## TSG-RAN Meeting \#8

Düsseldorf, Germany, 21-23 June 2000
Title: $\quad$ Agreed CRs to TS $\mathbf{2 5 . 2 2 2}$
Source: TSG-RAN WG1
Agenda item: 5.1.3

| No. | Doc \# | Spec | CR | Rev | Subject | Cat | Current_v | New_v |
| :---: | :---: | :---: | :---: | :---: | :--- | :---: | :---: | :---: |
| 1 | R1-000464 | 25.222 | 030 | - | Parity bit attachment to 0 size transport block | B | 3.2 .0 | 3.3 .0 |
| 2 | R1-000465 | 25.222 | 031 | - | Correction of the mapping formula | F | 3.2 .0 | 3.3 .0 |
| 3 | R1-000513 | 25.222 | 034 | - | Alignment of Multiplexing for TDD | F | 3.2 .0 | 3.3 .0 |
| 4 | R1-000716 | 25.222 | 036 | 2 | Bit separation of the Turbo encoded data | D | 3.2 .0 | 3.3 .0 |
| 5 | R1-000751 | 25.222 | 038 | 2 | Revision of code block segmentation description | D | 3.2 .0 | 3.3 .0 |
| 6 | R1-000743 | 25.222 | 039 | - | Editorial corrections in channel coding section | F | 3.2 .0 | 3.3 .0 |

## CHANGE REQUEST

25.222 CR 030 Current Version: 3.2.0

Please see embedded help file at the bottom of this page for instructions on how to fill in this form correctly.
Current Version: 3.2.0

For submission to: TSG RAN\# 8 list expected approval meeting \# here


Form: CR cover sheet, version 2 for 3GPP and SMG
The latest version of this form is available from: ftp://ftp.3gpp.org/Information/CR-Form-v2.doc
Proposed change affects:
(U)SIM $\square$

ME X
UTRAN / Radio $\qquad$ Core Network $\qquad$
(at least one should be marked with an X)

## Source:

TSG RAN WG1
Date: 28, March 2000
Subject: Parity bit attachment to 0 size transport block

## Work item: TS 25.222

## Category:

F Correction
A Corresponds to a correction in an earlier release
(only one category
B Addition of feature
shall be marked
C Functional modification of feature
with an $X$ )
D Editorial modification


Release: Phase 2
Release 96
Release 97
Release 98
Release 99
Release 00


Reason for Alignment with FDD - the attachment of CRC to a transport block of size $=0$ was change: approved for FDD

## Clauses affected: $\quad$ 4.2.1.1

Other specs affected:

| Other 3G core specifications <br> Other GSM core specifications | X | $\rightarrow$ List of CRs: |
| :---: | :---: | :---: |
|  |  | $\rightarrow$ List of CRs: |
| MS test specifications |  | $\rightarrow$ List of CRs: |
| BSS test specifications |  | $\rightarrow$ List of CRs: |
| O\&M specifications |  | $\rightarrow$ List of CRs: |

## Other

 comments:<--------- double-click here for help and instructions on how to create a CR.

### 4.2.1.1 CRC calculation

The entire transport block is used to calculate the CRC parity bits for each transport block. The parity bits are generated by one of the following cyclic generator polynomials:

$$
\begin{aligned}
& \mathrm{gCRC} 24(\mathrm{D})=\mathrm{D}^{24}+\mathrm{D}^{23}+\mathrm{D}^{6}+\mathrm{D}^{5}+\mathrm{D}+1 \\
& \mathrm{~g}_{\mathrm{CRC} 16}(\mathrm{D})=\mathrm{D}^{16}+\mathrm{D}^{12}+\mathrm{D}^{5}+1 \\
& \mathrm{~g}_{\mathrm{CRC} 12}(\mathrm{D})=\mathrm{D}^{12}+\mathrm{D}^{11}+\mathrm{D}^{3}+\mathrm{D}^{2}+\mathrm{D}+1 \\
& \mathrm{~g}_{\mathrm{CRC} 8}(\mathrm{D})=\mathrm{D}^{8}+\mathrm{D}^{7}+\mathrm{D}^{4}+\mathrm{D}^{3}+\mathrm{D}+1
\end{aligned}
$$

Denote the bits in a transport block delivered to layer 1 by $a_{i m 1}, a_{i m 2}, a_{i m 3}, \ldots, a_{i m A_{i}}$, and the parity bits by $p_{\text {im } 1}, p_{\text {im2 } 2}, p_{\text {im } 3}, \ldots, p_{\text {imL }} . A_{i}$ is the length of a transport block of $\operatorname{TrCH} i, m$ is the transport block number, and $L_{i}$ is $24,16,8$, or 0 depending on what is signalled from higher layers.
The encoding is performed in a systematic form, which means that in GF(2), the polynomial

$$
a_{i m 1} D^{A_{i}+23}+a_{i m 2} D^{A_{i}+22}+\ldots+a_{i m A_{i}} D^{24}+p_{i m 1} D^{23}+p_{i m 2} D^{22}+\ldots+p_{i m 23} D^{1}+p_{i m 24}
$$

yields a remainder equal to 0 when divided by $\mathrm{g}_{\text {CRC24 }}(D)$, polynomial

$$
a_{i m 1} D^{A_{i}+15}+a_{i m 2} D^{A_{i}+14}+\ldots+a_{i m A_{i}} D^{16}+p_{i m 1} D^{15}+p_{i m 2} D^{14}+\ldots+p_{i m 15} D^{1}+p_{i m 16}
$$

yields a remainder equal to 0 when divided by $\mathrm{g}_{\mathrm{CRC16}}(D)$, polynomial

$$
a_{i m 1} D^{A_{i}+11}+a_{i m 2} D^{A_{i}+10}+\ldots+a_{i m A_{i}} D^{12}+p_{i m 1} D^{11}+p_{i m 2} D^{10}+\ldots+p_{i m 7} D^{1}+p_{i m 12}
$$

yields a remainder equal to 0 when divided by $\mathrm{g}_{\mathrm{CRC12}}(\mathrm{D})$ and the polynomial

$$
a_{i m 1} D^{A_{i}+7}+a_{i m 2} D^{A_{i}+6}+\ldots+a_{i m A_{i}} D^{8}+p_{i m 1} D^{7}+p_{i m 2} D^{6}+\ldots+p_{i m 7} D^{1}+p_{i m 8}
$$

yields a remainder equal to 0 when divided by $\mathrm{g}_{\mathrm{CRC}}(D)$.
If no transport blocks are input to the CRC calculation ( $M_{i}=0$ ), no CRC attachment shall be done. If transport blocks are input to the CRC calculation $\left(M_{i} \neq 0\right)$ and the size of a transport block is zero $\left.\underline{\left(A_{i}\right.}=0\right)$, CRC shall be attached, i.e. all parity bits equal to zero.

### 4.2.1.2 Relation between input and output of the Cyclic Redundancy Check

The bits after CRC attachment are denoted by $b_{i m 1}, b_{i m 2}, b_{i m 3}, \ldots, b_{i m B_{i}}$, where $B_{i}=A_{i}+L_{i}$. The relation between $a_{i m k}$ and $b_{i m k}$ is:

$$
\begin{aligned}
& b_{i m k}=a_{i m k} \quad k=1,2,3, \ldots, A_{i} \\
& b_{i m k}=p_{i m\left(L_{i}+1-\left(k-A_{i}\right)\right)} \quad k=A_{i}+1, A_{i}+2, A_{i}+3, \ldots, A_{i}+L_{i}
\end{aligned}
$$

## CHANGE REQUEST

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25.222 CR 031

Current Version: V3.2.0

GSM (AA.BB) or 3G (AA.BBB) specification number $\uparrow$
$\uparrow$ CR number as allocated by MCC support team
For submission to: TSG RAN\#8
list expected approval meeting \# here

| for approval |  |
| ---: | ---: |
| for information | $\mathbf{X}$ |
|  |  |

$\square$ (for SMG
$\uparrow$ for information strategic
non-strategic use only)

Form: CR cover sheet, version 2 for 3GPP and SMG
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or for SMG, use the format P-99-xxx

nge affects:
(U)SIM $\square$ ME X
UTRAN / Radio $\qquad$ X

Core Network $\qquad$
$\frac{\text { Proposed change affects: }}{\text { (at least one should be marked with an } X}$

Source: TSG RAN WG1
Date: March 28, 2000
Subject: Correction of the mapping formula

## Work item: TS 25.222

Category: F Correction
A Corresponds to a correction in an earlier release
B Addition of feature
C Functional modification of feature
D Editorial modification

| $\mathbf{X}$ |
| :---: |
|  |
|  |
|  |
|  |

Release: Phase 2 Release 96 Release 97 Release 98
Release 99 Release 00


Reason for $\quad$ This CR corrects an error in the pseudo code used for channel mapping scheme. change:

Clauses affected:
4.2.11.1

Other specs
Other 3G core specifications
affected:
Other GSM core specifications MS test specifications BSS test specifications O\&M specifications

$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:
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$\rightarrow$ List of CRs:

## Other comments: <br> help.doc

<--------- double-click here for help and instructions on how to create a CR.

### 4.2.11.1 Mapping scheme

Notation used in this section:
$P_{\mathrm{t}}$ : number of physical channels for timeslot $\mathrm{t}, P_{t}=1 . .2$ for uplink ; $P_{t}=1 \ldots 16$ for downlink
$U_{t p}$ : capacity in bits for the physical channel p in timeslot t
$U_{t}$ : total number of bits to be assigned for timeslot t
$\mathrm{bs}_{\mathrm{p}}$ : number of consecutive bits to assign per code
for downlink all $\mathrm{bs}_{\mathrm{p}}=1$
for uplink if SF1 >= SF2 then $\mathrm{bs}_{1}=1 ; \mathrm{bs}_{2}=\mathrm{SF} 1 / \mathrm{SF} 2$;
if SF2 > SF1 then $\mathrm{bs}_{1}=\mathrm{SF} 2 / \mathrm{SF} 1 ; \mathrm{bs}_{2}=1$;
$\mathrm{fb}_{\mathrm{p}}$ : number of already written bits for each code
pos: intermediate calculation variable
for $\mathrm{p}=1$ to $P_{\mathrm{t}}$
-- reset number of already
written bits for every physical channel

$$
\mathrm{fb}_{\mathrm{p}}=0
$$

end for
$\mathrm{p}=1$
-- start with PhCH \#1
for $\mathrm{k}=1$ to $U_{t}$.
do while $\left(\mathrm{fb}_{\mathrm{p}}==U_{t p}\right) \quad$-- physical channel
filled up already?

```
\(-\mathrm{p}=\left((\mathrm{p}+1) \bmod \left(P_{t}+1\right)\right)+1\);
\(\mathrm{p}=\left(\mathrm{p} \bmod \mathrm{P}_{\mathrm{t}}\right)+1 ;\)
end do
```

if $(p \bmod 2)=0$
$\operatorname{pos}=U_{t p}-\mathrm{fb}_{\mathrm{p}} \quad$-- reverse order
else

endif

$$
\begin{array}{lr}
w_{\mathrm{tp}, \mathrm{pos}}=v_{\mathrm{t}, \mathrm{k}} & --\mathrm{assignment} \\
\mathrm{fb}_{\mathrm{p}}=\mathrm{fb}_{\mathrm{p}}+1 & -- \text { Incre }
\end{array}
$$

already written bits

$$
\text { if }\left(\mathrm{fb}_{\mathrm{p}} \bmod \mathrm{bs}_{\mathrm{p}}\right)==0
$$

to the next physical channel

$$
p=\left((p+1) \bmod \left(P_{t}+1\right)\right)+1
$$

$1 \quad \mathrm{p}=\left(\mathrm{p} \bmod \mathrm{P}_{\mathrm{t}}\right)+1 ;$
end if
end for


Reason for Some modifications have been done in order to align the TDD specification mode with change: the FDD spec. . The case of TFCS reconfiguration where a transport channel is removed from the CCTrCH is now described. Some formulae have been revised in order to avoid ambiguities in the order of calculation.

Clauses affected: $\quad 4.2 .4,4.2 .6,4.2 .7 .1,4.2 .7 .2,4.2 .12$


## Other comments:

### 4.2.4 Radio frame size equalisation

Radio frame size equalisation is padding the input bit sequence in order to ensure that the output can be segmented in $F_{i}$ data segments of same size as described in the subclause 4.2.6.

The input bit sequence to the radio frame size equalisation is denoted by $c_{i 1}, c_{i 2}, c_{i 3}, \ldots, c_{i E_{i}}$, where $i$ is $\operatorname{TrCH}$ number and $E_{i}$ the number of bits. The output bit sequence is denoted by $t_{i 1}, t_{i 2}, t_{i 3}, \ldots, t_{i T_{i}}$, where $T_{i}$ is the number of bits. The output bit sequence is derived as follows:
$t_{i k}=c_{i k}$, for $\mathrm{k}=1 \ldots E_{i}$ and
$t_{i k}=\left\{0 \vdash_{2} 1\right\}$ for $\mathrm{k}=E_{i}+1 \ldots T_{i}$, if $E_{i}<T_{i}$
where
$\mathrm{T}_{i}=F_{i} * N_{i}$ and
$N_{i}=\left\lfloor\left(E_{i}-1\right) / F_{i}\right\rfloor+1 N_{i}=\left\lceil E_{i} / F_{i}\right\rceil$ is the number of bits per segment after size equalisation.

### 4.2.6 Radio frame segmentation

When the transmission time interval is longer than 10 ms , the input bit sequence is segmented and mapped onto consecutive $\underline{F}_{i}$ radio frames. Following radio frame size equalisation the input bit sequence length is guaranteed to be an integer multiple of $F_{i}$.

The input bit sequence is denoted by $x_{i 1}, x_{i 2}, x_{i 3}, \ldots, x_{i X_{i}}$ where $i$ is the $\operatorname{TrCH}$ number and $X_{i}$ is the number bits. The $F \dot{t}_{\underline{i}}$ output bit sequences per TTI are denoted by $y_{i, n_{i} 1}, y_{i, n_{i} 2}, y_{i, n_{i} 3}, \ldots, y_{i, n_{i} Y_{i}}$ where $n_{i}$ is the radio frame number in current TTI and $Y_{i}$ is the number of bits per radio frame for $\operatorname{TrCH} i$. The output sequences are defined as follows:

$$
y_{i, n_{i} k}=x_{i,\left(\left(n_{i}-1\right) Y_{i}\right)+k}, n_{i}=1 \ldots F_{i}, k=1 \ldots Y_{i}
$$

where
$Y_{i}=\left(X_{i} / F_{i}\right)$ is the number of bits per segment $\overline{,}$.
$x_{i k}$ is the $\mathrm{k}^{\text {th }}$ bit of the input bit sequence and
$y_{i, n_{i} k}$ is the $\mathrm{k}^{\text {th }}$ bit of the output bit sequence correspending to the $\mathrm{n}^{\text {th }}$ radio frame

The $n_{i}$-th segment is mapped to the $n_{i}$-th radio frame of the transmission time interval.
The input bit sequence to the radio frame segmentation is denoted by $d_{i 1}, d_{i 2}, d_{i 3}, \ldots, d_{i T_{i}}$, where $i$ is the $\operatorname{TrCH}$ number and $T_{i}$ the number of bits. Hence, $x_{i k}=d_{i k}$ and $X_{i}=T_{i}$.

The output bit sequence corresponding to radio frame $n_{i}$ is denoted by $e_{i 1}, e_{i 2}, e_{i 3}, \ldots, e_{i N_{i}}$, where $i$ is the $\operatorname{TrCH}$ number and $N_{i}$ is the number of bits. Hence, $e_{i, k}=y_{i, n_{i} k}$ and $N_{i}=Y_{i}$.

### 4.2.7.1 Determination of rate matching parameters

The following relations, defined for all $\mathrm{TFC} j$, are used when calculating the rate matching pattern:

$$
Z_{0, j}=0
$$

$$
Z_{i j}=\left\lfloor\begin{array}{|l}
\sum_{m=1}^{i} R M_{m} \cdot N_{m j} \\
\sum_{m=1}^{I} R M_{m} \cdot N_{m j}
\end{array} N_{\text {data, } j} Z_{i j}=\left\lfloor\frac{\left\{\left(\sum_{m=1}^{i} R M_{m} \cdot N_{m j}\right) \cdot N_{d a t a, j}\right\}}{\sum_{m=1}^{I} R M_{m} \cdot N_{m j}}\right\rfloor \text { for all i }=1\right. \text {.. I }
$$

$$
\Delta N_{i j}=Z_{i j}-Z_{i-1, j}-N_{i j} \quad \text { for all } \mathrm{i}=1 . . \mathrm{I}
$$

Puncturing can be used to minimise the required transmission capacity. The maximum amount of puncturing that can be applied is signalled from higher layers and denoted by PL. The possible values for $\mathrm{N}_{\text {data }}$ depend on the number of physical channels $\mathrm{P}_{\text {max }}$, allocated to the respective CCTrCH , and on their characteristics (spreading factor, length of midamble and TFCI, usage of TPC and multiframe structure), which is given in [7].

Denote the number of data bits in each physical channel by $U_{p, S p}$, where $p$ refers to the sequence number $1 \leq p \leq P_{\max }$ of this physical channel in the allocation message, and the second index $S p$ indicates the spreading factor with the possible values $\{16,8,4,2,1\}$, respectively. For each physical channel an individual minimum spreading factor $S p_{\text {min }}$ is transmitted by means of the higher layer. Then, for $N_{\text {data }}$ one of the following values in ascending order can be chosen:
$\left\{U_{1,16}, \ldots, U_{1, S 1_{\min }}, U_{1, S 1_{\min }}+U_{2,16}, \ldots, U_{1, S 1_{\min }}+U_{2, S 2_{\min }}, \ldots, U_{1, S 1_{\min }}+U_{2, S 2_{\min }}+\ldots+U_{P_{\max }, 16}, \ldots, U_{1, S 1_{\min }}+U_{2, S 2_{\min }}+\ldots+U_{P_{\max }}\left(S P_{\max }\right)_{\min }\right\}$
$\mathrm{N}_{\mathrm{data}, \mathrm{j}}$ for the transport format combination j is determined by executing the following algorithm:

$$
\begin{aligned}
& \text { SET1 }=\left\{\mathrm{N}_{\text {data }} \text { such that } N_{\text {datata }}-P L \cdot \sum_{x=1}^{I} \frac{R M_{x}}{\min _{1 \leq y \leq I}\left\{R M_{y}\right\}} \cdot N_{x, j}^{\min _{1 \leq y \leq I}\left\{R M_{y}\right\} \cdot N_{\text {data }}-P L \cdot \sum_{x=1}^{I} R M_{x} \cdot N_{x, j} \text { is non }}\right. \\
& \quad \text { negative }\}
\end{aligned}
$$

$$
\mathrm{N}_{\mathrm{data}, \mathrm{j}}=\min \mathrm{SET} 1
$$

The number of bits to be repeated or punctured, $\Delta \mathrm{N}_{\mathrm{ij}}$, within one radio frame for each TrCH i is calculated with the relations given at the beginning of this subclause for all possible transport format combinations $j$ and selected every radio frame.

If $\Delta \mathrm{N}_{\mathrm{ij}}=0$ then the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.3 does not need to be executed.

Otherwise, the rate matching pattern is calculated with the algorithm described in subclause 4.2.7.3. For this algorithm the parameters $\mathrm{e}_{\mathrm{ini}}, \mathrm{e}_{\mathrm{plus}}, \mathrm{e}_{\text {minus }}$, and $X_{i}$ are needed, which are calculated according to the equations in subclauses 4.2.7.1.1 and 4.2.7.1.2.

### 4.2.7.2 Bit separation and collection for rate matching

The systematic bits (excluding bits for trellis termination) of turbo encoded TrCHs shall not be punctured. The systematic bit, first parity bit, and second parity bit in the bit sequence input to the rate matching block are therefore separated from each other. Puncturing is only applied to the parity bits and systematic bits used for trellis termination.

The bit separation function is transparent for uncoded TrCHs , convolutionally encoded TrCHs , and for turbo encoded TrCHs with repetition. The bit separation and bit collection are illustrated in figures 4 and 5 .


Figure 4: Puncturing of turbo encoded TrCHs


Figure 5: Rate matching for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition

The bit separation is dependent on the $1^{\text {st }}$ interleaving and offsets are used to define the separation for different TTIs. The offsets $\alpha_{b}$ for the systematic $(b=1)$ and parity bits $(b \in\{2,3\})$ are listed in table 4.

Table 4: TTI dependent offset needed for bit separation

| TTI (ms) | $\alpha_{1}$ | $\alpha_{2}$ | $\alpha_{3}$ |
| :---: | :---: | :---: | :---: |
| 10,40 | 0 | 1 | 2 |
| 20,80 | 0 | 2 | 1 |

The bit separation is different for different radio frames in the TTI. A second offset is therefore needed. The radio frame number for $\operatorname{TrCH} i$ is denoted by $n_{i}$. and the offset by $\beta_{n_{i}}$.

Table 5: Radio frame dependent offset needed for bit separation

| TTI (ms) | $\beta_{0}$ | $\beta_{1}$ | $\beta_{2}$ | $\beta_{3}$ | $\beta_{4}$ | $\beta_{5}$ | $\beta_{6}$ | $\beta_{7}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 10 | 0 | NA | NA | NA | NA | NA | NA | NA |
| 20 | 0 | 1 | NA | NA | NA | NA | NA | NA |
| 40 | 0 | 1 | 2 | 0 | NA | NA | NA | NA |
| 80 | 0 | 1 | 2 | 0 | 1 | 2 | 0 | 1 |

### 4.2.12 Multiplexing of different transport channels onto one CCTrCH, and mapping of one CCTrCH onto physical channels

Different transport channels can be encoded and multiplexed together into one Coded Composite Transport Channel $(\mathrm{CCTrCH})$. The following rules shall apply to the different transport channels which are part of the same CCTrCH :

1) Transport channels multiplexed into one CCTrCh shall have co-ordinated timings. When the TFCS of a CCTrCH is changed because ane or more transport channels $\dot{t}$ are ${ }^{\text {a }}$ added to the CCTrCH or reconfigured within the CCTrCH , or removed from the CCTrCH , the change may only be made at the TTI of transpert ehannel $i$ may only start in-of a radio frames with CFN fulfilling the relation
$\mathrm{CFN}_{\mathrm{i}} \bmod \mathrm{F}_{\max }=0$,
where $\mathrm{F}_{\max }$ denotes the maximum number of radio frames within the transmission time intervals of all transport channels which are multiplexed into the same CCTrCH , including any transport channels $i$ which isare added er reconfigured or have been removed, and $\mathrm{CFN}_{\mathrm{i}}$ denotes the connection frame number of the first radio frame of the changed CCTrCH.within the transmission time interval of transport channel $i$.

After addition or reconfiguration of a transport channel $i$ within a CCTrCH , the TTI of transport channel $i$ may only start in radio frames with CFN fulfilling the relation
$\mathrm{CFN}_{\mathrm{i}} \bmod \mathrm{F}_{\mathrm{i}}=0$.
2) Different CCTrCHs cannot be mapped onto the same physical channel.
3) One CCTrCH shall be mapped onto one or several physical channels.
4) Dedicated Transport channels and common transport channels cannot be multiplexed into the same CCTrCH.
5) For the common transport channels, only the FACH and PCH may belong to the same CCTrCH .
6) Each CCTrCH carrying a BCH shall carry only one BCH and shall not carry any other Transport Channel.
7) Each CCTrCH carrying a RACH shall carry only one RACH and shall not carry any other Transport Channel.

Hence, there are two types of CCTrCH .
CCTrCH of dedicated type, corresponding to the result of coding and multiplexing of one or several DCH.
CCTrCH of common type, corresponding to the result of the coding and multiplexing of a common channel, i.e. RACH and USCH in the uplink and DSCH, BCH, FACH or PCH in the downlink, respectively.

Transmission of TFCI is possible for CCTrCH containing Transport Channels of:

- dedicated type;
- USCH type;
- DSCH type;
- FACH and/or PCH type.

For submission to: TSG RAN \#8 list expected approval meeting \# here

| for approval |  |
| ---: | ---: |
| for information | $\mathbf{X}$ |
|  |  |

(for SMG use only)

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Proposed change affects:
(U)SIM $\square$ ME $\square$ UTRAN / Radio $\mathbf{X}$ Core Network $\qquad$
(at least one should be marked with an X)

Source: TSG RAN WG1
Date: 23, May 2000
Subject: $\quad$ Bit separation of the Turbo encoded data

## Work item: TS 25.222

Category:
(only one category
shall be marked
with an $X$ )

Reason for change:

F Correction
A Corresponds to a correction in an earlier release
B Addition of feature
C Functional modification of feature
D Editorial modification


Release: Phase 2 Release 96 Release 97 Release 98 Release 99 Release 00


Clauses affected: $\quad 4.2 .7 .2,4.2 .7 .2 .1$

Other specs Other 3G core specifications affected: Other GSM core specifications MS test specifications BSS test specifications O\&M specifications

$\rightarrow$ List of CRs: 25.212
$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:

## Other comments: <br> help.doc

<--------- double-click here for help and instructions on how to create a CR.

### 4.2.7.2 Bit separation and collection for rate matching

The systematic bits (exeluding bits for trellis termination) of turbo encoded TrCHs shall not be punctured-, however systematic bits for trellis termination may be punctured. The systematic bits, first parity bits, and second parity bits in the bit sequence input to the rate matching block are therefore separated from each other into three sequences, one sequence containing all of the systematic bits and some systematic, first and second parity trellis termination bits; the second sequence containing all of the first parity bits and some systematic, first and second parity trellis termination bits and the third sequence containing all of the second parity bits and some systematic, first and second parity trellis termination bits. Puncturing is only applied to the second and third sequences. parity bits and systematic bits used for trellis termination.
The bit separation function is transparent for uncoded TrCHs , convolutionally encoded TrCHs , and for turbo encoded TrCHs with repetition. The bit separation and bit collection are illustrated in figures 4 and 5.


Figure 4: Puncturing of turbo encoded TrCHs


Figure 5: Rate matching for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition.

The bit separation is dependent on the $1^{\text {st }}$ interleaving and offsets are used to define the separation for different TTIs. The sequence denoted as $b=1$ contains all of the systematic bits and some systematic, first and second parity trellis termination bits; the sequence denoted as $b=2$ contains all of the first parity bits and some systematic, first and second parity trellis termination bits; the sequence denoted as $b=3$ contains all of the second parity bits and some systematic, first and second parity trellis termination bits. The offsets $\alpha_{b}$ for the these sequences systematic ( $b=1$ ) and parity bits ( $b \in\{2,3\}$ ) are listed in table 4.

Table 4 : TTI dependent offset needed for bit separation

| TTI (ms) | $\alpha_{1}$ | $\alpha_{2}$ | $\alpha_{3}$ |
| :---: | :---: | :---: | :---: |
| 10,40 | 0 | 1 | 2 |
| 20,80 | 0 | 2 | 1 |

The bit separation is different for different radio frames in the TTI. A second offset is therefore needed. The radio frame number for $\operatorname{TrCH} i$ is denoted by $n_{i}$. and the offset by $\beta_{n_{i}}$.

Table 5 : Radio frame dependent offset needed for bit separation

| TTI (ms) | $\beta_{0}$ | $\beta_{1}$ | $\beta_{2}$ | $\beta_{3}$ | $\beta_{4}$ | $\beta_{5}$ | $\beta_{6}$ | $\beta_{7}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 10 | 0 | NA | NA | NA | NA | NA | NA | NA |
| 20 | 0 | 1 | NA | NA | NA | NA | NA | NA |
| 40 | 0 | 1 | 2 | 0 | NA | NA | NA | NA |
| 80 | 0 | 1 | 2 | 0 | 1 | 2 | 0 | 1 |

### 4.2.7.2.1 Bit separation

The bits input to the rate matching are denoted by $e_{i 1}, e_{i 2}, e_{i 3}, \ldots, e_{i N_{i}}$, where $i$ is the $\operatorname{TrCH}$ number and $N_{i}$ is the number of bits input to the rate matching block. Note that the transport format combination number $j$ for simplicity has been left out in the bit numbering, i.e. $N_{i}=N_{i j}$. The bits after separation are denoted by $x_{b i 1}, x_{b i 2}, x_{b i 3}, \ldots, x_{b i X_{i}}$. For turbo encoded TrCHs with puncturing, $b$ indicates the three sequences defined in section 4.2.7.2 systematic, first parity, or second parity bit. The sequence denoted as $b=1$ contains all of the systematic bits and some systematic, first and second parity trellis termination bits; the sequence denoted as $\mathrm{b}=2$ contains all of the first parity bits and some systematic, first and second parity trellis termination bits; the sequence denoted as $b=3$ contains all of the second parity bits and some systematic, first and second parity trellis termination bits.
For all other cases $b$ is defined to be $1 . X$ is the number of bits in each separated bit sequence. The relation between $e_{i k}$ and $x_{b i k}$ is given below.
For turbo encoded TrCHs with puncturing:
$x_{1, i, k}=e_{i, 3(k-1)+1+\left(\alpha_{1}+\beta_{n i}\right) \bmod 3} \quad k=1,2,3, \ldots, X_{i} \quad X_{i}=$
$\left\lfloor N_{i} / 3\right\rfloor$
$x_{1, i,\left[N_{i} / 3\right]+k}=e_{i, 3\left\lfloor N_{i} / 3\right]_{+k}} \quad k=1, \ldots, N_{i} \bmod 3 \quad$ Note: When $\left(N_{i} \bmod \right.$
$3)=0$ this row is not needed.
$x_{2, i, k}=e_{i, 3(k-1)+1+\left(\alpha_{2}+\beta_{n_{i}}\right) \bmod 3}$
$k=1,2,3, \ldots, X_{i}$
$X_{i}=\left\lfloor N_{i} / 3\right\rfloor$
$x_{3, i, k}=e_{i, 3(k-1)+1+\left(\alpha_{3}+\beta_{n}\right) \bmod 3}$
$k=1,2,3, \ldots, X_{i}$
$X_{i}=\left\lfloor N_{i} / 3\right\rfloor$
For uncoded TrCHs , convolutionally encoded TrCHs , and turbo encoded TrCH with repetition:
$x_{1, i, k}=e_{i, k}$

$$
k=1,2,3, \ldots, X_{i}
$$

$$
X_{i}=N_{i}
$$

### 25.222 CR 038r2

Current Version:
3.2.0

GSM (AA.BB) or 3G (AA.BBB) specification number $\uparrow$
$\uparrow C R$ number as allocated by MCC support team

For submission to: TSG RAN \#8
list expected approval meeting \# here $\qquad$
X for information

The latest version of
Form: CR cover sheet, version 2 for 3GPP and SMG
The la
Proposed change affects:
(U)SIM $\square$ ME $\mathbf{X}$
UTRAN / Radio $\qquad$ Core Network $\square$
(at least one should be marked with an X)

Source:
TSG RAN WG1
Subject: Revision of Code Block Segmentation Description

## Work item:

## Category:

(only one category
F Correction
A Corresponds to a correction in an earlier release shall be marked with an X)

C Functional modification of feature
D Editorial modification


Release: Phase 2 Release 96 Release 97 Release 98 Release 99 Release 00


Reason for The insertion of filler bits within the first block of code block segmentation does not change:
depend on the conditions $X_{i} \leq Z$ or $X_{i}>Z$. In that way the current specification is redundant. This is changed in the proposed modification. Additionally by reformulating some formulas a more precise description is given.

Clauses affected: $\quad$ Chapter 4.2.2.2


## Other <br> comments:

<--------- double-click here for help and instructions on how to create a CR.

### 4.2.2.2 Code block segmentation

Segmentation of the bit sequence from transport block concatenation is performed if $X_{i}>Z$. The code blocks after segmentation are of the same size. The number of code blocks on $\operatorname{TrCH} i$ is denoted by $C_{i}$. If the number of bits input to the segmentation, $X_{i}$, is not a multiple of $C_{i}$, filler bits are added to the beginning of the first block. If turbo coding is selected and $X_{i}<40$, filler bits are added to the beginning of the code block. The filler bits are transmitted and they are always set to 0 . The maximum code block sizes are:

- convolutional coding: $Z=504$;
- turbo coding: $Z=5114$;
- no channel coding: $Z=$ unlimited.

The bits output from code block segmentation are denoted by $o_{i r 1}, o_{i r 2}, o_{i r 3}, \ldots, o_{i r K_{i}}$, where $i$ is the $\operatorname{TrCH}$ number, $r$ is the code block number, and $K_{i}$ is the number of bits.

Number of code blocks: $C_{i}=\left\lceil X_{i} / Z\right\rceil$
Number of bits in each code block:
if $X_{i}<40$ and Turbo coding is used, then

$$
K_{i}=40
$$

else

$$
\begin{aligned}
& \qquad K_{i}=\left\lceil X_{i} / C_{i}\right\rceil \\
& \text { end if }
\end{aligned}
$$

Number of filler bits: $Y_{i}=C_{i} K_{i}-X_{i}$
If $X_{i} \leq Z$, then
$\theta_{i 1 k}=0-k=1,2, \ldots, Y_{i}$
for $k=1$ to $Y_{i} \quad$-- Insertion of filler bits
$\ldots o_{i 1 k}=0$
end for
$\theta_{i 1 k}=x_{i,\left(k-Y_{i}\right)}-k=Y_{i}+1, Y_{i}+2, \ldots, K_{i}$
for $k=Y_{i}+1$ to $K_{\underline{i}}$
$\quad_{i 1 k}=x_{i,\left(k-Y_{i}\right)}$
end for
end if
If $X_{i}>Z$, then
$o_{i 1 k}=0-k=1,2, \ldots, Y_{i}$
$\theta_{i 1 k}=x_{i,\left(k-Y_{i}\right)} \quad k=Y_{i}+1, Y_{i}+2, \ldots, K_{i}$
$r=2$ -- Segmentation
while $r \leq C_{i}^{i}$
for $k=1$ to $K_{\underline{i}}$

$$
o_{i 2 k}=x_{i,\left(k+K_{i}-Y_{i}\right)} o_{i r k}=x_{i,\left(k+(r-1) \cdot K_{i}-Y_{i}\right)-} k=1,2, \ldots, K_{i}
$$

end for
$r=r+1$
end while $\theta_{i 3 k}=x_{i,\left(k+2 K_{i}-Y_{i}\right)} \quad k=1,2, \ldots, K_{i}$
$o_{i C_{i} k}=x_{i,\left(k+\left(C_{i}-1\right) K_{i}-Y_{i}\right)} k=1,2, \ldots, K_{i}$
end if

## CHANGE REQUEST

Please see embedded help file at the bottom of this page for instructions on how to fill in this form correctly.
25.222 CR 039

Current Version: 3.2.0

GSM (AA.BB) or $3 G$ (AA.BBB) specification number $\uparrow$
$\uparrow$ CR number as allocated by MCC support team
For submission to: TSG RAN \#8 list expected approval meeting \# here

| for approval |  |
| ---: | ---: |
| for information | $\mathbf{X}$ |
|  |  |

Strategic $\square$ (for SMG non-strategic use only)

Form: CR cover sheet, version 2 for 3GPP and SMG
$\qquad$ (at least one should be marked with an $X$ )

TSG RAN WG1
Date: 24-May-2000
Subject: Editorial corrections in channel coding section

## Work item:

Category:
F Correction
A Corresponds to a correction in an earlier release
(only one category
B Addition of feature
shall be marked
with an $X$ )
C Functional modification of feature
D Editorial modification

| $\mathbf{X}$ |
| :---: |
|  |
|  |
|  |
|  |

Release: Phase 2
Release 96
Release 97
Release 98
Release 99
Release 00

$\begin{array}{ll}\text { Reason for } & \text { To give editorial corrections on descriptions of channel coding including Turbo code } \\ \text { change: } & \text { internal interleaver. }\end{array}$

Clauses affected: $\quad 4.2 .3$ of TS25.222
Other specs
Other 3G core specifications affected: Other GSM core specifications MS test specifications BSS test specifications O\&M specifications

$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:
$\rightarrow$ List of CRs:

## Other

 comments:end if

$$
\begin{aligned}
& \text { If } X_{i}>Z \text {, then } \\
& \qquad \begin{array}{l}
o_{i 1 k}=0 \quad k=1,2, \ldots, Y_{i} \\
o_{i 1 k}=x_{i,\left(k-Y_{i}\right)} \quad k=Y_{i}+1, Y_{i}+2, \ldots, K_{i} \\
o_{i 2 k}=x_{i,\left(k+K_{i}-Y_{i}\right)} k=1,2, \ldots, K_{i} \\
o_{i 3 k}=x_{i,\left(k+2 K_{i}-Y_{i}\right)} \quad k=1,2, \ldots, K_{i} \\
\ldots \\
o_{i C_{i} k}=x_{i,\left(k+\left(C_{i}-1\right) K_{i}-Y_{i}\right)} \quad k=1,2, \ldots, K_{i}
\end{array}
\end{aligned}
$$

end if

### 4.2.3 Channel coding

Code blocks are delivered to the channel coding block. They are denoted by $o_{i r 1}, o_{i r 2}, o_{i r 3}, \ldots, o_{i r K_{i}}$, where $i$ is the $\operatorname{TrCH}$ number, $r$ is the code block number, and $K_{i}$ is the number of bits in each code block. The number of code blocks on $\operatorname{TrCH} i$ is denoted by $C_{i}$. After encoding the bits are denoted by $y_{i r 1}, y_{i r 2}, y_{i r 3}, \ldots, y_{i r Y_{i}}$, where $Y_{i}$ is the number of encoded bits.

The relation between $o_{i r k}$ and $y_{i r k}$ and between $K_{i}$ and $Y_{i}$ is dependent on the channel coding scheme.
The following channel coding schemes can be applied to transport channels:

- convolutional coding;
- turbo coding;
- no coding.

Usage of coding scheme and coding rate for the different types of TrCH is shown in table 1 . The values of $Y_{i}$ in connection with each coding scheme:

- convolutional coding with rate $1 / 2: Y_{i}=2 * K_{i}+16$; rate $1 / 3: Y_{i}=3 * K_{i}+24$;
- turbo coding with rate $1 / 3: Y_{i}=3 * K_{i}+12$;
- no coding: $Y_{i}=K_{i}$.

Table 1: Usage of channel coding scheme and coding rate

| Type of TrCH | Coding scheme | Coding rate |
| :---: | :---: | :---: |
| BCH | Convolutional coding | 1/2 |
| PCH |  |  |
| RACH |  |  |
| DCH, DSCH, FACH, USCH |  | 1/3, 1/2 |
|  | Turbo coding | 1/3 |
|  | No coding |  |

If no code blocks are input to the channel coding $\left(C_{t}=0\right)$, no bits shall be output from the channel coding, i.e. $E_{t}=0$.

### 4.2.3.1 Convolutional coding

Convolutional codes with constraint length 9 and coding rates $1 / 3$ and $1 / 2$ are defined.
The configuration of the convolutional coder is presented in figure 2.
Output from the rate $1 / 3$ convolutional coder shall be done in the order output0, output1, output2, output0, output1, output 2 , output $0, \ldots$,output 2 . Output from the rate $1 / 2$ convolutional coder shall be done in the order output 0 , output 1 , output 0 , output 1 , output $0, \ldots$, output 1 .
8 tail bits with binary value 0 shall be added to the end of the code block before encoding.

The initial value of the shift register of the coder shall be "all 0 " when starting to encode the input bits.

(a) Rate $1 / 2$ convolutional coder

(b) Rate $1 / 3$ convolutional coder

Figure 2: Rate $1 / 2$ and rate $1 / 3$ convolutional coders

### 4.2.3.2 Turbo coding

### 4.2.3.2.1 Turbo coder

The scheme of Turbo coder is a Parallel Concatenated Convolutional Code (PCCC) with two 8-state constituent encoders and one Turbo code internal interleaver. The coding rate of Turbo coder is $1 / 3$. The structure of Turbo coder is illustrated in figure 3.
The transfer function of the 8 -state constituent code for PCCC is:

$$
\mathrm{G}(\mathrm{D})=\left[1, \frac{g_{1}(D)}{g_{0}(D)}\right],
$$

where

$$
\begin{aligned}
& g_{0}(D)=1+D^{2}+D^{3}, \\
& g_{1}(D)=1+D+D^{3} .
\end{aligned}
$$

The initial value of the shift registers of the 8 -state constituent encoders shall be all zeros when starting to encode the input bits.
Output from the Turbo coder is , $\mathrm{Y}^{\prime}(0), \mathrm{X}(1), \mathrm{Y}(1), \mathrm{Y}^{\prime}(1)$, etc:

$$
x_{1}, z_{1}, z_{1}^{\prime}, x_{2}, z_{2}, z_{2}^{\prime}, \ldots, x_{K}, z_{K}, z_{K}^{\prime},
$$

where $x_{1}, x_{2}, \ldots, x_{K}$ are the bits input to the Turbo coder i.e. both first 8 -state constituent encoder and Turbo code internal interleaver, and $K$ is the number of bits, and $z_{1}, z_{2}, \ldots, z_{K}$ and $z_{1}^{\prime}, z_{2}^{\prime}, \ldots, z_{K}^{\prime}$ are the bits output from first and second 8 -state constituent encoders, respectively.
The bits output from Turbo code internal interleaver are denoted by $x_{1}^{\prime}, x_{2}^{\prime}, \ldots, x_{K}^{\prime}$, and these bits are to be input to the second 8 -state constituent encoder.


Figure 3: Structure of rate $1 / 3$ Turbo coder (dotted lines apply for trellis termination only)

### 4.2.3.2.2 Trellis termination for Turbo coder

Trellis termination is performed by taking the tail bits from the shift register feedback after all information bits are encoded. Tail bits are padded after the encoding of information bits.
The first three tail bits shall be used to terminate the first constituent encoder (upper switch of figure 3 in lower position) while the second constituent encoder is disabled. The last three tail bits shall be used to terminate the second constituent encoder (lower switch of figure 3 in lower position) while the first constituent encoder is disabled.
The transmitted bits for trellis termination shall then be:

$$
\left.x_{K+1}, z_{K+1}, x_{K+2}, z_{K+2}, x_{K+3}, z_{K+3}, x_{K+1}^{\prime}, z_{K+1}^{\prime}, x_{K+2}^{\prime}, z_{K+2}^{\prime}, x_{K+3}^{\prime}, z_{K+3}^{\prime}\right) .
$$

### 4.2.3.2.3 Turbo code internal interleaver

The Turbo code internal interleaver consists of bits-input to a rectangular matrix, intra-row and inter-row permutations of the rectangular matrix, and bits-output from the rectangular matrix with pruning. The bits input to the Turbo code internal interleaver are denoted by $x_{1}, x_{2}, x_{3}, \ldots, x_{K}$, where $K$ is the integer number of the bits and takes one value of $40 \leq K \leq$ 5114. The relation between the bits input to the Turbo code internal interleaver and the bits input to the channel coding is defined by $x_{k}=o_{i r k}$ and $K=K_{i}$.

| The following subclause specific symbols are used in subclauses 4.2.3.2.3.1 to 4. |  |
| :--- | :--- |
| $K$ | Number of bits input to Turbo code internal interleaver |
| $R$ | Number of rows of rectangular matrix |
| $C$ | Number of columns of rectangular matrix |
| $p$ | Prime number |
| $v$ | Primitive root |
| $s(i)$ | Base sequence for intra-row permutation |
| $q_{j}$ | Minimum prime integers |
| $r_{j}$ | Permuted prime integers |
| $T(j)$ | Inter-row permutation pattern |
| $U_{j}(i)$ | Intra-row permutation pattern |
| $i$ | Index of matrix |
| $j$ | Index of matrix |
| $k$ | Index of bit sequence |

4.2.3.2.3.1

Bits-input to rectangular matrix
The bit sequence input to the Turbo code internal interleaver $x_{k}$ is written into the rectangular matrix as follows.
(1) Determine the number of rows $R$ of the rectangular matrix such that:
$R=\left\{\begin{array}{l}5, \text { if }(40 \leq K \leq 159) \\ 10, \text { if }((160 \leq K \leq 200) \text { or }(481 \leq K \leq 530)) \\ 20, \text { if }(K=\text { any other value })\end{array}\right.$
where the rows of rectangular matrix are numbered $0,1,2, \ldots, R-1$ from top to bottom.
(2) Determine the number of columns $C$ of rectangular matrix such that: if $(481 \leq K \leq 530)$ then

$$
p=53 \text { and } C=p
$$

else
Find minimum prime p such that,

$$
(p+1)-K / R \geq 0
$$

and determine $C$ such that
if $p-K / R \geq 0$ ) then
if $(p-1-K / R \geq 0)$ then

$$
C=p-1 .
$$

else

$$
C=p .
$$

end if
else
$\mathrm{C}=\mathrm{p}+1$.
end if
end if
where the columns of rectangular matrix are numbered $0,1,2, \ldots, C-1$ from left to right.
(3) Write the input bit sequence $x_{k}$ into the $R \times C$ rectangular matrix row by row starting with bit $x_{1}$ in column 0 of: row 0 :

$$
\left[\begin{array}{cccc}
x_{1} & x_{2} & x_{3} & \ldots x_{C} \\
x_{(C+1)} & x_{(C+2)} & x_{(C+3)} & \cdots x_{2 C} \\
\vdots & \vdots & \vdots & \cdots \\
x_{((R-1) C+1)} & x_{((R-1) C+2)} & x_{((R-1) C+3)} & \cdots x_{R C}
\end{array}\right] .
$$

### 4.2.3.2.3.2

 Intra-row and inter-row permutationsAfter the bits-input to the $R \times C$ rectangular matrix, the intra-row and inter-row permutations for the $R \underline{\times} \underline{C}$ rectangular matrix are performed by using the following algorithm:
(1) Select a primitive root $v$ from table 2 .
(2) Construct the base sequence $s(i)$ for intra-row permutation as: $s(i)=[v \times s(i-1)] \bmod p, \quad \mathrm{i}=1,2, \ldots(p-2) .$, and $s(0)=1$.
(3) Let $q_{0}=1$ be the first prime integer in $\left\{q_{j}\right\}$, and select the consecutive minimum prime integers $\left\{q_{\mathrm{j}}\right\}(\mathrm{j}=1,2, \ldots \mathrm{R}-1)$ : such that
g.c. $\left\{q_{\mathrm{j}}, p-1\right\}=1$,
$q_{j}>6$, and

$$
q_{j}>q_{(j-1)},
$$

where g.c.d. is greatest common divisor.
(4) Permute $\left\{q_{j}\right\}$ to make $\left\{r_{j}\right\}$ such that:

$$
r_{T(j)}=q_{j}, j=0,1, \ldots . \mathrm{R}-1,
$$

| where $T(j)$ indicates the original row position of the $j$ th permmted row, and $T(j)(j=0,1,2, \ldots, R-1)$ is the inter-row permutation pattern defined as the one of the following four kind of patterns: $\mathrm{Pat}_{1}, \mathrm{Pat}_{2}, \mathrm{Pat}_{3}$ and $\mathrm{Pat}_{4}$ depending on the number of input bits $K$.

$$
T(j) \underline{\{T(0), T(1), T(2), \ldots, T(R-1)\}}=\left\{\begin{array}{ll}
P_{4} t_{4} & \text { if }(40 \leq K \leq 159) \\
P_{3} & \text { if }(160 \leq K \leq 200) \\
P_{1} & \text { if }(201 \leq K \leq 480) \\
P a t_{3} & \text { if }(481 \leq K \leq 530) \\
P a t_{1} & \text { if }(531 \leq K \leq 2280) \\
P_{2} & \text { if }(2281 \leq K \leq 2480) \\
P_{1} & \text { if }(2481 \leq K \leq 3160) \\
P_{2} & \text { if }(3161 \leq K \leq 3210) \\
\text { Pat }_{1} & \text { if }(3211 \leq K \leq 5114)
\end{array},\right.
$$

where $\mathrm{Pat}_{1}, \mathrm{Pat}_{2}, \mathrm{Pat}_{3}$ and $\mathrm{Pat}_{4}$ have the following patterns respectively.
$P_{t}:\{19,9,14,4,0,2,5,7,12,18,10,8,13,17,3,1,16,6,15,11\}$
Pat $:\{19,9,14,4,0,2,5,7,12,18,16,13,17,15,3,1,6,11,8,10\}$
$P a t_{3}:\{9,8,7,6,5,4,3,2,1,0\}$
Pat $:\{4,3,2,1,0\}$
(5) Perform the $j$-th $(j=0,1,2, \ldots, R-1)$ intra-row permutation as:
if $(C=p)$ then

$$
U_{j}(i)=s\left(\left[i \times r_{j}\right] \bmod (p-1)\right), i=0,1,2, \ldots,(p-2) ., \text { and } U_{j}(p-1)=0,
$$

where $U_{j}(i)$ is the input bit position of $i$-th output after the permutation of $j$-th row.
end if
if $(\mathrm{C}=p+1)$ then
$U_{j}(i)=s\left(\left[i \times r_{j}\right] \bmod (p-1)\right), \quad i=0,1,2, \ldots,(p-2) ., \quad U_{j}(p-1)=0$, and $U_{j}(p)=p$,
where $U_{j}(i)$ is the input bit position of $i$-th output after the permutation of $j$-th row, and
if $(K=C \times R)$ then
Exhange $U_{R-I}(p)$ with $U_{R-I}(0)$.
end if
end if
if ( $C=p-1$ ) then

$$
U_{j}(i)=s\left(\left[i \times r_{j}\right] \bmod (p-1)\right)-1, i=0,1,2, \ldots,(p-2),
$$

where $U_{j}(i)$ is the input bit position of $i$-th output after the permutation of $j$-th row.
end if
(6) Perform the inter-row permutation based on the pattern $T(j)(j=0,1,2, \ldots, R-1)$,

Table 2: Table of prime $p$ and associated primitive root $v$

| $\mathbf{p}$ | $\mathbf{v}$ | $\mathbf{p}$ | $\mathbf{v}$ | $\mathbf{p}$ | $\mathbf{v}$ | $\mathbf{p}$ | $\mathbf{v}$ | $\mathbf{p}$ | $\mathbf{v}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | 3 | 47 | 5 | 101 | 2 | 157 | 5 | 223 | 3 |
| 11 | 2 | 53 | 2 | 103 | 5 | 163 | 2 | 227 | 2 |
| 13 | 2 | 59 | 2 | 107 | 2 | 167 | 5 | 229 | 6 |
| 17 | 3 | 61 | 2 | 109 | 6 | 173 | 2 | 233 | 3 |
| 19 | 2 | 67 | 2 | 113 | 3 | 179 | 2 | 239 | 7 |
| 23 | 5 | 71 | 7 | 127 | 3 | 181 | 2 | 241 | 7 |
| 29 | 2 | 73 | 5 | 131 | 2 | 191 | 19 | 251 | 6 |
| 31 | 3 | 79 | 3 | 137 | 3 | 193 | 5 | 257 | 3 |
| 37 | 2 | 83 | 2 | 139 | 2 | 197 | 2 |  |  |
| 41 | 6 | 89 | 3 | 149 | 2 | 199 | 3 |  |  |
| 43 | 3 | 97 | 5 | 151 | 6 | 211 | 2 |  |  |

### 4.2.3.2.3.3 Bits-output from rectangular matrix with pruning

After intra-row and inter-row permutations, the bits of the permuted rectangular matrix are denoted by $y_{k}^{\prime}$ :

$$
\left[\begin{array}{ccclc}
y_{1}^{\prime} & y_{(R+1)}^{\prime} & y_{(2 R+1)}^{\prime} & \cdots y_{((C-1) R+1)}^{\prime} \\
y_{2}^{\prime} & y_{(R+2)}^{\prime} & y_{(2 R+2)}^{\prime} & \cdots y_{((C-1) R+2)}^{\prime} \\
\vdots & \vdots & \vdots & \cdots & \vdots \\
y_{R}^{\prime} & y_{2 R}^{\prime} & y_{3 R}^{\prime} & \cdots & y_{C R}^{\prime}
\end{array}\right] .
$$

The output of the Turbo code internal interleaver is the bit sequence read out column by column from the intra-row and inter-row permuted $R \times C$ matrix starting with bit $y_{1}^{\prime}$ in row 0 of column 0 and ending with bit $y_{C R}^{\prime}$ in row $R-1$ of column $C-1$. The output is pruned by deleting bits that were not present in the input bit sequence, i.e. bits $y_{k}^{\prime}$ that corresponds to bits $x_{k}$ with $k>K$ are removed from the output. The bits output from Turbo code internal interleaver are denoted by $x_{1}^{\prime}, x_{2}^{\prime}$, $\ldots, x_{K}^{\prime}$, where $x_{1}^{\prime}$ corresponds to the bit $y_{k}^{\prime}$ with smallest index $k$ after pruning, $x_{2}^{\prime}$ to the bit $y_{k}^{\prime}$ with second smallest index $k$ after pruning, and so on. The number of bits output from Turbo code internal interleaver is $K$ and the total number of pruned bits is:
$R \times C-K$.

### 4.2.3.3 Concatenation of encoded blocks

After the channel coding for each code block, if $C_{i}$ is greater than 1, the encoded blocks are serially concatenated so that the block with lowest index $r$ is output first from the channel coding block, otherwise the encoded block is output from channel coding block as it is. The bits output are denoted by $c_{i 1}, c_{i 2}, c_{i 3}, \ldots, c_{i E_{i}}$, where $i$ is the $\operatorname{TrCH}$ number and $E_{i}=C_{i} Y_{i}$. The output bits are defined by the following relations:

$$
\begin{array}{ll}
c_{i k}=y_{i 1 k} & k=1,2, \ldots, Y_{i} \\
c_{i k}=y_{i, 2,\left(k-Y_{i}\right)} & k=Y_{i}+1, Y_{i}+2, \ldots, 2 Y_{i} \\
c_{i k}=y_{i, 3,\left(k-2 Y_{i}\right)} & k=2 Y_{i}+1,2 Y_{i}+2, \ldots, 3 Y_{i} \\
\ldots & \\
c_{i k}=y_{i, C_{i},\left(k-\left(C_{i}-1\right) Y_{i}\right)} \quad k=\left(C_{i}-1\right) Y_{i}+1,\left(C_{i}-1\right) Y_{i}+2, \ldots, C_{i} Y_{t} \underline{Y_{i}}
\end{array}
$$

If no code blocks are input to the channel coding $\left(C_{i}=0\right)$, no bits shall be output from the channel coding, i.e. $E_{i}=0$.

