}$ and $Y_i=V_i$.

When puncturing is used, $Y_i=X_i$ and bits $z_{i,k}$ with value δ , where $\delta \notin \{0, 1\}$, are removed from the bit sequence. Bit $f_{i,1}$ corresponds to the bit $z_{i,k}$ with smallest index *k* after puncturing, bit $f_{i,2}$ corresponds to the bit $z_{i,k}$ with second smallest index *k* after puncturing, and so on.

4.2.7.4 Bit separation and collection in downlink

- The systematic bits of turbo encoded TrCHs shall not be punctured, <u>the other bits</u> however, systematic bits for trellis termination may be punctured.

The systematic bits, first parity bits and second parity bits in the bit sequence input to the rate matching block are therefore separated into three sequences <u>of equal lengths</u>. <u>The first sequence contains :</u>

- All of the systematic bits that are from turbo encoded TrCHs.
- Some of the systematic, first parity and second parity bits that are for trellis termination.

The second sequence contains:

- All of the first parity bits that are from turbo encoded TrCHs.
- <u>Some of the systematic, first parity and second parity bits that are for trellis</u> <u>termination.</u>

The third sequence contains:

- All of the second parity bits that are from turbo encoded TrCHs.
- Some of the systematic, first parity and second parity bits that are for trellis termination.

, one sequence containing all of the systematic bits and some systematic, first and second parity trellis termination bits; the second sequence containing all of the first parity bits and some systematic, first and second parity trellis termination bits and the third sequence containing all of the second parity bits and some systematic, first and second parity trellis termination bits. Puncturing is only applied only to the second and third sequences. The bit separation function is transparent for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition. The bit separation and bit collection are illustrated in figures 7 and 8.

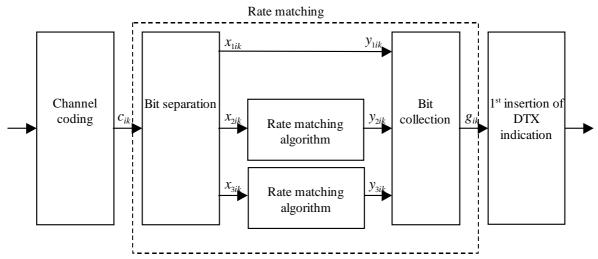


Figure 7: Puncturing of turbo encoded TrCHs in downlink

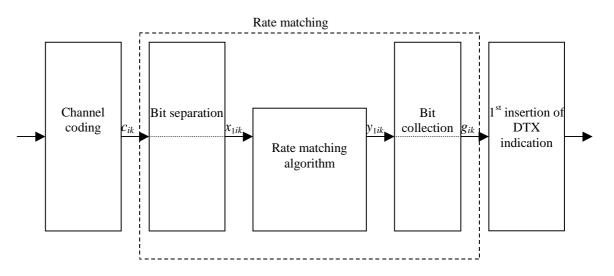


Figure 8: Rate matching for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition in downlink

4.2.7.4.1 Bit separation

The bits input to the rate matching are denoted by $c_{i1}, c_{i2}, c_{i3}, \dots, c_{iE_i}$, where *i* is the TrCH number and E_i is the number of bits input to the rate matching block. Note that E_i is a multiple of 3 for turbo encoded TrCHs and that the transport format *l* for simplicity has been left out in the bit numbering, i.e. $E_i = N_{il}^{TTI}$. The bits after separation are denoted by $x_{bi1}, x_{bi2}, x_{bi3}, \dots, x_{biX_i}$. For turbo encoded TrCHs with puncturing, *b* indicates the three sequences defined in section 4.2.7.4, with b=1 indicating the first sequence, and so forth.

For all other cases *b* is defined to be 1. X_i is the number of bits in each separated bit sequence. The relation between c_{ik} and x_{bik} is given below. For turbo encoded TrCHs with puncturing:

$$\begin{array}{ll} x_{1,i,k} = c_{i,3(k-1)+1} & k = 1, 2, 3, \dots, X_i & X_i = \\ E_i /3 & & \\ x_{2,i,k} = c_{i,3(k-1)+2} & k = 1, 2, 3, \dots, X_i & X_i = E_i /3 \\ x_{3,i,k} = c_{i,3(k-1)+3} & k = 1, 2, 3, \dots, X_i & X_i = E_i /3 \\ \end{array}$$

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

$$x_{1,i,k} = c_{i,k}$$
 $k = 1, 2, 3, ..., X_i$ $X_i = E_i$

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4.2.7 Rate matching

Rate matching means that bits on a transport channel are repeated or punctured. Higher layers assign a rate-matching attribute for each transport channel. This attribute is semistatic and can only be changed through higher layer signalling. The rate-matching attribute is used when the number of bits to be repeated or punctured is calculated. The number of bits on a transport channel can vary between different transmission time intervals. In the downlink the transmission is interrupted if the number of bits is lower than maximum. When the number of bits between different transmission time intervals in uplink is changed, bits are repeated or punctured to ensure that the total bit rate after TrCH multiplexing is identical to the total channel bit rate of the allocated dedicated physical channels.

If no bits are input to the rate matching for all TrCHs within a CCTrCH, the rate matching shall output no bits for all TrCHs within the CCTrCH and no uplink DPDCH will be selected in the case of uplink rate matching.

Notation used in subcaluse 4.2.7 and subclauses:

 $N_{i,j}$: For uplink: Number of bits in a radio frame before rate matching on TrCH *i* with transport format combination *j*.

For downlink: An intermediate calculation variable (not an integer but a multiple of 1/8).

- $N_{i,l}^{TTI}$: Number of bits in a transmission time interval before rate matching on TrCH *i* with transport format *l*. Used in downlink only.
- $\Delta N_{i,j}$: For uplink: If positive number of bits that should be repeated in each radio frame on TrCH *i* with transport format combination *j*.

If negative - number of bits that should be punctured in each radio frame on TrCH i with transport format combination j.

For downlink : An intermediate calculation variable (not an integer but a multiple of 1/8).

 $\Delta N_{i,l}^{TTI}$: If positive - number of bits to be repeated in each transmission time interval on TrCH *i* with transport format *l*.

If negative - number of bits to be punctured in each transmission time interval on TrCH i with transport format l.

Used in downlink only.

 $Np_{i,l}^{TTI,m}$, m=0 to $(F_{max}/F_i) - 1$: Positive or null: number of bits to be removed in TTI number m

within the largest TTI, to create the required gaps in the compressed radio frames of this TTI, in case of compressed mode by puncturing, for TrCH *i* with transport format *l*. In case of fixed positions and compressed mode by puncturing, this value is noted $Np_{i,\max}^{TTI,m}$ since it is calculated for all TrCH with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $Np_{i,l}^n$ n=0 to F_{max} -1:Positive or null: number of bits, in radio frame number *n* within the largest TTI, corresponding to the gap for compressed mode in this radio frame, for TrCH *i* with transport format *l*. The value will be null for the radio frames not overlapping with a transmission gap. In case of fixed positions and compressed mode by puncturing, this value is noted $Np_{i,max}^n$ since it is calculated for all TrCHs with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

- $N_{TGL}[k]$, k=0 to F_{max} -1 : Positive or null: number of bits in each radio frame corresponding to the gap for compressed mode for the CCTrCH.
- RM_i : Semi-static rate matching attribute for transport channel *i*. RM_i is provided by higher layers or takes a value as indicated in section 4.2.13.
- *PL:* Puncturing limit for uplink. This value limits the amount of puncturing that can be applied in order to avoid multicode or to enable the use of a higher spreading factor. Signalled from higher layers. The allowed puncturing in % is actually equal to (1-PL)*100.
- $N_{data,j}$: Total number of bits that are available for the CCTrCH in a radio frame with transport format combination *j*.
- *I:* Number of TrCHs in the CCTrCH.
- $Z_{i,j}$: Intermediate calculation variable.
- F_i : Number of radio frames in the transmission time interval of TrCH *i*.
- F_{max} Maximum number of radio frames in a transmission time interval used in the CCTrCH :

$$F_{\max} = \max_{1 \le i \le I} F_i$$

 n_i : Radio frame number in the transmission time interval of TrCH *i* ($0 \le n_i < F_i$).

- *q:* Average puncturing or repetition distance (normalised to only show the remaining rate matching on top of an integer number of repetitions). Used in uplink only.
- P1_{*F*}(n_i): The column permutation function of the 1st interleaver, P1_{*F*}(x) is the original position of column with number x after permutation. P1 is defined on table 3 of section 4.2.5.2 (note that the P1_{*F*} is self-inverse). Used for rate matching in uplink only.
- *S*[*n*]: The shift of the puncturing or repetition pattern for radio frame n_i when $n = P1_{F_i}(n_i)$. Used in uplink only.
- $TF_i(j)$: Transport format of TrCH *i* for the transport format combination *j*.
- TFS(i) The set of transport format indexes l for TrCH i.
- *TFCS* The set of transport format combination indexes *j*.
- e_{ini} Initial value of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.

e_{plus} Increment of variable *e* in the rate matching pattern determination algorithm of subclause4.2.7.5.

 e_{minus} Decrement of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.

b: Indicates systematic and parity bits

b=1: Systematic bit. x_k in subclause 4.2.3.2.1.

b=2: 1^{st} parity bit (from the upper Turbo constituent encoder). z_k in subcaluse 4.2.3.2.1.

b=3: 2^{nd} parity bit (from the lower Turbo constituent encoder). z'_k in subclause 4.2.3.2.1.

The * (star) notation is used to replace an index *x* when the indexed variable X_x does not depend on the index *x*. In the left wing of an assignment the meaning is that " $X_* = Y$ " is equivalent to "**for all** <u>x</u> **do** $X_x = Y$ ". In the right wing of an assignment, the meaning is that " $Y = X_*$ " is equivalent to "**take any** <u>x</u> **and do** $Y = X_x$ ".

The following relations, defined for all TFC j, are used when calculating the rate matching parameters:

 $Z_{0,j} = 0$

$$Z_{i,j} = \left[\frac{\left(\left(\sum_{m=1}^{i} RM_{m} \times N_{m,j} \right) \times N_{data,j} \right)}{\sum_{m=1}^{l} RM_{m} \times N_{m,j}} \right] \text{ for all } i = 1 \dots I$$

$$\Delta N_{i,j} = Z_{i,j} - Z_{i-1,j} - N_{i,j} \text{ for all } i = 1 \dots I$$

$$(1)$$

4.2.7.1 Determination of rate matching parameters in uplink

4.2.7.1.1 Determination of SF and number of PhCHs needed

In uplink, puncturing can be applied to match the CCTrCH bit rate to the PhCH bit rate. The bit rate of the PhCH(s) is limited by the UE capability and restrictions imposed by UTRAN, through limitations on the PhCH spreading factor. The maximum amount of puncturing that can be applied is <u>1-PL</u>, <u>PL</u> is signalled from higher layers and denoted by <u>PL</u>. The number of available bits in the radio frames of one PhCH for all possible spreading factors is given in [2]. Denote these values by N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , and N_4 , where the index refers to the spreading factor. The possible number of bits available to the CCTrCH on all PhCHs, N_{data} , then are { N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , N_4 , $2 \times N_4$, $3 \times N_4$, $4 \times N_4$, $5 \times N_4$, $6 \times N_4$ }.

For a RACH CCTrCH SET0 represents the set of N_{data} values allowed by the UTRAN, as set by the minimum SF provided by higher layers. SET0 may be a sub-set of { N_{256} , N_{128} , N_{64} , N_{32} }. SET0 does not take into account the UE's capability.

For other CCTrCHs, SET0 denotes the set of N_{data} values allowed by the UTRAN and supported by the UE, as part of the UE's capability. SET0 can be a subset of { N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , N_4 , $2 \times N_4$, $3 \times N_4$, $4 \times N_4$, $5 \times N_4$, $6 \times N_4$ }. $N_{data, j}$ for the transport format combination *j* is determined by executing the following algorithm:

SET1 = {
$$N_{data}$$
 in SET0 such that $\left(\min_{1 \le y \le I} \{RM_y\}\right) \times N_{data} - \sum_{x=1}^{I} RM_x \times N_{x,j}$ is non

negative }

If SET1 is not empty and the smallest element of SET1 requires just one PhCH then $N_{data,j} = \min \text{SET1}$

else

SET2 = {
$$N_{data}$$
 in SET0 such that $\left(\min_{1 \le y \le I} \{RM_y\}\right) \times N_{data} - PL \times \sum_{x=1}^{I} RM_x \times N_{x,j}$ is non

negative }

Sort SET2 in ascending order

 $N_{data} = \min \text{SET2}$

While N_{data} is not the max of SET2 and the follower of N_{data} requires no additional PhCH do

 N_{data} = follower of N_{data} in SET2

End while

 $N_{data,j} = N_{data}$

End if

For a RACH CCTrCH, if $N_{data,j}$ is not part of the UE's capability then the TFC *j* cannot be used.