TSG-RAN Meeting #8 Düsseldorf, Germany, 21-23 June 2000

Title: Agreed CRs to TS 25.223

Source: TSG-RAN WG1

Agenda item: 5.1.3

No.	Doc #	Spec	CR	Rev	Subject	Cat	Current_v	New_v
1	R1-000512	25.223	008	-	Editorial Modifications for 25.223	D	3.2.0	3.3.0
2	R1-000630	25.223	009	-	Editorial modification of 25.223	D	3.2.0	3.3.0
3	R1-000631	25.223	010	-	Editorial modification of 25.223	D	3.2.0	3.3.0
4	R1-000717	25.223	011	2	Editorial modification of 25.223	D	3.2.0	3.3.0
5	R1-000779	25.223	012	2	Modified code sets on SCH for cell search in	F	3.2.0	3.3.0
6	R1-000748	25.223	013	1	Editorial update of TS25.223	D	3.2.0	3.3.0

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Subject:	Editorial M	odifications for 25.	.223		
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1 Scope

The present document describes spreading and modulation for UTRA Physical Layer TDD mode.

The present document describes multiplexing, channel coding and interleaving for UTRA Physical Layer TDD mode.

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7 Synchronisation codes

7.1 Code Generation

The Primary code sequence, C_p is constructed as a so-called generalised hierarchical Golay sequence. The Primary SCH is furthermore chosen to have good aperiodic auto correlation properties.

Define $a = \langle x_1, x_2, x_3, ..., x_{16} \rangle = \langle 1, 1, 1, 1, 1, 1, 1, -1, -1, 1, -1, 1, -1, 1, -1, 1 \rangle$

The PSC code word is generated by repeating the sequence 'a' modulated by a Golay complementary sequence and creating a complex-valued sequence with identical real and imaginary components.

The PSC code word C_p is defined as $C_p = \langle y(0), y(1), y(2), ..., y(255) \rangle$

where $y = (1 + j) \times \langle a, a, a, -a, -a, a, a, -a, -a, a, a, a, -a, a, a, a \rangle$

and the left most index corresponds to the chip transmitted first in each time slot.

The <u>16-12</u> secondary synchronization code words, $\{C_0, \dots, C_{15}, C_{11}\}\$ are complex valued with identical real and imaginary components, and are constructed from the position wise multiplication of a Hadamard sequence and a sequence z, defined as

The Hadamard sequences are obtained as the rows in a matrix H_8 constructed recursively by:

$$\begin{aligned} H_0 &= (1) \\ H_k &= \begin{pmatrix} H_{k-1} & H_{k-1} \\ H_{k-1} & -H_{k-1} \end{pmatrix}, \quad k \geq 1 \end{aligned}$$

The rows are numbered from the top starting with row θ (the all zeros sequence).

Denote the *n*:th Hadamard sequence h_n as a row of H_{δ} numbered from the top, n = 0, 1, 2, ..., 255, in the sequel.

Furthermore, let $h_m(i)$ and z(i) denote the *i*:th symbol of the sequence h_m and z, respectively where i = 0, 1, 2, ..., 255 and i = 0 corresponds to the leftmost symbol.

The i:th SCH code word, $C_{SCH,i}$, $i = 0, ..., \frac{15 \cdot 11}{11}$ is then defined as

 $C_{\text{SCH,i}} = (1+j) \times \langle h_m(0) \times z(0), h_m(1) \times z(1), h_m(2) \times z(2), \dots, h_m(255) \times z(255) \rangle,$

where $m = (16 \times i)$ and the leftmost chip in the sequence corresponds to the chip transmitted first in time.

This code word is chosen from every 16th row of the matrix H_{sr} , which yields 16 possible codewords.

The Secondary SCH code words are defined in terms of $C_{SCH,i}$ and the definition of $\{C_0, \dots, C_{45}, C_{11}\}$ now follows as:

 $C_i = C_{SCH, i}, i=0,...,1$ 15

7.2 Code Allocation

Three SCH codes are QPSK modulated and transmitted in parallel with the primary synchronization code. The QPSK modulation carries the following information.

- The code group that the base station belongs to (5 bits; Cases 1, 2)
- The position of the frame within an interleaving period of 20 msec (1 bit, Cases 1, 2)

- The position of the slot within the frame (1 bit, Case 2)

The modulated codes are also constructed such that their cyclic-shifts are unique, i.e. a non-zero cyclic shift less than 2 (Case 1) and 4 (Case 2) of any of the sequences is not equivalent to some cyclic shift of any other of the sequences. Also, a non-zero cyclic shift less than 2 (Case 1) and 4 (Case 2) of any of the sequences is not equivalent to itself with any other cyclic shift less than 8. The secondary synchronization codes are partitioned into two code sets for Case 1 and four code sets for Case 2. The set is used to provide the following information:

Case 1:

Table 2: Code Set Allocation for Case 1

Code Set	Code Group
1	0-15
2	16-31

The code group and frame position information is provided by modulating the secondary codes in the code set.

Case 2:

Table 3: Code Set Allocation for Case 2

Code Set	Code Group
1	0-7
2	8-15
3	16-23
4	24-31

The slot timing and frame position information is provided by the comma free property of the code word and the Code group is provided by modulating some of the secondary codes in the code set.

The following SCH codes are allocated for each code set:

Case 1

Code set 1: C_0, C_1, C_2

Code set 2: C₃, C₄, C_{5.}

Case 2

Code set 1: C_0 , C_1 , C_2 .

Code set 2: C₃, C₄, C_{5.}

Code set 3: C₆, C₇, C_{8.}

Code set 4: C₉, C₁₀, C_{11.}

The following subsections 7.2.1 to 7.2.2 refer to the two cases of SCH/P-CCPCH usage as described in [7].

Note that in the Tables 4-6-5 corresponding to Cases 1,2, and 3, respectively, Frame 1 implies the frame with an odd SFN and Frame 2 implies the frame with an even SFN.

7.2.1 Code allocation for Case 1:

1

1

1

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NOTE: Modulation by "j" indicates that the code is transmitted on the Q channel.

Code Group	Code Set		Frame 1			Associated t _{offset}		
0	1	C ₀	C ₁	C ₂	C ₀	C ₁	-C2	t ₀
1	1	C ₀	-C1	C ₂	C ₀	-C1	-C2	t ₁
2	1	-C0	C ₁	C ₂	-C ₀	C ₁	-C2	t ₂
3	1	-C0	-C1	C ₂	-C ₀	-C1	-C2	t ₃
4	1	jC ₀	JC₁ <u>jC</u> 1	C2	jC ₀	jC₁	-C2	t4
5	1	jC ₀	-jC1	C ₂	jC ₀	-jC₁	-C2	t5
6	1	-jC ₀	JC₁ <u>jC</u> 1	C ₂	-jC ₀	jC₁	-C ₂	t ₆
7	1	-jCo	-jC₁	C2	-jC ₀	-jC₁	-C2	t7
8	1	jC ₀	JC<u>2</u>jC2	C ₁	jC ₀	jC ₂	-C1	t ₈
9	1	jC ₀	-jC ₂	C ₁	jC ₀	-jC ₂	-C1	t ₉
10	1	-jC ₀	JC<u>2</u>jC 2	C ₁	-jC ₀	jC ₂	-C1	t ₁₀
11	1	-jC ₀	-jC ₂	C ₁	-jC ₀	-jC2	-C1	t ₁₁
12	1	jC1	JC₂ <u>jC</u> ₂	C ₀	JC₁ <u>jC</u> 1	jC ₂	-C0	t ₁₂
13	1	jC₁	-jC ₂	C ₀	JC₁ <u>jC</u> 1	-jC ₂	-C0	t ₁₃
14	1	-jC₁	JC ₂ jC ₂	C ₀	-jC₁	jC ₂	-C0	t ₁₄
15	1	-jC₁	-jC ₂	C ₀	-jC₁	-jC ₂	-C0	t ₁₅
16	2	C ₃	C4	C_5	C ₃	C ₄	-C5	t ₁₆
17	2	C ₃	-C4	C_5	C ₃	-C4	-C5	t ₁₇
20	2	jC₃	<mark>JC</mark> ₄ <u>jC</u> ₄	C_5	jC₃	jC4	-C5	t ₂₀
24	2	jC₃	<mark>JC</mark> ₅ <u>jC</u> ₅	C_4	jC₃	JC₅jC₅	-C4	t ₂₄
31	2	-jC4	-jC ₅	C ₃	-jC4	-jC ₅	-C3	t ₃₁

Table 4: Code Allocation for Case 1

NOTE: The code construction for code groups 0 to 15 using only the SCH codes from code set 1 is shown. The construction for code groups 16 to 31 using the SCH codes from code set 2 is done in the same way.

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2 References

The following documents contain provisions which, through reference in this text, constitute provisions of the present document.

- References are either specific (identified by date of publication, edition number, version number, etc.) or non-specific.
- For a specific reference, subsequent revisions do not apply.
- For a non-specific reference, the latest version applies.

[1]	3G TS 25.201: "Physical layer - general description"
[2]	3G TS 25.211: "Physical channels and mapping of transport channels onto physical channels (FDD)"
[3]	3G TS 25.212: "Multiplexing and channel coding (FDD)"
[4]	3G TS 25.213: "Spreading and modulation (FDD)"
[5]	3G TS 25.214: "Physical layer procedures (FDD)"
[6]	3G TS 25.215: "Physical layer – Measurements (FDD)"
[7]	3G TS 25 221: "Physical channels and manning of transport channels onto physical

- [7] 3G TS 25.221: "Physical channels and mapping of transport channels onto physical channels (TDD)"
- [8] 3G TS 25.222: "Multiplexing and channel coding (TDD)"
- [9] 3G TS 25.223: "Spreading and modulation (TDD)"
- [10] 3G TS 25.224: "Physical layer procedures (TDD)"
- [11] 3G TS 25.225: "Physical layer Measurements (TDD)"

3 Abbreviations

For the purposes of the present document, the following abbreviations apply:

CDMA Code Division Multiple Access P-CCPCH Primary Common Control Physical Channel PN Pseudo Noise **OVSF** Orthogonal Variable Spreading Factor PRACH Physical Random Access Channel Quadrature Phase Shift Keying QPSK RACH Random Access Channel SCH Synchronisation Channel

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6.2 Channelisation codes

The elements $c_q^{(k)}$; k=1,...,K; q=1,...,Q_k; of the real valued channelisation codes $\mathbf{c}^{(k)} = (c_1^{(k)}, c_2^{(k)}, ..., c_{Q_k}^{(k)})$; k=1,...,K;

shall be taken from the set

$$\mathbf{V}_{c} = \left\{1, -1\right\} \tag{3}$$

The $\mathbf{c}_{Q_k}^{(k)}$ are Orthogonal Variable Spreading Factor (OVSF) codes, allowing to mix in the same timeslot channels with different spreading factors while preserving the orthogonality. The OVSF codes can be defined using the code tree of figure 1.

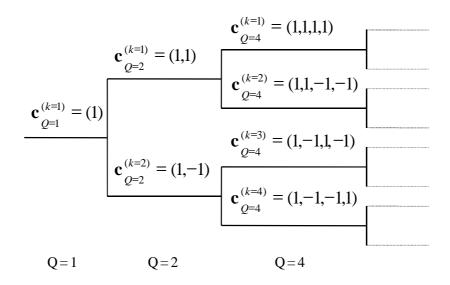


Figure 1:Code-tree for generation of Orthogonal Variable Spreading Factor (OVSF) codes for_Channelisation Operation

Each level in the code tree defines a spreading factor indicated by the value of Q in the figure. All codes within the code tree cannot be used simultaneously in a given timeslot. A code can be used in a timeslot if and only if no other code on the path from the specific code to the root of the tree or in the sub-tree below the specific code is used in this timeslot. This means that the number of available codes in a slot is not fixed but depends on the rate and spreading factor of each physical channel. The spreading factor goes up to $Q_{MAX}=16$.

6.3 Scrambling codes

The spreading of data by a real valued channelisation code $\mathbf{c}^{(k)}$ of length Q_k is followed by a cell specific complex scrambling sequence $\underline{\mathbf{i}} = (\underline{i}_1, \underline{i}_2, ..., \underline{i}_{16})$. The elements $\underline{i}_i; i = 1, ..., 16$ of the complex valued scrambling codes shall be taken from the complex set

$$\underline{V}_{v} = \{1, j, -1, -j\}$$
(54)

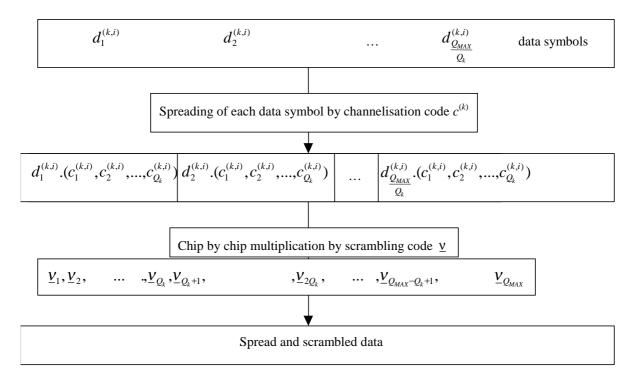
In equation 5-4 the letter j denotes the imaginary unit. A complex scrambling code \underline{i} is

generated from the binary scrambling codes $\mathbf{v} = (v_1, v_2, ..., v_{16})$ of length 16 shown in Annex A. The relation between the elements \mathbf{i} and \mathbf{i} is given by:

$$\underline{\mathbf{v}}_{i} = (\mathbf{j})^{i} \cdot \mathbf{v}_{i} \qquad \mathbf{v}_{i} \in \{1, -1\}, \ \mathbf{i} = 1, \dots, 16$$
(65)

Hence, the elements \underline{v}_i of the complex scrambling code \underline{i} are alternating real and imaginary.

The length matching is obtained by concatenating Q_{MAX}/Q_k spread words before the scrambling. The scheme is illustrated in figure <u>32</u> below and is described in more detail in section 6.4.





6.4 Spread signal of data symbols and data blocks

The combination of the user specific channelisation and cell specific scrambling codes can be seen as a user and cell specific spreading code $\mathbf{s}^{(k)} = (\mathbf{s}_n^{(k)})$ with

$$s_{p}^{(k)} = c_{1+[(p-1) \mod Q_{k}]}^{(k)} \cdot \dot{L}_{1+[(p-1) \mod Q_{MAX}]}, k=1,...,K, p=1,...,N_{k}Q_{k}.$$

With the root raised cosine chip impulse filter $Cr_0(t)$ the transmitted signal belonging to the data block $\underline{\mathbf{d}}^{(k,1)}$ of equation transmitted before the midamble is

$$\underline{d}^{(k,1)}(t) = \sum_{n=1}^{N_k} \underline{d}_n^{(k,1)} \sum_{q=1}^{Q_k} s_{(n-1)Q_k+q}^{(k)} \cdot Cr_o(t - (q-1)T_c - (n-1)Q_kT_c)$$
(36)

and for the data block $\underline{\mathbf{d}}^{(k,2)}$ of equation transmitted after the midamble

$$\underline{d}^{(k,2)}(t) = \sum_{n=1}^{N_k} \underline{d}_n^{(k,2)} \sum_{q=1}^{Q_k} s_{(n-1)Q_k+q}^{(k)} \cdot Cr_0(t - (q-1)T_C - (n-1)Q_kT_c - N_kQ_kT_c - L_mT_c).$$
(47)

where L_m is the number of midamble chips.

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Annex A (Normative): Scrambling Codes

The applicable scrambling codes are listed in below. Code numbers are referring to table 6 'Mapping scheme for Cell Parameters, Code Groups, Scrambling Codes, Midambles and t_{offset} ' in section 7.2-3 'Code AllocationEvaluation of synchronisation codes'.

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7 Synchronisation codes

7.1 Code Generation

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The Primary code sequence, C_p is constructed as a so-called generalised hierarchical Golay sequence. The Primary SCH is furthermore chosen to have good aperiodic auto correlation properties.

Define $a = \langle x_1, x_2, x_3, ..., x_{16} \rangle = \langle 1, 1, 1, 1, 1, 1, 1, -1, -1, 1, -1, 1, -1, 1, -1, 1 \rangle$

The PSC code word is generated by repeating the sequence 'a' modulated by a Golay complementary sequence and creating a complex-valued sequence with identical real and imaginary components.

The PSC code word C_p is defined as $C_p = \langle y(0), y(1), y(2), \dots, y(255) \rangle$

where $y = (1 + j) \times \langle a, a, a, -a, -a, a, a, -a, -a, a, a, -a, a, -a, a, a \rangle$

and the left most index corresponds to the chip transmitted first in each time slot.

The 1<u>26</u> secondary synchronization code words, { $C_0, C_1, C_3, C_4, C_5, C_6, C_8, C_{10}, C_{12}, C_{13}, C_{14}, ..., C_{15}$ } are complex valued with identical real and imaginary components, and are constructed from the position wise multiplication of a Hadamard sequence and a sequence z, defined as

The Hadamard sequences are obtained as the rows in a matrix H_8 constructed recursively by:

$$\begin{aligned} H_0 &= (1) \\ H_k &= \begin{pmatrix} H_{k-1} & H_{k-1} \\ H_{k-1} & -H_{k-1} \end{pmatrix}, \quad k \geq 1 \end{aligned}$$

The rows are numbered from the top starting with row 0 (the all zeros sequence).

Denote the *n*:th Hadamard sequence as a row of H_8 numbered from the top, n = 0, 1, 2, ..., 255, in the sequel.

Furthermore, let $h_m(i)$ and z(i) denote the *i*:th symbol of the sequence h_m and z, respectively where i = 0, 1, 2, ..., 255 and i = 0 corresponds to the leftmost symbol.

The i:th SCH code word, $C_{SCH,i}$, i = 0, <u>1</u>, <u>3</u>, <u>4</u>, <u>5</u>, <u>6</u>, <u>8</u>, <u>10</u>, <u>12</u>, <u>13</u>, <u>14</u>,..., 15 is then defined as

 $C_{\text{SCH,i}} = (1+j) \times \langle h_m(0) \times z(0), h_m(1) \times z(1), h_m(2) \times z(2), \dots, h_m(255) \times z(255) \rangle,$

where $m = (16 \times i)$ and the leftmost chip in the sequence corresponds to the chip transmitted first in time.

This code word is chosen from every 16^{th} row of the matrix $H_{8.}$, which yields 16 possible codewords.

The Secondary SCH code words are defined in terms of $C_{SCH,i}$ and the definition of $\{C_0, \underline{C_1}, \underline{C_3}, \underline{C_4}, \underline{C_5}, \underline{C_6}, \underline{C_8}, \underline{C_{10}}, \underline{C_{12}}, \underline{C_{13}}, \underline{C_{14}}, \dots, \underline{C_{15}}\}$ now follows as:

 $C_i = C_{SCH, i}, i=0, 1, 3, 4, 5, 6, 8, 10, 12, 13, 14$...,15

7.2 Code Allocation

Three SCH codes are QPSK modulated and transmitted in parallel with the primary synchronization code. The QPSK modulation carries the following information:

- the code group that the base station belongs to (5 bits; Cases 1, 2);

- the position of the frame within an interleaving period of 20 msec (1 bit, Cases 1, 2);
- the position of the slot within the frame (1 bit, Case 2).

The modulated codes are also constructed such that their cyclic-shifts are unique, i.e. a non-zero cyclic shift less than 2 (Case 1) and 4 (Case 2) of any of the sequences is not equivalent to some cyclic shift of any other of the sequences. Also, a non-zero cyclic shift less than 2 (Case 1) and 4 (Case 2) of any of the sequences is not equivalent to itself with any other cyclic shift less than 8. The secondary synchronization codes are partitioned into two code sets for Case 1 and four code sets for Case 2. The set is used to provide the following information:

Case 1:

Table 2: Code Set Allocation for Case 1

Code Set	Code Group
1	0-15
2	16-31

The code group and frame position information is provided by modulating the secondary codes in the code set.

Case 2:

Table 3: Code Set Allocation for Case 2

Code Set	Code Group
1	0-7
2	8-15
3	16-23
4	24-31

The slot timing and frame position information is provided by the comma free property of the code word and the Code group is provided by modulating some of the secondary codes in the code set.

The following SCH codes are allocated for each code set:

Case 1

Code set 1: $C_{\underline{\theta 1}}$, $C_{\underline{13}}$, $C_{\underline{25}}$.

Code set 2: C_{310} , C_{413} , C_{514} .

Case 2

Code set 1: $C_{\theta \underline{1}}$, $C_{4\underline{3}}$, $C_{2\underline{5}}$. Code set 2: $C_{3\underline{10}}$, $C_{4\underline{13}}$, $C_{5\underline{14}}$. Code set 3: $C_{6\underline{0}}$, $C_{7\underline{6}}$, $C_{8\underline{12}}$. Code set 4: $C_{9\underline{4}}$, $C_{4\underline{08}}$, $C_{4\underline{15}}$.

The following subclauses 7.2.1 to 7.2.2 refer to the two cases of SCH/P-CCPCH usage as described in [7].

Note that in the Tables 4-6 corresponding to Cases 1,2, and 3, respectively, Frame 1 implies the frame with an odd SFN and Frame 2 implies the frame with an even SFN.

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7.2.1 Code allocation for Case 1

NOTE: Modulation by "j" indicates that the code is transmitted on the Q channel.

Code Group	Code Set		Frame 1			Frame 2		Associated t _{offset}
0	1	C ₀₁	C ₄₃	C ₂₅	C ₀₁	C ₄₃	-C ₂₅	t ₀
1	1	C ₀₁	-C ₄₃	C ₂₅	C ₀₁	-C ₄₃	-C ₂₅	t ₁
2	1	-C ₀₁	C ₄₃	C ₂₅	-C ₀₁	C ₁₃	-C <u>25</u>	t ₂
3	1	-C ₀₁	-C ₄₃	C ₂₅	-C ₀₁	-C ₄₃	-C <u>25</u>	t ₃
4	1	jC ₀₁	JC ₁₃	C ₂₅	jC <mark>e1</mark>	jC <u>₄</u> 3	-C <u>25</u>	t4
5	1	jC _{θ1}	-jC ₁₃	C ₂₅	jC <mark>₀1</mark>	-jC₄ <u>3</u>	-C ₂₅	t ₅
6	1	-jC ₀₁	JC ₁₃	C ₂₅	-jC ₀₁	jC <u>₄</u> 3	-C <u>25</u>	t ₆
7	1	-jC ₀₁	-jC ₁₃	C ₂₅	-jC ₉₁	-jC <u>₄</u> 3	-C <u>25</u>	t7
8	1	jC _{θ1}	JC <u>25</u>	C ₁₃	jC <mark>₀1</mark>	јС <u>25</u>	-C ₁₃	t ₈
9	1	jC ₀₁	-jC <u>25</u>	C ₄₃	jC <mark>e1</mark>	-jC <u>25</u>	-C ₄₃	t ₉
10	1	-jC ₀₁	JC <u>25</u>	C ₄₃	-jC <mark>e1</mark>	јС <u>25</u>	-C ₄₃	t ₁₀
11	1	-jC ₀₁	-jC <u>25</u>	C ₁₃	-jC ₀₁	-jC ₂₅	-C ₄₃	t ₁₁
12	1	jC <u>+3</u>	JC <u>25</u>	C ₀₁	JC ₄₃	јС <u>₂₅</u>	-C ₀₁	t ₁₂
13	1	jC <u>₄3</u>	-jC <u>25</u>	C ₀₁	JC₄ <u>3</u>	-jC <u>25</u>	-C ₀₁	t ₁₃
14	1	-jC ₁₃	JC <u>25</u>	C ₀₁	-jC ₁₃	јС <u>25</u>	-C ₀₁	t ₁₄
15	1	-jC ₄₃	-jC <u>25</u>	C ₀₁	-jC ₄₃	-jC <u>25</u>	-C ₀₁	t ₁₅
16	2	C ₃₁₀	C4 <u>13</u>	C₅ <u>14</u>	C ₃₁₀	C ₄₁₃	-C <u>514</u>	t ₁₆
17	2	C ₃₁₀	-C4 <u>13</u>	C <u>514</u>	C ₃₁₀	-C4 <u>13</u>	-C ₅₁₄	t ₁₇
20	2	jC <u>310</u>	JC ₄₁₃	C <u>514</u>	јС _{з<u>10</u>}	jC4 <u>13</u>	-C <u>514</u>	t ₂₀
24	2	jC ₃₁₀	jC <u>₅14</u>	C ₄₁₃	jC <u>310</u>	JC <u>₅14</u>	-C4 <u>13</u>	t ₂₄
31	2	-jC ₄₁₃	-jC ₅₁₄	C ₃₁₀	-jC ₄₁₃	-jC ₅₁₄	-C ₃₁₀	t ₃₁

Table 4: Code Allocation for Case 1

NOTE: The code construction for code groups 0 to 15 using only the SCH codes from code set 1 is shown. The construction for code groups 16 to 31 using the SCH codes from code set 2 is done in the same way.

7.2.2 Code allocation for Case 2

Table 5: Code Allocation for Case 2

Code	Code			Fra	me 1					Frar	ne 2			Associated toffset
Group	Set		Slot k			Slot k+8			Slot k			Slot k+8		
0	1	C ₀₁	C ₁₃	C ₂₅	C ₀₁	C ₄₃	-C ₂₅	-C ₀₁	-C ₄₃	C ₂₅	-C ₀₁	-C ₄₃	-C ₂₅	to
1	1	C ₀₁	-C ₄₃	C ₂₅	C ₀₁	-C ₄₃	-C ₂₅	-C ₀₁	C ₄₃	C ₂₅	-C ₀₁	C ₄₃	-C ₂₅	t ₁
2	1	jC ₀₁	jC ₄₃	C ₂₅	јС <u>₀₁</u>	jC ₄₃	-C ₂₅	-jC ₀₁	-jC ₄₃	C ₂₅	-jC ₀₁	-jC ₄₃	-C ₂₅	t ₂
3	1	jC ₀₁	-jC ₁₃	C ₂₅	јС <u>ө1</u>	-jC ₁₃	-C <u>25</u>	-jC ₀₁	jC ₁₃	C ₂₅	-jC <u>01</u>	jC ₁₃	-C <u>25</u>	t ₃
4	1	jC ₀₁	jC <u>25</u>	C ₁₃	јС <u>е1</u>	jC <u>25</u>	-C ₁₃	-jC ₀₁	-jC <u>25</u>	C ₁₃	-jC <u>01</u>	-jC <u>25</u>	-C ₁₃	t4
5	1	jC ₀₁	-jC ₂₅	C ₄₃	jC <mark>₀1</mark>	-jC <u>25</u>	-C ₄₃	-jC ₀₁	jC <u>25</u>	C ₁₃	-jC ₀₁	jC <u>25</u>	-C ₄₃	t ₅
6	1	jC <u>43</u>	jC <u>25</u>	C ₀₁	jC <u>₄</u> 3	jC <u>25</u>	-C ₀₁	-jC <u>₄3</u>	-jC <u>25</u>	C ₀₁	-jC <u>+3</u>	-jC <u>25</u>	-C ₀₁	t ₆
7	1	jC ₁₃	-jC ₂₅	C ₀₁	jC <u>₄3</u>	-jC <u>25</u>	-C ₀₁	-jC ₁₃	jC <u>25</u>	С ₀₁	-jC ₁₃	jC <u>25</u>	-C ₀₁	t ₇
8	2	C <u>310</u>	C4 <u>13</u>	C <u>₅14</u>	C₃ <u>10</u>	C4 <u>13</u>	- C ₅₁₄	-C <u>310</u>	-C4 <u>13</u>	C <u>₅14</u>	-C <u>310</u>	-C4 <u>13</u>	- С ₅₁₄	t ₈
9	2	С <u>з10</u>	-C4 <u>13</u>	C <u>₅14</u>	С <u>з10</u>	-C4 <u>13</u>	- C ₅₁₄	-C <u>₃₁₀</u>	C4 <u>13</u>	C <u>514</u>	-C <u>₃10</u>	C _{4<u>13</u>}	- C ₅₁₄	t9
10	2	jC <u>₃10</u>	jC _{4<u>13</u>}	C ₅₁₄	jC _{3<u>10</u>}	jC _{4<u>13</u>}	- C ₅₁₄	-jC _{3<u>10</u>}	-jC _{4<u>13</u>}	C _{5<u>14</u>}	-jC _{3<u>10</u>}	-jC _{4<u>13</u>}	- C ₅₁₄	t ₁₀
11	2	jC <u>₃10</u>	- јС₄ <u>13</u>	C <u>₅14</u>	jC <u>₃10</u>	-jC4 <u>13</u>	- C ₅₁₄	-jC₃ <u>10</u>	jC4 <u>13</u>	C₅ <u>14</u>	-jC <u>₃10</u>	jC4 <u>13</u>	- C <u>514</u>	t ₁₁
12	2	jC <u>₃10</u>	jC <u>₅14</u>	C4 <u>13</u>	jC <u>₃10</u>	jC <u>₅14</u>	- C ₄₁₃	-jC _{3<u>10</u>}	-jC <u>₅14</u>	C4 <u>13</u>	-jC _{3<u>10</u>}	-jC <u>₅14</u>	- C ₄₁₃	t ₁₂
13	2	јС _{3<u>10</u>}	- jC ₅₁₄	C _{4<u>13</u>}	јС _{3<u>10</u>}	-jC <u>₅14</u>	- C ₄₁₃	-jC _{3<u>10</u>}	jC ₅₁₄	C _{4<u>13</u>}	-jC _{3<u>10</u>}	jC <u>₅14</u>	- C ₄₁₃	t ₁₃
14	2	jC4 <u>13</u>	jC <u>514</u>	C ₃₁₀	jC4 <u>13</u>	jC <u>514</u>	- C ₃₁₀	-jC4 <u>13</u>	-jC <u>₅14</u>	C _{3<u>10</u>}	-jC _{4<u>13</u>}	-jC <u>₅14</u>	- C ₃₁₀	t ₁₄
15	2	jC4 <u>13</u>	- jC ₅₁₄	C ₃₁₀	jC4 <u>13</u>	-jC <u>₅14</u>	- C ₃₁₀	-jC4 <u>13</u>	jC <u>₅14</u>	C ₃₁₀	-jC _{4<u>13</u>}	jC <u>₅14</u>	- C <u>310</u>	t ₁₅
16	3	C <u>€0</u>	C7 <u>6</u>	C _{8<u>12</u>}	C <u>60</u>	C7 <u>6</u>	- C ₈₁₂	-C ₆₀	-C7 <u>6</u>	C _{8<u>12</u>}	-C ₆₀	-C7 <u>6</u>	- C ₈₁₂	t ₁₆
23	3	jC <u>7</u> 6	- јС ₈₁₂	C <u>60</u>	jC7 <u>6</u>	-jC ₈₁₂	-C ₆₀	-jC <u>≠</u> 6	jC _{8<u>12</u>}	C ₆₀	-jC7 <u>6</u>	јС _{8<u>12</u>}	-C ₆₀	t ₂₀
24	4	C ₉₄	C ₁₀₈	C ₁₄₁ 5	C ₉₄	C ₁₀₈	- C ₁₁₁ 5	-C ₉₄	-C ₁₀₈	C ₄₄₁ <u>5</u>	-C ₉₄	-C ₁₀₈	- C ₁₁₁ 5	t ₂₄
31	4	jC <u>₁₀</u> 8	- jC ₁₁₁ 5	C <u>94</u>	jC ₁₀₈	- jC _{11<u>15</u>}	-C <u>94</u>	-jC <u>₁₀</u> 8	jC _{11<u>15</u>}	C ₉₄	-jC ₁₀₈	jC <u>1115</u>	-C <u>94</u>	t ₃₁

NOTE: The code construction for code groups 0 to 15 using the SCH codes from code sets 1 and 2 is shown. The construction for code groups 16 to 31 using the SCH codes from code sets 3 and 4 is done in the same way.

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2 References

The following documents contain provisions which, through reference in this text, constitute provisions of the present document.

- References are either specific (identified by date of publication, edition number, version number, etc.) or non-specific.
- For a specific reference, subsequent revisions do not apply.
- For a non-specific reference, the latest version applies.
- [1] 3G TS 25.201: "Physical layer general description".
- [2] 3G TS 25.211: "Physical channels and mapping of transport channels onto physical channels (FDD)".
- [3] 3G TS 25.212: "Multiplexing and channel coding (FDD)".
- [4] 3G TS 25.213: "Spreading and modulation (FDD)".
- [5] 3G TS 25.214: "Physical layer procedures (FDD)".
- [6] 3G TS 25.215: "Physical layer Measurements (FDD)".
- [7] 3G TS 25.221: "Physical channels and mapping of transport channels onto physical channels (TDD)".
- [8] 3G TS 25.222: "Multiplexing and channel coding (TDD)".
- [9] 3G TS 25.223: "Spreading and modulation (TDD)".
- [10] 3G TS 25.224: "Physical layer procedures (TDD)".
- [11] 3G TS 25.225: "Physical layer Measurements (TDD)".
- [12] 3G TS 25.102: "UTRA (UE) TDD; Radio Transmission and Reception"
- [13] 3G TS 25.105: "UTRA (BS) TDD; Radio Transmission and Reception"

Symbols and abbreviations

3.1 Symbols

<u>3</u>

For the purposes of the present document, the following symbols apply:

 $\begin{array}{ccc} \underline{C_p:} & \underline{PSC} \\ \underline{C_i:} & i: th secondary SCH code \end{array}$

3.2 Abbreviations

For the purposes of the present document, the following abbreviations apply:

el

6.5 Modulation

The complex-valued chip sequence is QPSK modulated as shown in Figure 3 below.

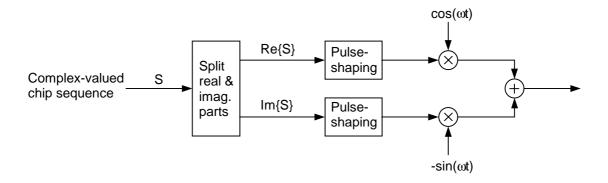


Figure 3: Modulation of complex valued chip sequences

The pulse-shaping characteristics are described in [12] and [13].

6.5.1 Combination of physical channels in uplink

Figure 4 illustrates how the maximum of two different physical uplink channels are combined within one timeslot. Each complex-valued spread channel is separately weighted by a weight factor G_i and combined using complex addition.

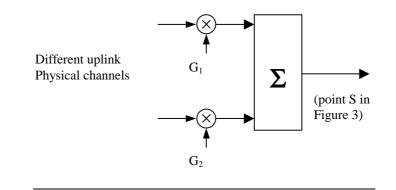
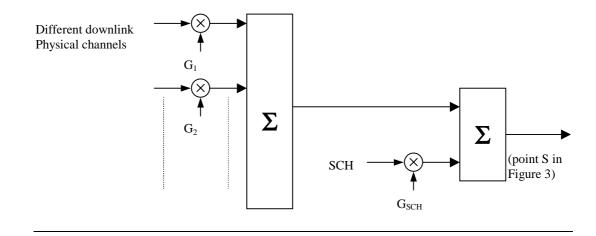


Figure 4: Combination of different physical channels in uplink

6.5.2 Combination of physical channels in downlink

Figure 5 illustrates how different physical downlink channels are combined within one timeslot. Each complex-valued spread channel is separately weighted by a weight factor G_i . If a timeslot contains the SCH, the complex-valued SCH, as described in [7] is separately weighted by a weight factor G_{SCH} . All downlink physical channels are then combined using complex addition.



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Figure 5: Combination of different physical channels in downlink in case of SCH timeslot

7 Synchronisation codes

7.1 Code Generation

The <u>pP</u>rimary <u>synchronisation</u> code <u>sequence(PSC)</u>, C_p is constructed as a so-called generalised hierarchical Golay sequence. The <u>PSCPrimary SCH</u> is furthermore chosen to have good aperiodic auto correlation properties.

The PSC code word is generated by repeating the sequence 'a' modulated by a Golay complementary sequence and creating a complex-valued sequence with identical real and imaginary components.

The PSC code word- C_p is defined as $C_p = \langle y(0), y(1), y(2), ..., y(255) \rangle$

where $y = (1 + j) \times \langle a, a, a, -a, -a, a, a, -a, -a, a, a, a, -a, a, a, a \rangle$

and the left most index corresponds to the chip transmitted first in each-time-slot.

The 16 secondary synchronization codes words, $\{C_0, ..., C_{15}\}$ are complex valued with identical real and imaginary components, and are constructed from the position wise multiplication of a Hadamard sequence and a sequence z, defined as:

) =

and $x_1, x_2, x_3, \dots, x_{16}$ are the same as in the definition of the sequence 'a' above.

The Hadamard sequences are obtained as the rows in a matrix H_8 constructed recursively by:

$$\begin{aligned} H_0 &= (1) \\ H_k &= \begin{pmatrix} H_{k-1} & H_{k-1} \\ H_{k-1} & -H_{k-1} \end{pmatrix} \quad k \geq 1 \end{aligned}$$

The rows are numbered from the top starting with row 0 (the all zeros <u>ones</u> sequence).

Denote the *n*:th Hadamard sequence as a row of H_8 numbered from the top, n = 0, 1, 2, ..., 255, in the sequel.

Furthermore, let $h_m(i)$ and z(i) denote the *i*:th symbol of the sequence h_m and *z*, respectively where i = 0, 1, 2, ..., 255 and i = 0 corresponds to the leftmost symbol.

The i:th <u>secondary</u> SCH code word, $C_{SCH,i}$, i = 0, ..., 15 is then defined as

 $C_{\text{SCH},i} = (1+j) \times \langle h_m(0) \times z(0), h_m(1) \times z(1), h_m(2) \times z(2), \dots, h_m(255) \times z(255) \rangle,$

where $m = (16 \times i)$ and the leftmost chip in the sequence corresponds to the chip transmitted first in time.

This code word is chosen from every 16^{th} row of the matrix $H_{8.}$, which yields 16 possible codewords.

The Secondary SCH code words are defined in terms of C_{SCH_i} and the definition of $\{C_0, \dots, C_{15}\}$ now follows as:

 $C_i = C_{SCH, i}, i=0,...,15$

7.3 Evaluation of synchronisation codes

The evaluation of information transmitted in SCH on code group and frame timing is shown in table 6, where the 32 code groups are listed. Each code group is containing 4 specific scrambling codes (cf. subclause 6.3), each scrambling code associated with a specific short and long basic midamble code.

Each code group is additionally linked to a specific t_{Offset} , thus to a specific frame timing. By using this scheme, the UE can derive the position of the frame border due to the position of the SCH sequence and the knowledge of t_{Offset} . The complete mapping of Code Group to Scrambling Code, Midamble Codes and t_{Offset} is depicted in table 6.

Table 6: Mapping scheme for Cell Parameters, Code Groups, Scrambling Codes, Midambles and toffset

CELL	Code	A	ssociated Code	S	Associat
PARA- METER	Group	Scrambling Code	Long Basic Midamble Code	Short Basic Midamble Code	ed t _{Offset}
0	Group 0	Code 0	m _{PL0}	m _{SL0}	to
1		Code 1	m _{PL1}	m _{SL1}	
2		Code 2	m _{PL2}	m _{SL2}	
3		Code 3	m _{PL3}	m _{SL3}	
4	Group 1	Code 4	m _{PL4}	m _{SL4}	t ₁
5		Code 5	m _{PL5}	m _{SL5}	
6		Code 6	m _{PL6}	m _{SL6}	
7		Code 7	m _{PL7}	m _{SL7}	
124	Group 31	Code 124	MPL124	m _{SL124}	t ₃₁
125		Code 125	m _{PL125}	m _{SL125}]
126		Code 126	m _{PL126}	m _{SL126}	
127		Code 127	M _{PL127}	m _{SL127}	

For basic midamble codes m_P cf. [7]TS 25.221, annex A 'Basic Midamble Codes'.

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Each cell shall cycle through two sets of cell parameters in a code group with the cell parameters changing each frame. Table 7 shows how the cell parameters are cycled according to the SFN.

Table 7: Alignment of	cell parameter	cycling and SFN
-----------------------	----------------	-----------------

Initial Cell Parameter Assignment	Code Group	Cell Parameter used when SFN mod 2 = 0	Cell Parameter used when SFN mod 2 = 1
0	Group 1 <u>0</u>	0	1
1		1	0
2		2	3
3		3	2
4	Group <u>21</u>	4	5
5		5	4
6		6	7
7		7	6
124	Group <u>3231</u>	124	125
125		125	124
	1	100	107
126		126	127

<u>Annex B (informative):</u> Generalised Hierarchical Golay Sequences

B.1 Alternative generation

The generalised hierarchical Golay sequences for the PSC described in 7.1 may be also viewed as generated (in real valued representation) by the following methods:

Method 1.

The sequence y is constructed from two constituent sequences x_1 and x_2 of length n_1 and n_2 respectively using the following formula:

 $- y(i) = x_2(i \mod n_2) * x_1(i \dim n_2), i = 0 \dots (n_1 * n_2) - 1.$

The constituent sequences x_1 and x_2 are chosen to be the following length 16 (i.e. $n_1 = n_2 = 16$) sequences:

- $\frac{-x_{1} \text{ is defined to be the length 16 (N^{(1)}=4) Golay complementary sequence obtained by the delay matrix D^{(1)} = [8, 4, 1, 2] and weight matrix W^{(1)} = [1, -1, 1, 1].}$
- $\frac{-x_2 \text{ is a generalised hierarchical sequence using the following formula, selecting s=2 and using the two Golay}{\text{ complementary sequences } x_3 \text{ and } x_4 \text{ as constituent sequences. The length of the sequence } x_3 \text{ and } x_4 \text{ is called } n_3 \text{ respectively } n_4.}$
- $x_2(i) = x_4(i \mod s + s^*(i \dim sn_3)) * x_3((i \dim s) \mod n_3), i = 0 \dots (n_3 * n_4) 1.$
- $\frac{-x_3}{2} \text{ and } x_4 \text{ are defined to be identical and the length 4 (N^{(3)} = N^{(4)} = 2) \text{ Golay complementary sequence obtained by}}{\frac{1}{2} \text{ the delay matrix } D^{(3)} = D^{(4)} = [1, 2] \text{ and weight matrix } W^{(3)} = W^{(4)} = [1, 1].}$

The Golay complementary sequences x_1, x_3 and x_4 are defined using the following recursive relation:

$$\underline{a_0(k) = \delta(k) \text{ and } b_0(k) = \delta(k);}$$

$$\underline{a_n(k) = a_{n-1}(k) + W^{(j)} \cdot \underline{b_{n-1}(k-D^{(j)}_{n});}$$

$$\underline{b_n(k) = a_{n-1}(k) - W^{(j)} \cdot \underline{b_{n-1}(k-D^{(j)}_{n});}$$

$$\underline{k = 0, 1, 2, \dots, 2^{**}N^{(j)} - 1;}$$

$$\underline{n = 1, 2, \dots, N^{(j)}.}$$

The wanted Golay complementary sequence x_j is defined by a_n assuming $n=N^{(j)}$. The Kronecker delta function is described by δ , k, j and n are integers.

Method 2

The sequence y can be viewed as a pruned Golay complementary sequence and generated using the following parameters which apply to the generator equations for a and b above:

(a) Let
$$j = 0$$
, $N^{(0)} = 8$.

(b) $[D_1^0, D_2^0, D_3^0, D_4^0, D_5^0, D_6^0, D_7^0, D_8^0] = [128, 64, 16, 32, 8, 1, 4, 2].$

(c) $[W_1^0, W_2^0, W_3^0, W_4^0, W_5^0, W_6^0, W_7^0, W_8^0] = [1, -1, 1, 1, 1, 1, 1].$

(d) For n = 4, 6, set $b_4(k) = a_4(k)$, $b_6(k) = a_6(k)$.

Annex <u>B-C (</u>informative): Change history

					Change history		
Date	TSG #	TSG Doc.	CR	Rev	Subject/Comment	Old	New
14/01/00	RAN_05	RP-99593	-		Approved at TSG RAN #5 and placed under Change Control	-	3.0.0
14/01/00	RAN_06	RP-99696	001	01	Primary and Secondary CCPCH in TDD	3.0.0	3.1.0
14/01/00	RAN_06	RP-99695	003	1	Alignment of Terminology Regarding Spreading for TDD Mode	3.0.0	3.1.0
14/01/00	RAN_06	RP-99696	004	-	Code allocation for Case 3	3.0.0	3.1.0
14/01/00	-	-	-		Change history was added by the editor	3.1.0	3.1.1
31/03/00	RAN_07	RP-000069	002	3	Cycling of cell parameters	3.1.1	3.2.0
31/03/00		RP-000069	005	-	Removal of Synchronisation Case 3 in TDD	3.1.1	3.2.0
31/03/00	RAN_07	RP-000069	006	1	Signal Point Constellation	3.1.1	3.2.0