TSG-RAN Meeting #8 Düsseldorf, Germany, 21-23 June 2000

Title: Agreed CRs to TS 25.212

Source: TSG-RAN WG1

Agenda item: 5.1.3

No.	Doc #	Spec	CR	Rev	Subject	Cat	Current_v	New_v
1	R1-000585	25.212	066	1	Section 4.4.5 and table 9 is moved to informative	F	3.2.0	3.3.0
2	R1-000539	25.212	068	-	Editorial modifications of 25.212	D	3.2.0	3.3.0
3	R1-000541	25.212	069	-	Removal of BTFD for flexible positions in Release	F	3.2.0	3.3.0
4	R1-000560	25.212	070	1	Editorial modifications	D	3.2.0	3.3.0
5	R1-000572	25.212	071	1	Corrections and editorial modifications of 25.212	F	3.2.0	3.3.0
6	R1-000735	25.212	072	4	Corrections to 25.212 (Rate Matching, p-bit	F	3.2.0	3.3.0
7	R1-000634	25.212	073	-	Editorial correction in 25.212 coding/multiplexing	F	3.2.0	3.3.0
8	R1-000716	25.212	074	2	Bit separation of the Turbo encoded data	D	3.2.0	3.3.0
9	R1-000723	25.212	076	1	Revision of code block segmentation description	D	3.2.0	3.3.0
10	R1-000657	25.212	077	-	Clarifications for TFCI coding	F	3.2.0	3.3.0
11	R1-000775	25.212	078	2	Clarifying the rate matching parameter setting for	F	3.2.0	3.3.0
12	R1-000706	25.212	080	-	Clarification on BTFD utilisation (single CCTrCH)	F	3.2.0	3.3.0
13	R1-000741	25.212	081	-	Correction of order of checking TFC during	F	3.2.0	3.3.0
14	R1-000743	25.212	082	-	Editorial corrections in channel coding section	F	3.2.0	3.3.0
15	R1-000744	25.212	083	-	Correction for bit separation and bit collection	F	3.2.0	3.3.0
16	R1-000776	25.212	084	1	Correction on the spreading factor selection for	F	3.2.0	3.3.0

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e.g. for 3GPP use the format TP-99xxx or for SMG, use the format P-99-xxx

	CHANGE	REQUEST Please	ase see embedded help f e for instructions on how	file at the bottom of this to fill in this form correctly.
	25.212	CR 066r1	Current Versio	on: 3.2.0
GSM (AA.BB) or 3G	(AA.BBB) specification number ↑	↑ CR numb	er as allocated by MCC s	support team
For submission to: TSG RAN #8 for approval X strategic list expected approval meeting # here for information non-strategic				
For	m: CR cover sheet, version 2 for 3GPP and SMG	G The latest version of this form is a	vailable from: ftp://ftp.3gpp.o	prg/Information/CR-Form-v2.doc
Proposed chang (at least one should be m	e affects: (U)SIM	ME UTRA	N / Radio X	Core Network
Source:	TSG RAN WG1		Date:	12-Apr-00
Subject:	Section 4.4.5 and table 9 is	moved to informative ar	nnex.	
Work item:				
Category:FA(only one categorybshall be markedCwith an X)D	Correction Corresponds to a correction Addition of feature Functional modification of fe Editorial modification	n in an earlier release eature	X <u>Release:</u>	Phase 2Release 96Release 97Release 98Release 99XRelease 00
<u>Reason for</u> change:	Table 9 contains some ambi informative, thus it is moved	iguities which are correct to an annex.	cted. Most of the t	table is
Clauses affected	Letter 4.4.5 and Annex B			
Other specs affected:	Other 3G core specifications Other GSM core specifications MS test specifications BSS test specifications O&M specifications	X \rightarrow List of CRs:	25.215-057	
<u>Other</u> <u>comments:</u>				

4.4.5 Parameters for downlink compressed mode

Table 9 shows the detailed parameters for each transmission gap length for the different transmission time reduction methods.

TGL	Frame Type	Spreading Factor	Idle length [ms]	Transmission time Reduction method	Idle frame Combining
3	A	512 – 4	1.73-1.99	Puncturing,	(S)
	₽	256 - 4	1.60-1.86	Spreading factor	(D) =(1,2) or (2,1)
4	A	512 - 4	2.40-2.66	division by 2 or	(S)
	₿	256 - 4	2.27-2.53	Higher layer	(D) =(1,3), (2,2) or (3,1)
5	A	512 - 4	3.07-3.33	scheduling	(S)
	₿	256 - 4	2.94-3.20		(D) = (1,4), (2,3), (3,2) or
7	A	512 - 4	4.40-4.66		(1,1) (S)
	₿	256 - 4	4 .27-4.53]	(D)=(1,6), (2,5), (3,4), (4,3),
					(5,2) or (6,1)
10	A	512 - 4	6.40-6.66		(D)=(3,7), (4,6), (5,5), (6,4) or
	₽	256 - 4	6.27-6.53		(7,3)
14	A	512 - 4	9.07-9.33		(D) = (7,7)
	₿	256 - 4	8.93-9.19		

Table 9: Parameters for compressed mode

(S): Single frame method as shown in figure 14 (1).

NOTE: Compressed mode by spreading factor reduction is not supported when SF=4 is used in normal mode.

<u>Annex B (informative):</u> Compressed mode idle lengths

The tables 9-11 show the resulting idle lengths for different transmission gap lengths, UL/DL modes and DL frame types. The idle lengths given are calculated purely from the slot and frame structures and the UL/DL offset. They do not contain margins for e.g. synthesizer switching.

B.1 Idle lengths for DL, UL and DL+UL compressed mode

<u>TGL</u>	<u>DL</u> <u>Frame</u> Type	Spreading Factor	Idle length [ms]	Transmission time Reduction method	Idle frame Combining
<u>3</u>	A B	<u>512 – 4</u>	<u>1.73 – 1.99</u> <u>1.60 – 1.86</u>	Puncturing,	<u>(D) =(1,2) or (2,1)</u>
<u>4</u>	<u>A</u> B		<u>2.40 - 2.66</u> <u>2.27 - 2.53</u>	Spreading factor division by 2 or	<u>(D) =(1,3), (2,2) or (3,1)</u>
<u>5</u>	<u>A</u> B		<u>3.07 - 3.33</u> <u>2.93 - 3.19</u>	<u>Higher layer</u> <u>scheduling</u>	$\frac{(S)}{(D) = (1,4), (2,3), (3,2) \text{ or}}$ $\frac{(4,1)}{(4,1)}$
<u>7</u>	<u>A</u> <u>B</u>		$\frac{4.40 - 4.66}{4.27 - 4.53}$		(<u>S)</u> (<u>D)=(1,6), (2,5), (3,4), (4,3),</u> (5,2) or (6,1)
<u>10</u>	A B		<u>6.40 - 6.66</u> <u>6.27 - 6.53</u>		(D)=(3,7), (4,6), (5,5), (6,4) or (7,3)
<u>14</u>	<u>A</u> B		<u>9.07 - 9.33</u> <u>8.93 - 9.19</u>		<u>(D) =(7,7)</u>

Table 9: Parameters for DL compressed mode

Table 10: Parameters for UL compressed mode

TGL	Spreading Factor	Idle length [ms]	Transmission time Reduction method	Idle frame Combining
<u>3</u>	<u>256 – 4</u>	2.00	Spreading factor	$\frac{(S)}{(D) = (1,2) \text{ or } (2,1)}$
<u>4</u>		<u>2.67</u>	division by 2 or Higher layer	<u>(D) =(1,3), (2,2) or (3,1)</u>
<u>5</u>		<u>3.33</u>	<u>scheduling</u>	$\frac{(S)}{(D) = (1,4), (2,3), (3,2) \text{ or}}$
<u>7</u>		<u>4.67</u>		(D) = (1,6), (2,5), (3,4), (4,3), (5,2) or (6,1)
<u>10</u>		<u>6.67</u>		$\frac{(0,2) \text{ or } (0,1)}{(D)=(3,7), (4,6), (5,5), (6,4) \text{ or}}$ (7,3)
<u>14</u>		<u>9.33</u>		<u>(D) =(7,7)</u>

<u>TGL</u>	<u>DL</u> <u>Frame</u> <u>Type</u>	Spreading Factor	Idle length [ms]	Transmission time Reduction method	Idle frame Combining
<u>3</u>	A or B	DL:	<u>1.47 – 1.73</u>	DL:	<u>(D) =(1,2) or (2,1)</u>
<u>4</u>		<u>512 – 4</u>	<u>2.13 – 2.39</u>	Puncturing, Spreading factor	(D) =(1,3), (2,2) or (3,1)
<u>5</u>		<u>UL:</u> 256 – 4	<u>2.80 - 3.06</u>	division by 2 or <u>Higher layer</u> <u>scheduling</u>	$\frac{(S)}{(D) = (1,4), (2,3), (3,2) \text{ or}}$ $\frac{(4,1)}{(4,1)}$
<u>7</u>			<u>4.13 – 4.39</u>	UL: Spreading factor	(<u>S)</u> (<u>D)=(1,6), (2,5), (3,4), (4,3),</u> (5,2) or (6,1)
<u>10</u>			<u>6.13 – 6.39</u>	division by 2 or <u>Higher layer</u>	$\frac{(D)=(3,7), (4,6), (5,5), (6,4) \text{ or}}{(7,3)}$
<u>14</u>			8.80 - 9.06	scheduling	(D) = (7,7)

Table 11: Parameters for combined UL/DL compressed mode

(S): Single-frame method as shown in figure 14 (1).

(D): Double-frame method as shown in figure 14 (2). (x,y) indicates x: the number of idle slots in the first frame, y: the number of idle slots in the second frame.

NOTE: Compressed mode by spreading factor reduction is not supported when SF=4 is used in normal mode

Annex <u>B-C</u> (informative): Change history

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	25.212 CR 068 Current Version: 3.2.0				
GSM (AA.BB) or 3G (AA.BBB) specification number ↑ ↑ CR number as allocated by MCC support team					
For submission to: TSG RAN#8 for approval X strategic (for SI list expected approval meeting # here for information for information non-strategic use or					
101					
Proposed change (at least one should be m	e affects: (U)SIM ME X UTRAN / Radio X Core Network arked with an X)				
<u>Source:</u>	TSG RAN WG1 Date:				
Subject:	Editorial modifications of 25.212				
Work item:	TS 25.212				
Category:FA(only one categoryshall be markedCwith an X)D	CorrectionRelease:Phase 2Corresponds to a correction in an earlier releaseRelease 96Addition of featureRelease 97Functional modification of featureRelease 98Editorial modificationX				
<u>Reason for</u> <u>change:</u>	Editorial modifications correcting small mistakes left in change request 25.212-042-r5 (Downlink compressed mode by puncturing)				
Clauses affected	<u>4.2, 4.2.5.3, 4.2.5.4, 4.2.7.2, 4.2.9.1, 4.2.10</u>				
Other specs	Other 3G core specifications \rightarrow List of CRs:Other GSM core specifications \rightarrow List of CRs:VS test specifications \rightarrow List of CRs:OSS test specifications \rightarrow List of CRs:O&M specifications \rightarrow List of CRs:				
<u>Other</u> comments:					

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e.g. for 3GPP use the format TP-99xxx or for SMG, use the format P-99-xxx

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<----- double-click here for help and instructions on how to create a CR.

4.2 Transport-channel coding/multiplexing

Data arrives to the coding/multiplexing unit in form of transport block sets once every transmission time interval. The transmission time interval is transport-channel specific from the set {10 ms, 20 ms, 40 ms, 80 ms}.

The following coding/multiplexing steps can be identified:

- add CRC to each transport block (see subclause 4.2.1);
- transport block concatenation and code block segmentation (see subclause 4.2.2);
- channel coding (see subclause 4.2.3);
- rate matching (see subclause 4.2.7);
- insertion of discontinuous transmission (DTX) indication bits (see subclause 4.2.9);
- interleaving (two steps, see subclauses 4.2.4 and 4.2.11);
- radio frame segmentation (see subclause 4.2.6);
- multiplexing of transport channels (see subclause 4.2.8);
- physical channel segmentation (see subclause 4.2.10);
- mapping to physical channels (see subclause 4.2.12).

The coding/multiplexing steps for uplink and downlink are shown in figure 1 and figure 2 respectively.



Figure 1: Transport channel multiplexing structure for uplink





Figure 2: Transport channel multiplexing structure for downlink

The single output data stream from the TrCH multiplexing, including DTX indication bits in downlink, is denoted *Coded Composite Transport Channel (CCTrCH)*. A CCTrCH can be mapped to one or several physical channels.

4.2.5 1st interleaving

In Compressed Mode by puncturing, bits marked with a fourth value on top of $\{0, 1, \delta\}$ and noted p, are introduced in the radio frames to be compressed, in positions corresponding to the first bits of the radio frames. They will be removed in a later stage of the multiplexing chain to create the actual gap. Additional puncturing has been performed in the rate matching step, over the TTI containing the compressed radio frame, to create room for these p-bits. The following subclause describes this feature.

4.2.5.1 Insertion of marked bits in the sequence to be input in first interleaver

In normal mode, compressed mode by higher layer scheduling, and compressed mode by spreading factor reduction:

 $x_{ik} \equiv Z_{ik}$ and $X_i = Z_i$.

In case of compressed mode by puncturing and fixed positions, sequence $x_{i,k}$ which will be input to first interleaver for TrCh *i* and TTI *m* within largest TTI, is built from bits $z_{i,k}$, $k=1, ...Z_i$, plus $Np^{TTI, m}_{i,max}$ bits marked p and $X_i = Z_i + Np^{TTI, m}_{i,max}$, as is described thereafter.

 $Np^{TTI, m}_{i,max}$ is defined in the Rate Matching subclause 4.2.7.

 $P_{Fi}[x]$ defines the inter column permutation function for a TTI of length Fi *10ms, as defined in Table 3 above. $P_{Fi}[x]$ is the Bit Reversal function of x on $log_2(Fi)$ bits.

- NOTE 1: C[x], x=0 to Fi 1, the number of bits p which have to be inserted in each of the Fi segments of the TTI, i.e. in each column of the first interleaver. C[x] is equal to $Np^{x}_{i,max}$ for x equal 0 to Fi -1 for fixed positions. It is noted Np^{x}_{i} in the following initialisation step.
- NOTE 2: cbi[x], x=0 to Fi 1, the counter of the number of bits p inserted in each of the Fi segments of the TTI, i.e. in each column of the first interleaver.

col = 0

while $\operatorname{col} < F_i$ do

$C[col] = Np_{i}^{col}$	initialisation of number of bits p to be inserted in each of the Fi segments of the TTI

cbi[col] = 0 -- initialisation of counter of number of bits p inserted in each of the Fi segments of the TTI

end do

n = 0, m = 0

while $n < X_i$ do

```
col = n \mod F_i
```

if $cbi[col] < C[P_{Fi}(col)]$ **do**

 $\mathbf{x}_{i,n} = \mathbf{p}$

-- insert one p bit

cbi[col] = cbi[col]+1

-- update counter of number of bits p inserted

```
else
```

-- no more p bit to insert in this segment

 $x_{i,n} = z_{i,m}$

```
m = m+1
```

endif

n = n + 1

end do

4.2.5.2 1st interleaver operation

The 1st interleaving is a block interleaver with inter-column permutations. The input bit sequence to the 1st interleaver is denoted by $x_{i1}, x_{i2}, x_{i3}, \ldots, x_{iX_i}$, where *i* is TrCH number and X_i the number of bits (at this stage X_i is assumed and guaranteed to be an integer multiple of TTI). The output bit sequence is derived as follows:

- (1) Select the number of columns C_I from table 3.
- (2) Determine the number of rows R_I defined as:

 $R_{\rm I} = X_{\rm i}/C_{\rm I}$

(3) Write the input bit sequence into the $R_I \times C_I$ rectangular matrix row by row starting with bit $x_{i,1}$ in the first column of the first row and ending with bit $x_{i,(R_IC_I)}$ in column C_I of row R_I :

$$\begin{bmatrix} X_{i1} & X_{i2} & X_{i3} & \dots & X_{iC_{I}} \\ X_{i,(C_{I}+1)} & X_{i,(C_{I}+2)} & X_{i,(C_{I}+3)} & \dots & X_{i,(2C_{I})} \\ \vdots & \vdots & \vdots & & \vdots \\ X_{i,((R_{I}-1)C_{I}+1)} & X_{i,((R_{I}-1)C_{I}+2)} & X_{i,((R_{I}-1)C_{I}+3)} & \dots & X_{i,(R_{I}C_{I})} \end{bmatrix}$$

(4) Perform the inter-column permutation based on the pattern $\{P_1(j)\}$ (j=0,1, ..., C-1) shown in table 3, where $P_1(j)$ is the original column position of the *j*-th permuted column. After permutation of the columns, the bits are denoted by y_{ik} :

y_{i1}	$\mathcal{Y}_{i,(R_I+1)}$	$\mathcal{Y}_{i,(2R_I+1)}$	$\cdots \mathcal{Y}_{i,((C_I-1)R_I+1)}$
y_{i2}	$\mathcal{Y}_{i,(R_I+2)}$	$y_{i,(2R_I+2)}$	$\cdots \mathcal{Y}_{i,((C_I-1)R_I+2)}$
:	÷	÷	:
y_{iR_I}	$\mathcal{Y}_{i,(2R_I)}$	$\mathcal{Y}_{i,(3R_I)}$	$\dots \mathcal{Y}_{i,(C_I R_I)}$

(5) Read the output bit sequence $y_{i1}, y_{i2}, y_{i3}, \dots, y_{i,(C_lR_l)}$ of the 1st interleaving column by column from the intercolumn permuted $R_l \times C_l$ matrix. Bit $y_{i,1}$ corresponds to the first row of the first column and bit $y_{i,(R_lC_l)}$ corresponds to row R_l of column C_l .

TTI	Number of columns C ₁	Inter-column permutation patterns
10 ms	1	{0}
20 ms	2	{0,1}
40 ms	4	{0,2,1,3}
80 ms	8	{0,4,2,6,1,5,3,7}

Table 3

4.2.5.3 Relation between input and output of 1st interleaving in uplink

The bits input to the 1st interleaving are denoted by $t_{i1}, t_{i2}, t_{i3}, \dots, t_{iT_i}$, where *i* is the TrCH number and T_i the number of bits. Hence, $z_{ik} = t_{ik}$ and $Z_{i} = T_i$.

The bits output from the 1st interleaving are denoted by $d_{i1}, d_{i2}, d_{i3}, \dots, d_{iT_i}$, and $d_{ik} = y_{ik}$.

4.2.5.4 Relation between input and output of 1st interleaving in downlink

If fixed positions of the TrCHs in a radio frame is used then the bits input to the 1st interleaving are denoted by $h_{i1}, h_{i2}, h_{i3}, \dots, h_{i(F_iH_i)}$, where *i* is the TrCH number. Hence, $z_{ik} = h_{ik}$ and $Z_i = \underline{D}_i F_i * H_i - Np^{\frac{TT_i}{m}}$ in compressed mode by puncturing, and $Z_i = F_i H_i$ otherwise.

If flexible positions of the TrCHs in a radio frame is used then the bits input to the 1st interleaving are denoted by $g_{i1}, g_{i2}, g_{i3}, \dots, g_{iG_i}$, where *i* is the TrCH number. Hence, $z_{ik} = g_{ik}$ and $Z_i = G_i$.

The bits output from the 1st interleaving are denoted by $q_{i1}, q_{i2}, q_{i3}, \dots, q_{iQ_i}$, where *i* is the TrCH number and Q_i is the number of bits. Hence, $q_{ik} = y_{ik}, Q_i = F_i H_i$ if fixed positions are used, and $Q_i = G_i$ if flexible positions are used.

4.2.7.2 Determination of rate matching parameters in downlink

For downlink $N_{data,j}$ does not depend on the transport format combination *j*. $N_{data,*}$ is given by the channelization code(s) assigned by higher layers. Denote the number of physical channels used for the CCTrCH by *P*. $N_{data,*}$ is the number of bits available to the CCTrCH in one radio frame and defined as $N_{data,*}=P(15N_{data1}+15N_{data2})$, where N_{data1} and N_{data2} are defined in [2]. Note that contrary to the uplink, the same rate matching patterns are used in normal and compressed mode by spreading factor reduction or higher layer scheduling.

In the following, the total amount of puncturing or repetition for the TTI is calculated.

Additional calculations for compressed mode by puncturing in case of fixed positions are performed to determine this total amount of rate matching needed.

For compressed mode by puncturing, in TTIs where some compressed radio frames occur, the puncturing is increased or the repetition is decreased compared to what is calculated according to the rate matching parameters provided by higher layers. This allows to create room for later insertion of marked bits, noted p-bits, which will identify the positions of the gaps in the compressed radio frames.

The amount of additional puncturing corresponds to the number of bits to create the gap in the TTI for TrCh_i. In case of fixed positions, it is calculated in addition to the amount of rate matching indicated by higher layers. It is noted $Np^{TTI, m}_{i,max}$.

In fixed positions case, to obtain the total rate matching $\Delta N_{i,\max}^{TTI,cm,m}$ to be performed on the TTI *m*, $Np^{TTI,m}_{i,\max}$ is sub-stracted from $\Delta N^{TTI,m}_{i,\max}$ (calculated based on higher layers RM parameters as for normal rate matching). This allows to create room for the $Np^{TTI,m}_{i,\max}$ bits p to be inserted later. If the result is null, i.e. the amount of repetition matches exactly the amount of additional puncturing needed, then no rate matching is necessary.

In case of compressed mode by puncturing and fixed positions, for some calculations, $N'_{data,*}$ is used for radio frames with gap instead of $N_{data,*}$, where $N'_{data,*} = P(15N'_{data1} + 15N'_{data2})$. N'_{data1} and N'_{data2} are the number of bits in the data fields of the slot format used for the current compressed mode, i.e. slot format A or B as defined in [2] corresponding to the Spreading Factor and the number of transmitted slots in use.

The number of bits corresponding to the gap for TrCh i, in each radio frame of its TTI is calculated using the number of bits to remove on <u>each-all</u> Physical Channels $N_{TGL}[k]$, where k is the radio frame number in the TTI.

For each radio frame k of the TTI, N_{TGL}[k] is given by the relation:

$$N_{TGL} = \begin{cases} \frac{TGL}{15} N'_{data,*}, \text{ if } N_{first} + TGL \le 15 \\ \frac{15 - N_{first}}{15} N'_{data,*}, \text{ in first radio frame of the gap if } N_{first} + TGL > 15 \\ \frac{TGL - (15 - N_{first})}{15} N'_{data,*}, \text{ in second radio frame of the gap if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

Note that N $_{TGL}[k] = 0$ if radio frame k is not compressed.

4.2.9.1 1st insertion of DTX indication bits

This step of inserting DTX indication bits is used only if the positions of the TrCHs in the radio frame are fixed. With fixed position scheme a fixed number of bits is reserved for each TrCH in the radio frame.

The bits from rate matching are denoted by $g_{i1}, g_{i2}, g_{i3}, \dots, g_{iG_i}$, where G_i is the number of bits in one TTI of TrCH *i*. Denote the number of bits in one radio frame of TrCH *i* by H_i . Denote D_i the number of bits output of the first DTX insertion block.

In normal or compressed mode by spreading factor reduction, H_i is constant and corresponds to the maximum number of bits from TrCH *i* in one radio frame for any transport format of TrCH *i*. and $D_i = F_i * H_i$.

In compressed mode by puncturing, additional puncturing is performed in the rate matching block. The empty positions resulting from the additional puncturing are used to insert p-bits in the first interleaving block, the DTX insertion is therefore limited to allow for later insertion of p-bits. Thus DTX bits are inserted until the total number of bits is Di where $D_i = F_i * H_{i,*} - Np^{TTI, m}_{i,max} + \Delta N^{TTI}$ and $H_i = N_{i,*} + \Delta N_{i,*}$.

The bits output from the DTX insertion are denoted by h_{il} , h_{i2} , h_{i3} , ..., h_{iDi} Note that these bits are three valued. They are defined by the following relations:

$$h_{ik} = g_{ik} \ k = 1, 2, 3, ..., G_i$$

$$h_{ik} = \delta$$
 $k = G_i + 1, G_i + 2, G_i + 3, ..., D_i$

where DTX indication bits are denoted by δ . Here $g_{ik} \in \{0, 1\}$ and $\delta \notin \{0, 1\}$.

4.2.10 Physical channel segmentation

When more than one PhCH is used, physical channel segmentation divides the bits among the different PhCHs. The bits input to the physical channel segmentation are denoted by $x_1, x_2, x_3, \ldots, x_Y$, where *Y* is the number of bits input to the physical channel segmentation block. The number of PhCHs is denoted by *P*.

The bits after physical channel segmentation are denoted $u_{p1}, u_{p2}, u_{p3}, \dots, u_{pU}$, where *p* is PhCH number and *U* is the number of bits in one radio frame for each PhCH, i.e. $U = (Y - N_{TGL})/P$ for compressed mode by puncturing, and

 $U = \frac{Y}{P}$ otherwise. The relation between x_k and u_{pk} is given below.

For all modes, some bits of the input flow are mapped to each code until the number of bits on the code is 4U. For modes other than compressed mode by puncturing, all bits of the input flow are taken to be mapped to the codes. For compressed mode by puncturing, only the bits of the input flow not corresponding to bits p are taken to be mapped to the codes, each bit p is removed to ensure creation the gap required by the compressed mode, as described below.

Bits on first PhCH after physical channel segmentation:

 $u_{1, k} = x_{i, f(k)} \ k = 1, 2, ..., U$

Bits on second PhCH after physical channel segmentation:

 $u_{2,k} = x_{i,f(k+U)} k = 1, 2, ..., U$

...

Bits on the P^{th} PhCH after physical channel segmentation:

 $u_{P, k} = x_{i, f(k+(P-1) U)} k = 1, 2, ..., U$

where f is such that :

- for modes other than compressed mode by puncturing, $x_{i, f(k)} = x_{i, k}$, i.e. f(k) = k, for all k.
- for compressed mode by puncturing, bit u_{1,1} corresponds to the bit x_{i,k} with smallest index k when the bits p are not counted, bit u_{1,2} corresponds to the bit x_{i,k} with second smallest index k when the bits p are not counted, and so on for bits u_{1,3} ... u_{1,4U}, u_{2,2}, ..., u_{2,4U}, ..., u_{P,1}, u_{P,2}, ..., u_{P,4U},

4.2.10.1 Relation between input and output of the physical segmentation block in uplink

The bits input to the physical segmentation are denoted by $s_1, s_2, s_3, \dots, s_s$. Hence, $x_k = s_k$ and Y = S.

4.2.10.2 Relation between input and output of the physical segmentation block in downlink

The bits input to the physical segmentation are denoted by $W_1, W_2, W_3, \dots, W_{(PU)}$. Hence, $x_k = W_k$ and Y = PU.

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4.2.12 Physical channel mapping

The PhCH for both uplink and downlink is defined in [2]. The bits input to the physical channel mapping are denoted by $v_{p1}, v_{p2}, \dots, v_{pU}$, where *p* is the PhCH number and *U* is the number of bits in one radio frame for one PhCH. The bits v_{pk} are mapped to the PhCHs so that the bits for each PhCH are transmitted over the air in ascending order with respect to *k*.

In compressed mode, no bits are mapped to certain slots of the PhCH(s). If $N_{first} + TGL \le 15$, no bits are mapped to slots N_{first} to N_{last} . If $N_{first} + TGL > 15$, i.e. the transmission gap spans two consecutive radio frames, the mapping is as follows:

- In the first radio frame, no bits are mapped to slots N_{first} , $N_{first}+1$, $N_{first}+2$, ..., 14.
- In the second radio frame, no bits are mapped to the slots $0, 1, 2, ..., N_{last}$.

TGL, N_{first} , and N_{last} are defined in section 4.4.

4.2.12.1 Uplink

In uplink, the PhCHs used during a radio frame are either completely filled with bits that are transmitted over the air or not used at all. The only exception is when the UE is in compressed mode. The transmission can then be turned off during consecutive slots of the radio frame.

4.2.12.2 Downlink

In downlink, the PhCHs do not need to be completely filled with bits that are transmitted over the air. Bits $v_{pk} \notin \{0, 1\}$ are not transmitted.

During compressed mode by reducing the spreading factor by 2, no bits are mapped to the DPDCH field as follows:

If $N_{first} + TGL \le 15$, i.e. the transmission gap spans one radio frame,

if
$$N_{first} + 7 \le 14$$

no bits are mapped to slots N_{first} , N_{first} + 1, N_{first} + 2, ..., N_{last} + (7 - TGL)

no bits are mapped to the first $(N_{Data1} + N_{Data2})/2$ bit positions of slot $N_{last} + (8 - TGL)$

else

no bits are mapped to slots N_{first} , N_{first} + 1, N_{first} + 2,..., 14

no bits are mapped to slots N_{first} - 1, N_{first} - 2, N_{first} - 3, ..., N_{first} - (7 - TGL - (14 - N_{last}))

no bits are mapped to the last (N_{Data1}+ N_{Data2})/2 bit positions of slot N_{first} - (8 - TGL - (14 - N_{last}))

end if

If $N_{first} + TGL > 15$, i.e. the transmission gap spans two consecutive radio frames,

In the first radio frame, no bits are mapped to last $(N_{Data1} + N_{Data2})/2$ bit positions in slot 7 as well as to slots 8, 9, 10, ..., 14.

In the second radio frame, no bits are mapped to slots 0, 1, 2, ..., 6 as well as to first $(N_{Data1} + N_{Data2})/2$ bit positions in slot 7.

N_{Data1}and N_{Data2} are defined in [2].

The following rules should be used for the selection of fixed or flexible positions of the TrCHs in the radio frame:

For TrCHs not relying on TFCI for transport format detection (blind transport format detection), the positions of the transport channels within the radio frame should be fixed. In a limited number of cases, where there are a small number of transport format combinations, it is possible to allow flexible positions.

- For TrCHs relying on TFCI for transport format detection, higher layer signal whether the positions of the transport channels should be fixed or flexible.

A.2 Blind transport format detection with flexible positions

In certain cases where the CCTrCH consists of multiple transport channels and a small number of transport format combinations are allowed, it is possible to allow blind transport format detection with flexible positions.

Several examples for how the blind transport format detection with flexible positions might be performed are:

The blind transport format detection starts at a fixed position and identifies the transport format of the first present transport channel and stops. The position of the other transport channels and their transport format being derived on the basis of the allowed transport format combinations, assuming that there is a one to one relationship between the transport format combination and the transport format of the first present transport channel.

The blind rate detection evaluates all transport format combinations and picks the most reliable one.

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4.2.3.3 Concatenation of encoded blocks

After the channel coding for each code block, if C_i is greater than 1, the encoded blocks are serially concatenated so that the block with lowest index *r* is output first from the channel coding block, otherwise the encoded block is output from channel coding block as it is. The bits output are denoted by $c_{i1}, c_{i2}, c_{i3}, \ldots, c_{iE_i}$, where *i* is the TrCH number and $E_i = C_i Y_i$. The output bits are defined by the following relations:

 $c_{ik} = y_{i1k} \quad k = 1, 2, ..., Y_i$ $c_{ik} = y_{i,2,(k-Y_i)} \quad k = Y_i + 1, Y_i + 2, ..., 2Y_i$ $c_{ik} = y_{i,3,(k-2Y_i)} \quad k = 2Y_i + 1, 2Y_i + 2, ..., 3Y_i$... $c_{ik} = y_{i,C_i,(k-(C_i-1)Y_i)} \quad k = (C_i - 1)Y_i + 1, (C_i - 1)Y_i + 2, ..., C_iY_{ij}$

If no code blocks are input to the channel coding ($C_i = 0$), no bits shall be output from the channel coding, i.e. $E_i = 0$.

4.2.5.2 1st interleaver operation

The 1st interleaving is a block interleaver with inter-column permutations. The input bit sequence to the 1st interleaver is denoted by $x_{i1}, x_{i2}, x_{i3}, \ldots, x_{iX_i}$, where *i* is TrCH number and X_i the number of bits (at this stage X_i is assumed and guaranteed to be an integer multiple of TTI). The output bit sequence is derived as follows:

- (1) Select the number of columns C_I from table 3.
- (2) Determine the number of rows R_I defined as:

 $R_I = X_i/C_I$

(3) Write the input bit sequence into the $R_I \times C_I$ rectangular matrix row by row starting with bit $x_{i,1}$ in the first column of the first row and ending with bit $x_{i,(R_IC_I)}$ in column C_I of row R_I :

$$\begin{bmatrix} x_{i1} & x_{i2} & x_{i3} & \dots & x_{iC_I} \\ x_{i,(C_I+1)} & x_{i,(C_I+2)} & x_{i,(C_I+3)} & \dots & x_{i,(2C_I)} \\ \vdots & \vdots & \vdots & & \vdots \\ x_{i,((R_I-1)C_I+1)} & x_{i,((R_I-1)C_I+2)} & x_{i,((R_I-1)C_I+3)} & \dots & x_{i,(R_IC_I)} \end{bmatrix}$$

(4) Perform the inter-column permutation based on the pattern $\{P_1(j)\}$ (*j*=0,1, ..., C-1) shown in table 3, where $P_1(j)$ is the original column position of the *j*-th permuted column. After permutation of the columns, the bits are denoted by y_{ik} :

$$\begin{bmatrix} y_{i1} & y_{i,(R_{I}+1)} & y_{i,(2R_{I}+1)} & \cdots & y_{i,((C_{I}-1)R_{I}+1)} \\ y_{i2} & y_{i,(R_{I}+2)} & y_{i,(2R_{I}+2)} & \cdots & y_{i,((C_{I}-1)R_{I}+2)} \\ \vdots & \vdots & \vdots & & \vdots \\ y_{iR_{I}} & y_{i,(2R_{I})} & y_{i,(3R_{I})} & \cdots & y_{i,(C_{I}R_{I})} \end{bmatrix}$$

(5) Read the output bit sequence $y_{i1}, y_{i2}, y_{i3}, \dots, y_{i,(C_IR_I)}$ of the 1st interleaving column by column from the intercolumn permuted $R_I \times C_I$ matrix. Bit $y_{i,1}$ corresponds to the first row of the first column and bit $y_{i,(R_IC_I)}$ corresponds to row R_I of column C_I .

Table 3 Inter-column permutation patterns for 1st inter	leaving
---	---------

TTI	Number of columns C ₁	Inter-column permutation patterns
10 ms	1	{0}
20 ms	2	{0,1}
40 ms	4	{0,2,1,3}
80 ms	8	{0,4,2,6,1,5,3,7}

4.2.7 Rate matching

Rate matching means that bits on a transport channel are repeated or punctured. Higher layers assign a rate-matching attribute for each transport channel. This attribute is semi-static and can only be changed through higher layer signalling. The rate-matching attribute is used when the number of bits to be repeated or punctured is calculated.

The number of bits on a transport channel can vary between different transmission time intervals. In the downlink the transmission is interrupted if the number of bits is lower than maximum. When the number of bits between different transmission time intervals in uplink is changed, bits are repeated or punctured to ensure that the total bit rate after TrCH multiplexing is identical to the total channel bit rate of the allocated dedicated physical channels.

If no bits are input to the rate matching for all TrCHs within a CCTrCH, the rate matching shall output no bits for all TrCHs within the CCTrCH and no uplink DPDCH will be selected in the case of uplink rate matching.

Notation used in subcaluse 4.2.7 and subclauses:

 N_{ij} : For uplink: Number of bits in a radio frame before rate matching on TrCH *i* with transport format combination *j*.

For downlink: An intermediate calculation variable (not an integer but a multiple of 1/8).

- N_{il}^{TTI} : Number of bits in a transmission time interval before rate matching on TrCH *i* with transport format *l*. Used in downlink only.
- ΔN_{ij} : For uplink: If positive number of bits that should be repeated in each radio frame on TrCH *i* with transport format combination *j*.

If negative - number of bits that should be punctured in each radio frame on TrCH i with transport format combination j.

For downlink : An intermediate calculation variable (not an integer but a multiple of 1/8).

 ΔN_{il}^{TTI} : If positive - number of bits to be repeated in each transmission time interval on TrCH *i* with transport format *j*.

If negative - number of bits to be punctured in each transmission time interval on TrCH i with transport format j.

Used in downlink only.

 $Np^{TTI, m}_{i,b}$ m=0 to $F_{max}/F_i - 1$: Positive or null: number of bits to be removed in TTI number *m* within the largest TTI, to create the required gaps in the compressed radio frames of this TTI, in case of compressed mode by puncturing, for TrCh *i* with transport format *l*. In case of fixed positions and compressed mode by puncturing, this value is noted $Np^{TTI, m}_{i,max}$ since it is calculated for all TrCh with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $Np_{i,l}^{n} = 0$ to F_{max} -1:Positive or null: number of bits, in radio frame number *n* within the largest TTI, corresponding to the gap for compressed mode in this radio frame, for TrCH *i* with transport format *l*. The value will be null for the un-compressed radio frames. In case of fixed positions and compressed mode by puncturing, this value is noted $Np_{i,max}^{n}$ since it is calculated for all TrChs with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $N_{TGL}[k]$, k=0 to $F_i - 1$: Positive or null: number of bits in each radio frame corresponding to the gap for compressed mode for the CCTrCh.

- *RM_i*: Semi-static rate matching attribute for transport channel *i*. Signalled from higher layers.
- *PL:* Puncturing limit for uplink. This value limits the amount of puncturing that can be applied in order to avoid multicode or to enable the use of a higher spreading factor. Signalled from higher layers.

- $N_{data,j}$: Total number of bits that are available for the CCTrCH in a radio frame with transport format combination *j*.
- *I:* Number of TrCHs in the CCTrCH.
- Z_{ij} : Intermediate calculation variable.
- F_i : Number of radio frames in the transmission time interval of TrCH *i*.
- F_{max} Maximum number of radio frames in a transmission time interval used in the CCTrCH :

$$F_{\max} = \max_{1 \le i \le I} F_i$$

- n_i : Radio frame number in the transmission time interval of TrCH i ($0 \le n_i < F_i$).
- *q:* Average puncturing or repetition distance (normalised to only show the remaining rate matching on top of an integer number of repetitions). Used in uplink only.
- $I_F(n_i)$: The inverse interleaving function of the 1st interleaver (note that the inverse interleaving function is identical to the interleaving function itself for the 1st interleaver). Used in uplink only.
- $S(n_i)$: The shift of the puncturing or repetition pattern for radio frame n_i . Used in uplink only.
- $TF_i(j)$: Transport format of TrCH *i* for the transport format combination *j*.
- TFS(i) The set of transport format indexes *l* for TrCH *i*.
- *TFCS* The set of transport format combination indexes *j*.
- e_{ini} Initial value of variable e in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- e_{plus} Increment of variable *e* in the rate matching pattern determination algorithm of subclause4.2.7.5.
- e_{minus} Decrement of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- *b:* Indicates systematic and parity bits

b=1: Systematic bit. $\frac{X(t)x_k}{x_k}$ in subclause 4.2.3.2.1.

b=2: 1st parity bit (from the upper Turbo constituent encoder). $\frac{Y(t)_{zk}}{z}$ in subcaluse 4.2.3.2.1.

 $b=3: 2^{nd}$ parity bit (from the lower Turbo constituent encoder). $\frac{Y'(t)z'_k}{z'_k}$ in subclause 4.2.3.2.1.

The * (star) notation is used to replace an index x when the indexed variable X_x does not depend on the index x. In the left wing of an assignment the meaning is that " $X_* = Y$ " is equivalent to "**for all** <u>x</u> **do** $X_x = Y$ ". In the right wing of an assignment, the meaning is that " $Y = X_*$ " is equivalent to "**take any** <u>x</u> **and do** $Y = X_x$ ".

The following relations, defined for all TFC j, are used when calculating the rate matching parameters:

$$Z_{0,i} = 0$$

$$Z_{ij} = \left[\frac{\left\{ \left(\sum_{m=1}^{i} RM_{m} \cdot N_{mj} \right) \cdot N_{data, j} \right\}}{\sum_{m=1}^{l} RM_{m} \cdot N_{mj}} \right] \text{ for all } i = 1 \dots I$$

$$N_{m} = Z_{m} - Z_{m} - N_{m} \text{ for all } i = 1 \dots I$$

$$(1)$$

$$\Delta N_{ij} = Z_{ij} - Z_{i-1,j} - N_{ij} \qquad \text{for all } i = 1 \dots I$$

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4.2.11 2nd interleaving

The 2^{nd} interleaving is a block interleaver with inter-column permutations. The bits input to the 2^{nd} interleaver are denoted $u_{p1}, u_{p2}, u_{p3}, \dots, u_{pU}$, where *p* is PhCH number and *U* is the number of bits in one radio frame for one PhCH.

- (1) Set the number of columns $C_2 = 30$. The columns are numbered 0, 1, 2, ..., C_2 -1 from left to right.
- (2) Determine the number of rows R_2 by finding minimum integer R_2 such that:

-

 $U \leq R_2 C_2.$

(3) The bits input to the 2^{nd} interleaving are written into the $R_2 \times C_2$ rectangular matrix row by row.

$$\begin{bmatrix} u_{p1} & u_{p2} & u_{p3} & \dots & u_{p30} \\ u_{p31} & u_{p32} & u_{p33} & \dots & u_{p60} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ u_{p,((R_2-1)30+1)} & u_{p,((R_2-1)30+2)} & u_{p,((R_2-1)30+3)} & \dots & u_{p,(R_230)} \end{bmatrix}$$

(4) Perform the inter-column permutation based on the pattern $\{P_2(j)\}$ $(j = 0, 1, ..., C_2-1)$ that is shown in table 6, where $P_2(j)$ is the original column position of the *j*-th permuted column. After permutation of the columns, the bits are denoted by y_{pk} .

(5) The output of the 2nd interleaving is the bit sequence read out column by column from the inter-column permuted $R_2 \times C_2$ matrix. The output is pruned by deleting bits that were not present in the input bit sequence, i.e. bits y_{pk} that corresponds to bits u_{pk} with k>U are removed from the output. The bits after 2nd interleaving are denoted by $v_{p1}, v_{p2}, \ldots, v_{pU}$, where v_{p1} corresponds to the bit y_{pk} with smallest index *k* after pruning, v_{p2} to the bit y_{pk} with second smallest index *k* after pruning, and so on.

Table 6 Inter-column	permutation	pattern fo	or 2nd	interleaving
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Number of column C ₂	Inter-column permutation pattern
30	{0, 20, 10, 5, 15, 25, 3, 13, 23, 8, 18, 28, 1, 11, 21, 6, 16, 26, 4, 14, 24, 19, 9, 29, 12, 2, 7, 22, 27, 17}

3GPP TSG RAN Meeting #8 Düsseldorf, Germany, 21-23 June 2000

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Work item:	TS 25.212
Category:FA(only one categoryshall be markedCwith an X)D	CorrectionXRelease:Phase 2Corresponds to a correction in an earlier releaseRelease 96Release 96Addition of featureRelease 97Release 97Functional modification of featureRelease 98Release 98Editorial modificationRelease 00X
<u>Reason for</u> change:	Editorial corrections on section 4.2.9.2, reflecting the remarks made by Nortel Networks on the RAN1 reflector between RAN1#11 and RAN1#12.
Clauses affected	<u>1:</u> 4.2.9.2
Other specs affected:	Other 3G core specifications \rightarrow List of CRs:Other GSM core specifications \rightarrow List of CRs:MS test specifications \rightarrow List of CRs:BSS test specifications \rightarrow List of CRs:O&M specifications \rightarrow List of CRs:
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e.g. for 3GPP use the format TP-99xxx or for SMG, use the format P-99-xxx

4.2.9.2 2nd insertion of DTX indication bits

The DTX indication bits inserted in this step shall be placed at the end of the radio frame. Note that the DTX will be distributed over all slots after 2nd interleaving.

The bits input to the DTX insertion block are denoted by $s_1, s_2, s_3, \ldots, s_s$, where S is the number of bits from TrCH multiplexing. The number of PhCHs is denoted by P and the number of bits in one radio frame, including DTX indication bits, for each PhCH by R..

In normal mode $R = \frac{N_{data,*}}{P} = 15N_{data1} + 15N_{data2}$, where N_{data1} and N_{data2} are defined in [2].

For compressed mode, $N'_{data,*}$ is defined as $N_{data,*} = P(15N_{data1} + 15N_{data2})$. N_{data1} and N_{data2} are the number of bits in the data fields of the slot format used for the current compressed mode, i.e. slot format A or B as defined in [2] corresponding to the Spreading Factor and the number of transmitted slots in use.

In case of compressed mode by puncturing and fixed positions, DTX shall be inserted until $N'_{data,*}$ bits, since the exact room for the gap is already reserved thanks to the earlier insertion of the p-bits. Therefore R is defined as $R = N'_{data,*} / P_{\tau}$

In compressed mode by SF reduction and by higher layer scheduling, additional DTX with respect to normal mode shall be inserted if the transmission time reduction method transmission time reduction by higher layer scheduling does not exactly create a transmission gap of the desired TGL.

The number of bits available to the CCTrCH in one radio frame in compressed mode by SF reduction and by higher

layer scheduling is denoted by $N_{data,*}^{cm}$ and $R = \frac{N_{data,*}^{cm}}{P}$.

The exact value of N_{data*}^{cm} is dependent on the TGL and the transmission time reduction method, which are signalled

from higher layers. For the transmission time reduction by SF/2 method in compressed mode $N_{data,*}^{cm} = \frac{N'_{data,*}}{2}$,

and <u>fF</u>or other methods <u>compressed mode by higher layer scheduling the exact value of</u> $N_{data,*}^{cm}$ is dependent on the <u>TGL</u> which is signalled from higher layers. <u>iI</u>t can be calculated as $N_{data,*}^{cm} = N_{data,*}^{'} - N_{TGL}$. For every transmission time reduction method- $N_{data,*}^{'} = P(15N_{data1}^{'} + 15N_{data2}^{'})$, where $N_{data1}^{'}$ and $N_{data2}^{'}$ are the number of bits in the data fields of a slot for slot format A or B as defined in [2].

 N_{TGL} is the number of bits that are located within the transmission gap and defined as:

$$N_{TGL} = \begin{cases} \frac{TGL}{15} N'_{data,*}, \text{ if } N_{first} + TGL \le 15 \\ \frac{15 - N_{first}}{15} N'_{data,*}, \text{ in first frame if } N_{first} + TGL > 15 \\ \frac{TGL - (15 - N_{first})}{15} N'_{data,*}, \text{ in second frame if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

NOTE : In compressed mode by SF/2 method DTX is also added in physical channel mapping stage (subclause 4.2.12.2). During 2nd DTX insertion the number of CCTrCH bits is kept the same as in normal mode.

The bits output from the DTX insertion block are denoted by $w_1, w_2, w_3, \ldots, w_{(PR)}$. Note that these bits are four valued in case of compressed mode by puncturing, and three valued otherwise. They are defined by the following relations:

$$w_k = s_k$$
 k = 1, 2, 3, ..., S

$$w_k = \delta$$
 k = S+1, S+2, S+3, ..., PR

where DTX indication bits are denoted by δ . Here $s_k \in \{0,1, p\}$ and $\delta \notin \{0,1\}$.

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For submission list expected appro	to: TSG RA wal meeting # here ↑	N#8 for ap for infor	oproval mation	X		strates non-strates	gic (for SMG gic use only)
	Form: CR	cover sheet, version 2	? for 3GPF	P and SMG	The la ftp://	atest version of this /ftp.3gpp.org/Inform	s form is available from: nation/CR-Form-v2.doc
Proposed change (at least one should l	ge affects: be marked with ar	(U)SIM	ME	X	UTRAN	/ Radio X	Core Network
Source:	TSG RAN V	/G1				Date:	25 th May 2000
Subject:	Corrections	to 25.212 (Rate N	Matching	<mark>g, p-bit in</mark>	sertion, I	PhCH segment	ation)
Work item:	TS 25.212						
Category:F(only one categoryEshall be marked with an X)C	Correction Correspond Addition of f Functional r Editorial mo	s to a correction i eature nodification of fea dification	in an ea ature	rlier relea	ase	Release:	Phase 2Release 96Release 97Release 98Release 99XRelease 00
<u>Reason for</u> <u>change:</u>	 P1 1st IL formula PhCH set Indexing Specification Plenty context end 	column permuta $\Delta N^{TTI,m}_{b max} = \Sigma_{n=1}^{TTI,m}$ gmentation uses so by frame number f ation was using the f editorials, like p ditor for formulas	tion is n $o^{n=Fi} \Delta N^n$ metimes for CM b fact that utting sy , along v	ot named symbol V y punctur P1 is self whols in with using	d clearly I sum up t / instead of ring was r / inverse n italic an g unified	to F_{i} -1 of <i>U</i> not done with reg nd using equation mathematical	gards to largest TTI on editor instead notations.
Clauses affecte	d: 4.2.5. ;	<mark>4.2.7. ; 4.2.9.2. ;</mark>	<mark>4.2.10,</mark>	4.2.11			
<u>Other specs</u> affected:	Other 3G core Other GSM co specificati MS test speci BSS test speci O&M specific	e specifications ore ons fications cifications ations		$\begin{array}{l} \rightarrow \ \text{List of} \\ \rightarrow \ \text{List of} \end{array}$	f CRs: f CRs: f CRs: f CRs: f CRs: f CRs:		
<u>Other</u> comments:							

1st interleaving 4.2.5

In Compressed Mode by puncturing, bits marked with a fourth value on top of $\{0, 1, \delta\}$ and noted p, are introduced in the radio frames to be compressed, in positions corresponding to the first bits of the radio frames. They will be removed in a later stage of the multiplexing chain to create the actual gap. Additional puncturing has been performed in the rate matching step, over the TTI containing the compressed radio frame, to create room for these p-bits. The following subclause describes this feature.

4.2.5.1 Insertion of marked bits in the sequence to be input in first interleaver

In normal mode, compressed mode by higher layer scheduling, and compressed mode by spreading factor reduction:

 $\mathbf{x}_{ik} = \mathbf{Z}_{ik}$ and $\mathbf{X}_i = \mathbf{Z}_i$, $x_{ik} = z_{ik}$ and $X_i = Z_i$

In case of compressed mode by puncturing and fixed positions, sequence $\underline{x}_{i,k} \neq_{i-k}$ which will be input to first interleaver for TrChH *i* and TTI *m* within largest TTI, is built from bits $\underline{Z_{i,k}} = 1, \dots, Z_i$, plus $Np_{i,\max}^{TTI,m} Np^{TTI,m}$ bits

marked p and $X_i = Z_i + N p_{i,max}^{TTI,m} N p_{i,max}^{TTI,m}$, as is described thereafter.

 $Np_{i,\max}^{TTI,m} Np^{TTI,m}$ is defined in the Rate Matching subclause 4.2.7.

 $P_{1} = P_{1} + 1$ defines the inter column permutation function for a TTI of length $F_{1} = 10$ ms, as defined in Table 3 above in section 4.2.5.2. $P_{1_{F_i} \to F_i}(x_i)$ is the Bit Reversal function of x on $\log_2(F_i \to F_i)$ bits.

- NOTE 1: C[x], x=0 to $F_{i}F_{i} 1$, the number of bits p which have to be inserted in each of the $F_{i}F_{i}$ segments of the TTI, where x is the column number before permutation, i.e. in each column of the first interleaver. $C[\underline{P1}_{Fi}(x)]$ is equal to $Np_{i,\max}^{m \cdot F_i + x} Np_{i,\max}^{*}$ for x equal 0 to $F_i = -1$ for fixed positions. It is noted $Np_i^{m \times F_i + x} Np_i^{*}$ in the following initialisation step.
- NOTE 2: cbi[x], x=0 to F_{F} 1, the counter of the number of bits p inserted in each of the F_{F} segments of the TTI, i.e. in each column of the first interleaver. x is the column number before permutation.

col = 0

while $col < F_i$ do -- here col is the column number after column permutation

segments of the TTI number m

 $\operatorname{cbi}[\underline{\operatorname{P1}}_{Fi}(\operatorname{col})] = 0$ -- initialisation of counter of number of bits p inserted in each of the F.F. segments of the TTI

col = col + 1

end do

n = 0, m = 0

while $n < X_i - X_j - from$ here col is the column number before column permutation

 $col = n \mod F_i$

if $cbi[col] < C[P_{Fi}(col)]$ do

 $\mathbf{x}_{i,n} \cdot \underline{x}_{i,n} = \mathbf{p}$

-- insert one p bit cbi[col] = cbi[col]+1-- update counter of number of bits p inserted

-- no more p bit to insert in this segment

else

```
x_{i,n} = z_{i,n}, x_{j,n} = z_{i,m}
m = m+1
endif
n = n + 1
```

end do

4.2.5.2 1st interleaver operation

The 1st interleaving is a block interleaver with inter-column permutations. The input bit sequence to the 1st interleaver is denoted by $x_{i,1}, x_{i,2}, x_{i,3}, \dots, x_{i,X_i}, x_{i1}, x_{i2}, x_{i3}, \dots, x_{iX_i}$, where *i* is TrCH number and X_i the number of bits (at this stage X_i is assumed and guaranteed to be an integer multiple of TTI). The output bit sequence is derived as follows:

- (1) Select the number of columns $\underline{C1}C_{4}$ from table 3.
- (2) Determine the number of rows $\underline{R1}R_{I}$ defined as:
 - $\underline{\mathbf{R1}}\mathbf{R}_{\mathrm{H}} = \underline{X_{i}}/\underline{\mathbf{C1}}\underline{\mathbf{X}_{i}}/\underline{\mathbf{C}}_{\mathrm{H}}$
- (3) Write the input bit sequence into the <u>R1R_r × C1C_t</u> rectangular matrix row by row starting with bit $x_{i,1}$ in the first column of the first row and ending with bit $x_{i,(R1\times C1)} \cdot x_{i,(R_tC_1)}$ in column <u>C1C_t</u> of row <u>R1R_t</u>:

$$\begin{bmatrix} x_{i,1} & x_{i,2} & x_{i,3} & \dots & x_{i,C1} \\ x_{i,(C1+1)} & x_{i,(C1+2)} & x_{i,(C1+3)} & \dots & x_{i,(2\times C1)} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x_{i,((R1-1)\times C1+1)} & x_{i,((R1-1)\times C1+2)} & x_{i,((R1-1)\times C1+3)} & \dots & x_{i,(R1\times C1)} \end{bmatrix}$$

$$\begin{bmatrix} x_{i1} & x_{i2} & x_{i3} & \dots & x_{iC_{I}} \\ x_{i,(C_{I}+1)} & x_{i,(C_{I}+2)} & x_{i,(C_{I}+3)} & \dots & x_{i,(2C_{I})} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ x_{i,((R_{I}-1)C_{I}+1)} & x_{i,((R_{I}-1)C_{I}+2)} & x_{i,((R_{I}-1)C_{I}+3)} & \dots & x_{i,(R_{I}C_{I})} \end{bmatrix}$$

(4) Perform the inter-column permutation based on the pattern $\langle P1_{C1}(j) \rangle_{j \in \{0,1,...,C1-1\}} \{P_{1}(j)\} (j=0,1,...,C-1)$ shown in table 3, where $\underline{P1_{C1}}_{P_{1}}P_{1}(j)$ is the original column position of the *j*-th permuted column. After permutation of the columns, the bits are denoted by $y_{i,k}$:

$\int y_{i,1}$	$y_{i,(R1+1)}$	$y_{i,(2 \times \text{R1+1})}$	y	$V_{i,((C1-1)\times R1+1)}$	y_{i1}	$\mathcal{Y}_{i,(R_I+1)}$	$y_{i,(2R_I+1)}$	y	$\langle i, ((C_I - 1)R_I + 1) \rangle$
<i>Y</i> _{<i>i</i>,2}	$y_{i,(\text{R1+2})}$	$y_{i,(2 \times \text{R1}+2)}$	y	,((C1–1)×R1+2)	y_{i2}	$\mathcal{Y}_{i,(R_I+2)}$	$y_{i,(2R_I+2)}$	y	$V_{i,((C_I - 1)R_I + 2)}$
:	:	:		÷	:	:	:		:
$y_{i,R1}$	$y_{i,(2 \times R1)}$	$y_{i,(3 \times R1)}$		$y_{i,(C1\times R1)}$	y_{iR_I}	$y_{i,(2R_I)}$	$y_{i,(3R_I)}$		$y_{i,(C_I R_I)}$

(5) Read the output bit sequence $y_{i1}, y_{i2}, y_{i3}, \dots, y_{i,(Cl\times R1)}, y_{i1}, y_{i2}, y_{i3}, \dots, y_{i,(C_lR_l)}$ of the 1st interleaving column by column from the inter-column permuted $\underline{R1R_t} \times \underline{C1C_t}$ matrix. Bit $y_{i,1}$ corresponds to the first row of the first column and bit $y_{i,(R1\times C1)}, y_{i,(R_lC_l)}$ corresponds to row $\underline{R1R_t}$ of column $\underline{C1C_t}$.

TTI	Number of columns C1C	Inter-column permutation patterns <u><p1<sub>C1(0),, P1_{C1}(C1-1)></p1<sub></u>
10 ms	1	<u><</u> {0} <u>></u>
20 ms	2	<u> </u>
40 ms	4	<u><{</u> 0,2,1,3 <u>}≥</u>
80 ms	8	<u><</u> {0,4,2,6,1,5,3,7 <u>}</u> ≥

Table 3

4.2.5.3 Relation between input and output of 1st interleaving in uplink

The bits input to the 1st interleaving are denoted by $t_{i,1}, t_{i,2}, t_{i,3}, \dots, t_{i,T_i} = t_{i1}, t_{i2}, t_{i3}, \dots, t_{iT_i}$, where *i* is the TrCH number and T_i the number of bits. Hence, $z_{ik} = t_{ik} - \underline{z_{i,k}} = t_{i,k}$ and $X_i = T_i$.

The bits output from the 1st interleaving are denoted by $d_{i,1}, d_{i,2}, d_{i,3}, \dots, d_{i,T_i}, \frac{d_{i,1}, d_{i,2}, d_{i,3}, \dots, d_{i,T_i}}{d_{i,1}, d_{i,2}, d_{i,3}, \dots, d_{i,T_i}}$, and $\underline{d_{ik}} = \underline{y_{ik}} \underline{d_{i,k}}$.

4.2.7 Rate matching

Rate matching means that bits on a transport channel are repeated or punctured. Higher layers assign a rate-matching attribute for each transport channel. This attribute is semi-static and can only be changed through higher layer signalling. The rate-matching attribute is used when the number of bits to be repeated or punctured is calculated.

The number of bits on a transport channel can vary between different transmission time intervals. In the downlink the transmission is interrupted if the number of bits is lower than maximum. When the number of bits between different transmission time intervals in uplink is changed, bits are repeated or punctured to ensure that the total bit rate after TrCH multiplexing is identical to the total channel bit rate of the allocated dedicated physical channels.

If no bits are input to the rate matching for all TrCHs within a CCTrCH, the rate matching shall output no bits for all TrCHs within the CCTrCH and no uplink DPDCH will be selected in the case of uplink rate matching.

Notation used in subcaluse 4.2.7 and subclauses:

 $N_{i,j}$: For uplink: Number of bits in a radio frame before rate matching on TrCH *i* with transport format combination *j*.

For downlink: An intermediate calculation variable (not an integer but a multiple of 1/8).

 $\frac{N_{i,l}^{TTI}}{M_{il}^{TTI}} \cdot \frac{N_{il}^{TTI}}{M_{il}}$: Number of bits in a transmission time interval before rate matching on TrCH *i* with transport format *l*. Used in downlink only.

 $\frac{\Delta N_{i,j}}{\Delta N_{ij}}$: For uplink: If positive - number of bits that should be repeated in each radio frame on TrCH *i* with transport format combination *j*.

If negative - number of bits that should be punctured in each radio frame on TrCH *i* with transport format combination *j*.

For downlink : An intermediate calculation variable (not an integer but a multiple of 1/8).

 $\Delta N_{i,l}^{TTI} \Delta N_{il}^{TTI}$: If positive - number of bits to be repeated in each transmission time interval on TrCH *i* with transport format <u>*j*l</u>.

If negative - number of bits to be punctured in each transmission time interval on TrCH *i* with transport format $j\underline{l}$.

Used in downlink only.

 $\frac{Np_{i,l}^{TTI,m}}{Mp} \xrightarrow{P} \frac{Mp}{k} m = 0 \text{ to } (F_{max} / F_{i}) - 1 \text{ :Positive or null: number of bits to be removed in TTI number$ *m*within the largest TTI, to create the required gaps in the compressed radio frames of this TTI, in case of compressed mode by puncturing, for TrChH*i*with transport format*l*. In case of fixed positions and

compressed mode by puncturing, this value is noted $Np_{i,max}^{TTI,m} Np_{i,max}^{TTI,m}$ since it is calculated for all

 $TrCh\underline{H}$ with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $Np_{i,l}^n Np_{i,l}^n = 0$ to F_{max} -1:Positive or null: number of bits, in radio frame number *n* within the largest TTI,

corresponding to the gap for compressed mode in this radio frame, for TrCH *i* with transport format *l*. The value will be null for the <u>un-compressed</u> radio frames <u>not overlapping with a transmission gap</u>. In case of fixed positions and compressed mode by puncturing, this value is noted $Np_{i,\max}^n Np_{i,\max}^n$ since it is calculated for all TrChHs with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $N_{TGL}[k] N_{TGL}[k]$, k=0 to $F_{e} - 1 F_{max} - 1$: Positive or null: number of bits in each radio frame corresponding to the gap for compressed mode for the CCTrChH.

- *RM_i*: Semi-static rate matching attribute for transport channel *i*. Signalled from higher layers.
- *PL:* Puncturing limit for uplink. This value limits the amount of puncturing that can be applied in order to avoid multicode or to enable the use of a higher spreading factor. Signalled from higher layers.
- $N_{data,j}$: Total number of bits that are available for the CCTrCH in a radio frame with transport format combination *j*.
- *I:* Number of TrCHs in the CCTrCH.
- $Z_{i,j}$: Intermediate calculation variable.
- F_i : Number of radio frames in the transmission time interval of TrCH *i*.
- F_{max} Maximum number of radio frames in a transmission time interval used in the CCTrCH :

$$F_{\max} = \max_{1 \le i \le I} F_i$$

- n_i : Radio frame number in the transmission time interval of TrCH i ($0 \le n_i < F_i$).
- *q:* Average puncturing or repetition distance (normalised to only show the remaining rate matching on top of an integer number of repetitions). Used in uplink only.
- $I_E \underline{P1}_E(n_i)$: The inverse interleaving column permutation function of the 1st interleaver, $\underline{P1}_E(x)$ is the original position of column with number x after permutation. P1 is defined on table 3 of section 4.2.5.2 (note that the $\underline{P1}_E$ inverse interleaving function is identical to the interleaving function itself for the 1st interleaver). is self-inverse. Used for rate matching in uplink only.
- <u>S[n]</u> <u>S(n_i)</u>: The shift of the puncturing or repetition pattern for radio frame n_i when $n = P1_{F_i}(n_i)$. Used in uplink only.
- <u> $TF_i(j)$ </u> Transport format of TrCH *i* for the transport format combination *j*.
- <u>*TFS(i)*</u> The set of transport format indexes l for TrCH i.
- *TFCS* The set of transport format combination indexes *j*.
- *e*_{ini} Initial value of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- e_{plus} Increment of variable *e* in the rate matching pattern determination algorithm of subclause4.2.7.5.
- e_{minus} Decrement of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- *b:* Indicates systematic and parity bits

b=1: Systematic bit. X(t) in subclause 4.2.3.2.1.

b=2: 1^{st} parity bit (from the upper Turbo constituent encoder). *Y*(*t*) in subcaluse 4.2.3.2.1.

b=3: 2^{nd} parity bit (from the lower Turbo constituent encoder). *Y*(*t*) in subclause 4.2.3.2.1.

The * (star) notation is used to replace an index x when the indexed variable X_x does not depend on the index x. In the left wing of an assignment the meaning is that " $X_* = Y$ " is equivalent to "**for all** <u>x</u> **do** $X_x = Y$ ". In the right wing of an assignment, the meaning is that " $Y = X_*$ " is equivalent to "**take any** <u>x</u> **and do** $Y = X_x$ ".

The following relations, defined for all TFC j, are used when calculating the rate matching parameters:

$$Z_{0,i} = 0$$



4.2.7.1 Determination of rate matching parameters in uplink

4.2.7.1.1 Determination of SF and number of PhCHs needed

In uplink, puncturing can be applied to match the CCTrCH bit rate to the PhCH bit rate. The bit rate of the PhCH(s) is limited by the UE capability and restrictions imposed by UTRAN, through limitations on the PhCH spreading factor. The maximum amount of puncturing that can be applied is signalled from higher layers and denoted by *PL*. The number of available bits in the radio frames of one PhCH for all possible spreading factors is given in [2]. Denote these values by N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , and N_4 , where the index refers to the spreading factor. The possible number of bits available to the CCTrCH on all PhCHs, N_{data} , then are $\{N_{2567}, N_{1285}, N_{647}, N_{327}, N_{167}, N_{37}, 4N_{47}, 5N_{47}, 6N_{47}\}$ [N_{2562} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , N_4 , $2\times N_4$, $3\times N_4$, $4\times N_4$, $5\times N_4$, $6\times N_4$]. Depending on the UE capability and the restrictions from UTRAN, the allowed set of N_{datar} , denoted SET0, can be a subset of { $N_{256}, N_{128}, N_{64}, N_{32}, N_{16}, N_8, N_4, 2\times N_4, 3\times N_4$, $4\times N_4$, $5\times N_4$, $4N_4$, $5N_4$, $6N_4$]. N_{data} , for the transport format combination *j* is determined by executing the following algorithm:

SET1 = {
$$N_{data}$$
 in SET0 such that

$$\left(\min_{1 \le y \le I} \left\{ RM_y \right\} \right) \times N_{data} - \sum_{x=1}^{I} RM_x \times N_{x,j} - \frac{\min_{1 \le y \le I} \left\{ RM_y \right\} N_{data} - \sum_{x=1}^{I} RM_x - N_{x,j}} \text{ is non negative } \}$$

If SET1 is not empty and the smallest element of SET1 requires just one PhCH then

$$N_{data,i} = \min \text{SET1}$$

else

SET2 = {
$$N_{data}$$
 in SET0 such that

$$\frac{\left(\min_{1\leq y\leq I} \left\{RM_{y}\right\}\right) \times N_{data} - PL \times \sum_{x=1}^{I} RM_{x} \times N_{x,j}}{\max_{1\leq y\leq I} \left\{RM_{y}\right\} N_{data} - PL \cdot \sum_{x=1}^{I} RMx \cdot N_{x,j}} \text{ is non}$$
negative }

0

Sort SET2 in ascending order

 $N_{data} = \min \text{SET2}$

While N_{data} is not the max of SET2 and the follower of N_{data} requires no additional PhCH do

 N_{data} = follower of N_{data} in SET2

End while

$$\mathbf{N}_{\text{data},i} = \mathbf{N}_{\text{data}} N_{\text{data},i} = N_{\text{data}}$$

End if

4.2.7.1.2 Determination of parameters needed for calculating the rate matching pattern

The number of bits to be repeated or punctured, $\Delta N_{i,j}$, within one radio frame for each TrCH *i* is calculated with equation 1 for all possible transport format combinations *j* and selected every radio frame. $N_{data,j}$ is given from subclause 4.2.7.1.1.
In compressed mode $N_{data,j}$ is replaced by $N_{data,j}^{cm}$ in Equation 1. $N_{data,j}^{cm}$ is given as follows:

In compressed mode by higher layer scheduling, $N_{data,j}^{cm}$ is obtained by executing the algorithm in subclause 4.2.7.1.1 but with the number of bits in one radio frame of one PhCH reduced to $\frac{N_{tr}}{15}$ of the value in normal mode.

 N_{tr} is the number of transmitted slots in a compressed radio frame and is defined by the following relation:

$$N_{tr} = \begin{cases} 15 - TGL, \text{ if } N_{first} + TGL \le 15\\ N_{first}, \text{ in first frame if } N_{first} + TGL > 15\\ 30 - TGL - N_{first}, \text{ in second frame if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

In compressed mode by spreading factor reduction,
$$N_{data,j}^{cm} = 2 \times (N_{data,j} - N_{TGL}) \cdot N_{data,j}^{cm} = 2N_{data,j} - 2N_{TGL}$$
,
where $N_{TGL} = \frac{15 - N_{tr}}{15} \times N_{data,j} \cdot N_{TGL} = \frac{15 - N_{tr}}{15} \cdot N_{data,j}$

If $\Delta N_{i,j} = 0$ then the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed.

If $\Delta N_{i,j} \neq 0$ the parameters listed in subclauses 4.2.7.1.2.1 and 4.2.7.1.2.2 shall be used for determining e_{ini} , e_{plus} , and e_{minus} (regardless if the radio frame is compressed or not).

4.2.7.1.2.1 Uncoded and convolutionally encoded TrCHs

 $R = \underline{AN_{ij} \mod N_{ij}} \underbrace{M_{ij} \mod N_{i,j}}_{i,j} - \text{ note: in this context } \underline{\Delta N_{i,j}} \mod N_{i,j} \underbrace{M_{ij}}_{i,j} \mod N_{i,j} \text{ is in the range of } 0 \text{ to } N_{i,j} - 1 \text{ i.e. -1} \\ \mod 10 = 9.$

if $R \neq 0$ and $2 \leq R \leq N_{i,j}$ then $q = \left\lceil \frac{N_{ij}N_{ij}}{R_{ij}} \right\rceil$

else

 $\mathbf{q} = \left[\underline{N}_{\underline{i},j} \mathbf{N}_{\underline{i},j} / (\mathbf{R} \cdot \underline{R} \cdot \underline{N}_{\underline{i},j} \mathbf{N}_{\underline{i},j}) \right]$

endif

-- note: <u>q-q</u> is a signed quantity.

if q is even

then $q' = q + gcd(|q|, \underline{F}_i F_i) / \underline{F}_i F_i$ -- where gcd $(|q|, \underline{F}_i F_i)$ means greatest common divisor of |q| and $\underline{F}_i F_i$

```
-- note that q' q' is not an integer, but a multiple of 1/8
```

else

q' = q

endif

for x = 0 to $F_i F_i - 1$

 $S[(I_{F} (| x \times x^{q}] | mod \underline{F}, F_{i}))] = (| x^{q}| | div \underline{F}, F_{i})$

end for

 $\Delta \mathbf{N}_{i} = \Delta \mathbf{N}_{i,j} \ \Delta \underline{N}_{\underline{i}} = \Delta \underline{N}_{\underline{i},j}$

a = 2

For each radio frame, the rate-matching pattern is calculated with the algorithm in subclause 4.2.7.5, where :

$$\begin{split} \underline{X_i = N_{i,j}, \text{ and}} \\ \underline{e_{ini} = (a \times S[P1_{Fi}(n_i)] \times |\Delta N_i| + 1) \mod (a \cdot N_{ij}).} \\ \underline{e_{plus} = a \times N_{i,j}} \\ \underline{e_{minus} \equiv a \times |\Delta N_i|} \\ \underline{X_i = N_{i,j}, \text{ and}} \\ \underline{e_{ini} = (a \cdot S(n_i) \cdot |\Delta N_i| + 1) \mod (a \cdot N_{ij}).} \\ \underline{e_{plus} = a \cdot N_{ij}} \\ \underline{e_{minus} = a \cdot |\Delta N_i|} \\ puncturing for \underline{AN \Delta N} < 0, repetition otherwise. \end{split}$$

4.2.7.1.2.2 Turbo encoded TrCHs

If repetition is to be performed on turbo encoded TrCHs, i.e. $\underline{A\Delta}N_{i,j} > 0$, the parameters in subclause 4.2.7.1.2.1 are used.

If puncturing is to be performed, the parameters below shall be used. Index b is used to indicate systematic (b=1), 1^{st} parity (b=2), and 2^{nd} parity bit (b=3).

a=2 when b=2 a=1 when b=3 $\Delta N_{i} = \begin{cases} \left[\Delta N_{i,j} / 2 \right], & b=2\\ \left[\Delta N_{i,j} / 2 \right], & b=3 \end{cases}$

If ΔN_i is calculated as 0 for b=2 or b=3, then the following procedure and the rate matching algorithm of subclause 4.2.7.5 don't need to be performed for the corresponding parity bit stream.

$$\mathbf{X}_{i} = \left\lfloor \mathbf{N}_{i,j} / 3 \right\rfloor,$$
$$\mathbf{q} = \left\lfloor \mathbf{X}_{i} / |\Delta \mathbf{N}_{i}| \right\rfloor$$

 $if(q \le 2)$

for $x\underline{r}=0$ to F_i -1

 $S[\underline{I_{F}}(3 \times r \times + b-1) \mod \underline{F_{r}}] = \underline{K_{r}} \mod 2;$

end for

else

```
if q is even
```

```
then \underline{q'}\underline{q}' = \underline{q}\underline{-}gcd(\underline{q}\underline{q}, F_i)/F_i -- where gcd(\underline{q}\underline{q}, F_i) means greatest common divisor of \underline{q}\underline{q} and F_i -- note that \underline{q'}\underline{q'} is not an integer, but a multiple of 1/8
```

else q' = q

endif

for x=0 to F_i -1

 $r = \left[\frac{x + x + q}{x + q} \right] \mod \frac{F_i F_i}{F_i};$

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$$S[I_{F}(3 \times r+b-1) \mod F_{i} \underbrace{F_{i}}] = [\times \underbrace{x \times q'} \dim F_{i} \underbrace{F_{i}}];$$

endfor

endif

For each radio frame, the rate-matching pattern is calculated with the algorithm in subclause 4.2.7.5, where:

 X_i is as above:

 $\mathbf{e}_{\text{ini}} = (\mathbf{a} \times \mathbf{S}[\underline{P1}_{i'i'}(\underline{n}_i)](\underline{\mathbf{n}}_i) \times |\Delta \mathbf{N}_i| + \mathbf{X}_i) \mod (\mathbf{a} \cdot \mathbf{X}_i \times \mathbf{X}_i), \text{ if } \mathbf{e}_{\text{ini}} = 0 \text{ then } \underline{e}_{ini} = \mathbf{a} \cdot \mathbf{X}_i \cdot \mathbf{e}_{\text{ini}} = \mathbf{a} \cdot \mathbf{X}_i \cdot \mathbf{x}_i$

 $e_{\text{plus}} = \underline{a \times X_i} a \cdot X_i$ $e_{\text{minus}} = \underline{a \times |\Delta N_i|} a \cdot |\Delta N_i|$

4.2.7.2 Determination of rate matching parameters in downlink

For downlink $N_{data,j}$ does not depend on the transport format combination j. $N_{data,*}$ is given by the channelization code(s) assigned by higher layers. Denote the number of physical channels used for the CCTrCH by P. N_{data,*} is the number of bits available to the CCTrCH in one radio frame and defined as $N_{data} = P \times 15 \times (15N_{data1} + 15N_{data2})$, where N_{data1} and N_{data2} are defined in [2]. Note that contrary to the uplink, the same rate matching patterns are used in normal and compressed mode by spreading factor reduction or higher layer scheduling.

In the following, the total amount of puncturing or repetition for the TTI is calculated.

Additional calculations for compressed mode by puncturing in case of fixed positions are performed to determine this total amount of rate matching needed.

For compressed mode by puncturing, in TTIs where some compressed radio frames occur, the puncturing is increased or the repetition is decreased compared to what is calculated according to the rate matching parameters provided by higher layers. This allows to create room for later insertion of marked bits, noted p-bits, which will identify the positions of the gaps in the compressed radio frames.

The amount of additional puncturing corresponds to the number of bits to create the gap in the TTI for TrChH *i*. In case of fixed positions, it is calculated in addition to the amount of rate matching indicated by higher layers. It is noted $Np_{i,\max}^{TTI,m} Np^{TTI,m}$

In fixed positions case, to obtain the total rate matching $\Delta N_{i,\max}^{TTI,cm,m}$ to be performed on the TTI *m*, $Np_{i,\max}^{TTI,m} Np_{i,\max}^{TTI,m}$ $_{i,max}$ is sub-stracted from $\Delta N_{i,max}^{TTI,m} \Delta N_{i,max}^{TTI,m}$ (calculated based on higher layers RM parameters as for normal rate

matching). This allows to create room for the $Np_{i,\max}^{TTI,m} Np_{i,\max}^{TTI,m}$ bits p to be inserted later. If the result is null, i.e. the

amount of repetition matches exactly the amount of additional puncturing needed, then no rate matching is necessary.

In case of compressed mode by puncturing and fixed positions, for some calculations, N'_{data,*} is used for radio frames with gap instead of $N_{data,*}$, where $N_{data,*} = P \times 15 \times (N_{data1} + N_{data2}) \cdot N_{data,*} = P(15N_{data1} + 15N_{data2})$.

 N_{data1} and N_{data2} are the number of bits in the data fields of the slot format used for the current compressed mode, i.e. slot format A or B as defined in [2] corresponding to the Spreading Factor and the number of transmitted slots in use.

The number of bits corresponding to the gap for TrCh i TrCH i, in each radio frame of its TTI is calculated using the number of bits to remove on each Physical Channel $N_{TGL}[k]$, where k is the radio frame number in the largest TTI.

For each radio frame k of the largest TTI that is overlapping with a transmission gap, N_{TGL}[k] is given by the relation:

$$\frac{TGL}{15} \times N_{data,*}^{'} \frac{TGL}{15} N_{data,*}^{'}, \text{ if } N_{first} + TGL \le 15$$

 $N_{TGL} = \left\{ \right.$

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$$\frac{\frac{15 - N_{first}}{15} \times N'_{data,*}}{\frac{15 - N_{first}}{15}} \frac{N'_{data,*}}{15}$$
, in first radio frame of the gap if $N_{first} + TGL > 15$
$$\frac{TGL - (15 - N_{first})}{\frac{15}{TGL > 15}} \times N'_{data,*} \frac{TGL - (15 - N_{first})}{15} N'_{data,*}$$
, in second radio frame of the gap if $N_{first} + 15$

 N_{first} and TGL are defined in subclause 4.4.

Note that $N_{\text{TGL}}[k] = 0$ if radio frame k is not <u>compressed</u> overlapping with a transmission gap.

4.2.7.2.1 Determination of rate matching parameters for fixed positions of TrCHs

4.2.7.2.1.1 Calculation of $\Delta N_{i,max}$ for normal mode and compressed mode by higher layer scheduling and spreading factor reduction

First an intermediate calculation variable $N_{i,*}$ is calculated for all transport channels *i* by the following formula:

$$N_{i,*} = \frac{1}{F_i} \times \left(\max_{l \in TFS(i)} N_{i,l}^{TTI}\right) N_{i,*} - \frac{1}{F_i} \cdot \max_{l \in TFS(i)} N_{i,l}^{TTI}$$

In order to compute The computation of the $\Delta N_{i,l}^{TTI}$ parameters is then performed in for all TrCH *i* and all TF *l*, we first compute an intermediate parameter $\Delta N_{i,max}$ by the following formula, where $\Delta N_{i,*}$ is derived from $N_{i,*}$ by the formula given at subclause 4.2.7:

$$\Delta N_{i,max} = F_i \times \Delta N_{i,*} - \frac{\Delta N_{max} = F_i \cdot \Delta N_{i,*}}{\Delta N_{max} = F_i \cdot \Delta N_{i,*}}$$

If $\Delta N_{i,max} = 0$ $\Delta N_{max} = 0$ then, for TrCH *i*, the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed. In this case we have :

$$\forall l \in TFS(i) \Delta N_{i,l}^{TTI} = 0$$

If $\Delta N_{i,max} \neq 0$ $\Delta N_{max} \neq 0$ the parameters listed in subclauses 4.2.7.2.1.3 and 4.2.7.2.1.4 shall be used for determining e_{ini} , e_{plus} , and e_{minus} and $\Delta N_{i,l}^{TTI}$.

4.2.7.2.1.2 Calculations for compressed mode by puncturing

Calculations of $\Delta N_{i,\max}^{TTI,m} \Delta N^{TTI,m}$ for all TTI *m* within largest TTI, for all TrChH *i*

First an intermediate calculation variable $N^{n}_{i,\pm}$ is calculated for all transport channels *i* and all frames *n* in TTI *m* within the largest TTI, using the same formula as for normal mode above by replacing $N^{TTI}_{i,i}$ by $N^{TTI,m}_{i,i}$, the number of bits in TTI *m*.

First an intermediate calculation variable $N_{i,*}$ is calculated for all transport channels *i* by the following formula:

$$N_{i,*} = \frac{1}{F_i} \times \left(\max_{l \in TFS(i)} N_{i,l}^{TTI} \right)$$

Then an intermediate calculation variable $\Delta N_{i,*}^n$ is derived from $N_{i,*}$ by the formula given at subclause 4.2.7, for all

<u>TrCH *i*</u> and all frames *n* in the largest TTI, from the formula given at subclause 4.2.7 using $N_{data,*}$ when index *n* designates a radio frame of the largest TTI that is not overlapping with a transmission gap, and using $N'_{data,*}$ instead of $N_{data,*}$, when index *n* designate a frame that is overlapping with a transmission gap.

In order to compute the $\Delta N_{i,l}^{TTI,m}$ parameters for all TrCH *i*, all TF *l* and all TTI with number *m* in the largest TTI, we first compute an intermediate parameter $\Delta N_{i,max}^{m}$ by the following formula ::

$$\Delta N_{i,\max}^{m} = \sum_{n=m\times F_{i}}^{n=(m+1)\times F_{i}-1} \Delta N_{i,*}^{n}$$

The computation of the $\Delta N^{TTI,m}$ -i.max parameters is then performed for all TrCH *i* by the following formula,

$$4N^{TTI,m} - \sum_{n=0}^{n=Fi} 4N^{n} - i *$$

where all $\Delta N^{n}_{-i,*}$ - are derived from $N^{n}_{-i,*}$ for all TrCh i and all frames *n* in TTI *m*, from the formula given at subclause 4.2.7 using $N_{data,*}$ for the non compressed frames of TTI *m* and using $N'_{data,*}$ instead of $N_{data,*}$, for the compressed frames of TTI *m*.

Calculations of $Np_{i,\max}^n Np^{-m}$ and $Np_{i,\max}^{TTI,m} Np^{-TTI,m}$

Let $Np_{i,\max}^n Np_{i,\max}^n$ be the number of bits to eliminate on TrChH *i* to create the gap for compressed mode, in each radio frame k of the TTI, calculated for the Transport Format Combination of TrChH *i*, in which the number of bits of TrChH *i* is at its maximum.

 $Np_{i,\max}^n Np_{i,\max}^n$ is calculated for each radio frame k of the TTI in the following way.

Intermediate variables Z_i for i = 1 to I are calculated using the formula (1) in 4.2.7, by replacing $N_{-data,i}$ by $N_{TGL}[\underline{kn}]$.

Then
$$Np_{i,\max}^n Np_{i,\max}^m = (Z_i - Z_{i-1})$$
 for $i = 1$ to I

The total number of bits $Np_{i,\max}^{TTI,m} Np_{i,\max}^{TTI,m}$ corresponding to the gaps for compressed mode for TrChH i in the TTI is calculated as:

$$Np_{i,\max}^{TTI,m} = \sum_{n=m\times F_i}^{n=(m+1)\times F_i-1} Np_{i,\max}^n$$

 $Np^{TTI, m}_{i,max} = \Sigma_{n=0}^{Fi-1} Np^{m}_{i,max}$

If $\Delta N_{max} = Np^{TTI, m}$, then, for TrCH *i*, the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed. If $\Delta N_{max} \neq Np^{-TTI, m}$, then, for TrCH *i*, the rate matching algorithm of subclause 4.2.7.5 needs to be executed.

The amount of rate matching $\Delta N_{i,\max}^{TTI,cm,m}$ for the highest TrCH bit rate is then computed by the following formula :

$$\Delta N_{i,\max}^{TTI,cm,m} = \Delta N_{i,\max}^{m} \Delta N^{TTI,m} - Np_{i,\max}^{TTI,m} Np^{TTI,m} Np^{TTI,m} + Np^{TTI,m} Np^{TTI,m} + Np^{$$

If $\Delta N_{i,\text{max}}^{TT1,cm,m} = 0$, then, for TrCH *i*, the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed.

If $\Delta N_{i,\max}^{TTI,cm,m} \neq 0$, then, for TrCH *i*, the rate matching algorithm of subclause 4.2.7.5 needs to be executed, and the parameters listed in subclauses 4.2.7.2.1.3 and 4.2.7.2.1.4 shall be used for determining e_{inis}, e_{plus} , and e_{minus} and $\Delta N_{i,l}^{TTI,m}$.

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4.2.7.2.1.3 Determination of rate matching parameters for uncoded and convolutionally encoded TrCHs

$$\Delta N_i = \Delta N_{i,max} \frac{\Delta N_i = \Delta N_{max}}{\Delta N_i = \Delta N_{max}}$$

For compressed mode by puncturing, ΔN_i is defined as: $\Delta N_i = \Delta N_{i,\max}^{TTI,cm,m}$, instead of the previous relation.

a=2

$$N_{max} = \max_{l \in TFS(i)} N_{il}^{TTI}$$

For each transmission time interval of TrCH *i* with TF *l*, the rate-matching pattern is calculated with the algorithm in subclause 4.2.7.5. The following parameters are used as input:

$$X_{i} = N_{il}^{III}$$

$$e_{ini} = 1$$

$$e_{plus} = a \times N_{max} \cdot e_{plus} = a \cdot N_{max}$$

$$e_{\min us} = a \times |\Delta N_{i}| \cdot e_{\min us} = a \cdot |\Delta N_{i}|$$

Puncturing if $\Delta N_i < 0$, repetition otherwise. The values of $\Delta N_{i,l}^{TTI}$ may be computed by counting repetitions or puncturing when the algorithm of subclause 4.2.7.5 is run. The resulting values of $\Delta N_{i,l}^{TTI}$ can be represented with following expression.

$$\Delta N_{i,l}^{TTI} = \left[\frac{|\Delta N_i| \times X_i}{N_{max}}\right] \times \operatorname{sgn}(\Delta N_i) \quad \Delta N_{i,l}^{TTI} = \left[\frac{|\Delta N_{max}| \times X_i}{N_{max}}\right] \times \operatorname{sgn}(\Delta N_{max})$$

For compressed mode by puncturing, the above formula produces $\Delta N_{i,l}^{TTI,m}$ instead of $\Delta N_{i,l}^{TTI}$.

4.2.7.2.1.4 Determination of rate matching parameters for Turbo encoded TrCHs

If repetition is to be performed on turbo encoded TrCHs, i.e.- $\Delta N_{i,max} > 0$ $\Delta N_{max} > 0$, the parameters in subclause 4.2.7.2.1.3 are used.

If puncturing is to be performed, the parameters below shall be used. Index b is used to indicate systematic (b=1), 1^{st} parity (b=2), and 2^{nd} parity bit (b=3).

a=2 when b=2

$$a=1$$
 when $b=3$

The bits indicated by b=1 shall not be punctured.

$$\Delta N_i^b = \begin{cases} \left[\Delta N_{i,max} / 2 \right], & \text{for } b = 2\\ \left[\Delta N_{i,max} / 2 \right], & \text{for } b = 3 \end{cases} \xrightarrow{\Delta N_i} = \begin{cases} \left[\Delta N_{max} / 2 \right], & b = 2\\ \left[\Delta N_{max} / 2 \right], & b = 3 \end{cases}$$

In Compressed Mode by puncturing, the following relations are used instead of the previous ones:

$$\frac{\Delta N_{i}^{b}}{\Delta N_{i}} = \frac{\left[\Delta N_{i,\max}^{TTI,cm,m} / 2\right]}{\Delta N_{i}^{b}} + \frac{\Delta N_{i,\max}^{b}}{\Delta N_{i}} = \frac{\left[\Delta N_{i,\max}^{TTI,cm,m} / 2\right]}{\left[\Delta N_{i,\max}^{TTI,cm,m} / 2\right]} + \frac{\Delta N_{i,\max}^{b}}{\Delta N_{i}} + \frac{\left[\Delta N_{i,\max}^{TTI,cm,m} / 2\right]}{N_{max}} + \frac{\left[\Delta N_{i,\max}^{TTI,cm,m} / 2\right]}{\Delta N_{i}} + \frac{\Delta N_{i}^{b}}{\Delta N_{i}} + \frac{\Delta N_{i}^{TTI,cm,m}}{\Delta N_{i}} + \frac{\Delta N_{i}^{TTI,cm,m}}{\Delta$$

For each transmission time interval of TrCH *i* with TF *l*, the rate-matching pattern is calculated with the algorithm in subcaluse 4.2.7.5. The following parameters are used as input:

$$X_{i} = N_{il}^{III} / 3$$

$$e_{ini} = N_{max}$$

$$e_{plus} = a \times N_{max} \cdot e_{plus} = a \cdot N_{max}$$

$$e_{min\,us} = a \times \left| \Delta N_{i}^{b} \right| \cdot e_{min\,us} = a \cdot \left| \Delta N_{i} \right|$$

The values of $\Delta N_{i,l}^{TTI}$ may be computed by counting puncturing when the algorithm of subclause 4.2.7.5 is run. The resulting values of $\Delta N_{i,l}^{TTI}$ can be represented with following expression.

$$\frac{\Delta N_{i,l}^{TTI} = -\left\lfloor \frac{\left| \Delta N_i^2 \right| \times X_i}{N_{max}} + 0.5 \right\rfloor - \left\lfloor \frac{\left| \Delta N_i^3 \right| \times X_i}{N_{max}} \right\rfloor}{\Delta N_{i,l}^{TTI} = -\left\lfloor \frac{\left| \left\lfloor \Delta N_{max} / 2 \right\rfloor \right| \times X_i}{N_{max}} + 0.5 \right\rfloor - \left\lfloor \frac{\left| \left\lceil \Delta N_{max} / 2 \right\rceil \right| \times X_i}{N_{max}} \right\rfloor}$$

In the above equation, the first term of the right hand side represents the amount of puncturing for b=2 and the second term represents the amount of puncturing for b=3.

For compressed mode by puncturing, the above formula produces $\Delta N_{i,l}^{TTI,m}$ instead of $\Delta N_{i,l}^{TTI}$.

4.2.9.2 2nd insertion of DTX indication bits

The DTX indication bits inserted in this step shall be placed at the end of the radio frame. Note that the DTX will be distributed over all slots after 2nd interleaving.

The bits input to the DTX insertion block are denoted by $s_1, s_2, s_3, \ldots, s_s$, where *S* is the number of bits from TrCH multiplexing. The number of PhCHs is denoted by *P* and the number of bits in one radio frame, including DTX indication bits, for each PhCH by *R*.

In normal mode-
$$R = \frac{N_{data,*}}{P} = 15 \times \left(N_{data1} + N_{data2}\right) - \frac{N_{data,*}}{P} = \frac{N_{data,*}}{P} = 15N_{data1} + 15N_{data2}$$
, where N_{data1} and N_{data2}

are defined in [2].

For compressed mode, $N'_{data,*}$ is defined as $N'_{data,*} = P \times 15 \times (N'_{data1} + N'_{data2}) \cdot N'_{data,*} = P(15N'_{data1} + 15N'_{data2})$.

 N_{data1} and N_{data2} are the number of bits in the data fields of the slot format used for the current compressed mode, i.e. slot format A or B as defined in [2] corresponding to the Spreading Factor and the number of transmitted slots in use.

In case of compressed mode by puncturing and fixed positions, DTX shall be inserted until $N'_{data,*}$ bits, since the exact room for the gap is already reserved thanks to the earlier insertion of the p-bits. Therefore R is defined as $R = N'_{data,*} / P$.

In compressed mode by SF reduction and by higher layer scheduling, additional DTX shall be inserted if the transmission time reduction method does not exactly create a transmission gap of the desired <u>TGL TGL</u>. The number of bits available to the CCTrCH in one radio frame in compressed mode by SF reduction and by higher layer scheduling is denoted by $N_{data,*}^{cm}$ and $R = \frac{N_{data,*}^{cm}}{P}$. The exact value of $N_{data,*}^{cm}$ is dependent on the *TGL* and the transmission time

reduction method, which are signalled from higher layers. For transmission time reduction by SF/2 method in compressed mode $N_{data,*}^{cm} = \frac{N'_{data,*}}{2}$, and for other methods it can be calculated as $N_{data,*}^{cm} = N'_{data,*} - N_{TGL}$. For every transmission time reduction method- $N'_{data,*} = P \times 15 \times (N'_{data1} + N'_{data2}) \cdot \frac{N'_{data,*}}{N'_{data,*}} = P(15N'_{data1} + 15N'_{data2})$,

where N_{data1} and N_{data2} are the number of bits in the data fields of a slot for slot format A or B as defined in [2]. N_{TGL} is the number of bits that are located within the transmission gap and defined as:

$$N_{TGL} = \begin{cases} \frac{TGL}{15} \times N'_{data,*} \frac{TGL}{15} N'_{data,*}, \text{ if } N_{first} + TGL \le 15 \\ \frac{15 - N_{first}}{15} \times N'_{data,*} \frac{15 - N_{first}}{15} N'_{data,*}, \text{ in first frame if } N_{first} + TGL > 15 \\ \frac{TGL - (15 - N_{first})}{15} \times N'_{data,*} \frac{TGL - (15 - N_{first})}{15} N'_{data,*}, \text{ in second frame if } N_{first} + TGL > 15 \end{cases}$$

 N_{first} and TGL are defined in subclause 4.4.

NOTE : In compressed mode by SF/2 method DTX is also added in physical channel mapping stage (subclause 4.2.12.2). During 2nd DTX insertion the number of CCTrCH bits is kept the same as in normal mode.

The bits output from the DTX insertion block are denoted by $w_1, w_2, w_3, \ldots, w_{(PR)}$. Note that these bits are four valued in case of compressed mode by puncturing, and three valued otherwise. They are defined by the following relations:

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 $W_k = S_k$ k = 1, 2, 3, ..., S

 $w_k = \delta$ k = S+1, S+2, S+3, ..., <u>PP</u>·R

where DTX indication bits are denoted by δ . Here $S_k \in \{0,1, p\}$ and $\delta \notin \{0,1\}$.

4.2.10 Physical channel segmentation

When more than one PhCH is used, physical channel segmentation divides the bits among the different PhCHs. The bits input to the physical channel segmentation are denoted by $x_1, x_2, x_3, ..., x_X \xrightarrow{x_1, x_2, x_3, ..., x_Y}$, where \underbrace{YX} is the number of bits input to the physical channel segmentation block. The number of PhCHs is denoted by *P*.

The bits after physical channel segmentation are denoted $u_{p1}, u_{p2}, u_{p3}, \dots, u_{pU}$, where *p* is PhCH number and *U* is the number of bits in one radio frame for each PhCH, i.e. $U = (\underline{Y} - \underline{X} - N_{TGL}) / P$ for compressed mode by puncturing, and $U = \frac{X}{P} - \frac{Y}{P}$ otherwise. The relation between x_k and u_{pk} is given below.

For all modes, some bits of the input flow are mapped to each code until the number of bits on the code is 4U. For modes other than compressed mode by puncturing, all bits of the input flow are taken to be mapped to the codes. For compressed mode by puncturing, only the bits of the input flow not corresponding to bits p are taken to be mapped to the codes, each bit p is removed to ensure creation the gap required by the compressed mode, as described below.

Bits on first PhCH after physical channel segmentation:

 $u_{l, k} = x_{i, f(k)} \ k = 1, 2, ..., U$

Bits on second PhCH after physical channel segmentation:

 $u_{2, k} = x_{i, f(k+U)} k = 1, 2, ..., U$

•••

Bits on the *P*th PhCH after physical channel segmentation:

 $u_{P, k} = x_{i, f(k+(P-1) \leq -U)}$ k = 1, 2, ..., U

where *f* is such that :

- for modes other than compressed mode by puncturing, $x_{i,f(k)} = x_{i,k}$, i.e. f(k) = k, for all- $\frac{k}{k}$.
- for compressed mode by puncturing, bit $u_{1,1}$ corresponds to the bit $x_{i,k}$ with smallest index k when the bits p are not counted, bit $u_{1,2}$ corresponds to the bit $x_{i,k}$ with second smallest index k when the bits p are not counted, and so on for bits $u_{1,3}$... $u_{1,4\underline{U}}$, $u_{2,2}$, $u_{2,4\underline{U}}$, $u_{P,1}$, $u_{P,2}$, $u_{P,4\underline{U}}$,

4.2.11 2nd interleaving

The 2nd interleaving is a block interleaver with inter-column permutations. The bits input to the 2nd interleaver are denoted $u_{p,1}, u_{p,2}, u_{p,3}, \dots, u_{p,U}$ $u_{p1}, u_{p2}, u_{p3}, \dots, u_{pU}$, where *p* is PhCH number and *U* is the number of bits in one radio frame for one PhCH.

- (1) Set the number of columns $\underline{C2}C_2 = 30$. The columns are numbered 0, 1, 2, ..., $\underline{C2}C_2$ -1 from left to right.
- (2) Determine the number of rows $\underline{R2R}_2$ by finding minimum integer $\underline{R2R}_2$ such that:

 $U \leq \underline{\mathbf{R}2 \times \mathbf{C}2} \mathbf{R}_2 \mathbf{C}_2.$

(3) The bits input to the 2^{nd} interleaving are written into the <u>R2</u> × <u>C2R₂</u> + <u>C2</u> rectangular matrix row by row.

$$\begin{bmatrix} u_{p,1} & u_{p,2} & u_{p,3} & \dots & u_{p,30} \\ u_{p,31} & u_{p,32} & u_{p,33} & \dots & u_{p,60} \\ \vdots & \vdots & \vdots & & \vdots & \dots & \vdots \\ u_{p,((R2-1)\times30+1)} & u_{p,((R2-1)\times30+2)} & u_{p,((R2-1)\times30+3)} & \dots & u_{p,(R2\times30)} \end{bmatrix}$$

$$\begin{bmatrix} u_{p1} & u_{p2} & u_{p3} & \dots & u_{p30} \\ u_{p31} & u_{p32} & u_{p33} & \dots & u_{p60} \\ \vdots & \vdots & \vdots & & \vdots & \dots & \vdots \\ u_{p,((R_2-1)30+1)} & u_{p,((R_2-1)30+2)} & u_{p,((R_2-1)30+3)} & \dots & u_{p,(R_230)} \end{bmatrix}$$

(4) Perform the inter-column permutation based on the pattern $(P_2(j)) P_2(j) (j = 0, 1, ..., C_2C_2 - 1))$ that is shown in table 6, where $P_2P_2(j)$ is the original column position of the *j*-th permuted column. After permutation of the columns, the bits are denoted by y_{pk} .

$y_{p,1}$	$\mathcal{Y}_{p,(\text{R2+1})}$	$y_{p,(2 \times \text{R2+1})}$	$\dots \mathcal{Y}_{p,(29 \times \text{R2+1})}$	$\int y_{p1}$	$\mathcal{Y}_{p,(R_2+1)}$	$\mathcal{Y}_{p,(2R_2+1)}$	$\cdots \mathcal{Y}_{p,(29R_2+1)}$
$y_{p,2}$	$\mathcal{Y}_{p,(\text{R2+2})}$	$\mathcal{Y}_{p,(2\times R2+2)}$	$\cdots \mathcal{Y}_{p,(29 \times \text{R2}+2)}$	y_{p2}	$\mathcal{Y}_{p,(R_2+2)}$	$\mathcal{Y}_{p,(2R_2+2)}$	$\cdots \mathcal{Y}_{p,(29R_2+2)}$
÷			:	1			:
$\mathcal{Y}_{p,\mathrm{R2}}$	$\mathcal{Y}_{p,(2 \times \mathbb{R}^2)}$	$y_{p,(3\times R2)}$	$\dots y_{p,(30 \times R2)}$	y_{pR_2}	$\mathcal{Y}_{p,(2R_2)}$	$\mathcal{Y}_{p,(3R_2)}$	$\cdots y_{p,(30R_2)}$

(5) The output of the 2nd interleaving is the bit sequence read out column by column from the inter-column permuted $\frac{R2 \times C2R_2 \times C_2}{R_2 \times C_2}$ matrix. The output is pruned by deleting bits that were not present in the input bit sequence, i.e. bits y_{pk} that corresponds to bits u_{pk} with k>U are removed from the output. The bits after 2nd interleaving are denoted by $v_{p,1}, v_{p,2}, \dots, v_{p,U} \neq v_{p1}, v_{p2}, \dots, v_{pU}$, where v_{p_1} corresponds to the bit y_{p_2k} with smallest index k after pruning, $v_{p,2}$ to the bit y_{pk} with second smallest index k after pruning, and so on.

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Number of column C ₂ C2	Inter-column permutation pattern < P2(0), P2(1),,P2(29) >
30	 ≤{0, 20, 10, 5, 15, 25, 3, 13, 23, 8, 18, 28, 1, 11, 21, 6, 16, 26, 4, 14, 24, 19, 9, 29, 12, 2, 7, 22, 27, 17≥}

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4.2 Transport-channel coding/multiplexing

Data arrives to the coding/multiplexing unit in form of transport block sets- once every transmission time interval. The transmission time interval is transport-channel specific from the set {10 ms, 20 ms, 40 ms, 80 ms}.

8

The following coding/multiplexing steps can be identified:

- add CRC to each transport block (see subclause 4.2.1);
- transport block concatenation and code block segmentation (see subclause 4.2.2);
- _____channel coding (see subclause 4.2.3);
- radio frame equalisation (see subclause 4.2.4);
- rate matching (see subclause 4.2.7);
- insertion of discontinuous transmission (DTX) indication bits (see subclause 4.2.9);
- interleaving (two steps, see subclauses 4.2.54 and 4.2.11);
- radio frame segmentation (see subclause 4.2.6);
- multiplexing of transport channels (see subclause 4.2.8);
- physical channel segmentation (see subclause 4.2.10);
- mapping to physical channels (see subclause 4.2.12).

The coding/multiplexing steps for uplink and downlink are shown in figure 1 and figure 2 respectively.

3GPP TSG RAN Meeting #8 Düsseldorf, Germany, 21-23 June 2000

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e.g. for 3GPP use the format TP-99xxx or for SMG, use the format P-99-xxx

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4.2.7.3 Bit separation and collection in uplink

The systematic bits (excluding bits for trellis termination) of turbo encoded TrCHs shall not be punctured, however, systematic bits for trellis termination may be punctured. The systematic bits, first parity bits, and second parity bits in the bit sequence input to the rate matching block are therefore separated from each other into three sequences, one sequence containing all of the systematic bits and some systematic, first and second parity trellis termination bits; the second sequence containing all of the first parity bits and some systematic, first and second parity trellis termination bits and the third sequence containing all of the second parity bits and some systematic, first and second parity trellis termination bits. Puncturing is only applied to the parity bits and systematic bits used for trellis termination. second and third sequences.

The bit separation function is transparent for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition. The bit separation and bit collection are illustrated in figures 6 and 7.



Figure 5: Puncturing of turbo encoded TrCHs in uplink



Figure 6: Rate matching for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition in uplink.

The bit separation is dependent on the 1st interleaving and offsets are used to define the separation for different TTIs. The sequence denoted as b=1 contains all of the systematic bits and some systematic, first and second parity trellis termination bits; the sequence denoted as b=2 contains all of the first parity bits and some systematic, first and second parity trellis termination bits; the sequence denoted as b=3 contains all of the second parity bits and some systematic, first and second parity trellis termination bits. The offsets α_b for the these sequences systematic (*b*=1) and parity bits (*b* \in {2, 3}) are listed in table 4.

TTI (ms)	α1	α2	A 3
10, 40	0	1	2
20, 80	0	2	1

Table 4: TTI dependent offset needed for bit separa	ation
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The bit separation is different for different radio frames in the TTI. A second offset is therefore needed. The radio frame number for TrCH *i* is denoted by n_i . and the offset by β_{n_i} .

Table 5: Radio	frame dependent	t offset needed f	for bit separation

TTI (ms)	β_0	β 1	β2	β_3	β_4	β_5	β_6	β_7
10	0	NA	NA	NA	NA	NA	NA	NA
20	0	1	NA	NA	NA	NA	NA	NA
40	0	1	2	0	NA	NA	NA	NA
80	0	1	2	0	1	2	0	1

4.2.7.3.1 Bit separation

The bits input to the rate matching are denoted by $e_{i1}, e_{i2}, e_{i3}, \dots, e_{iN_i}$, where *i* is the TrCH number and N_i is the number of bits input to the rate matching block. Note that the transport format combination number *j* for simplicity has been left out in the bit numbering, i.e. $N_i=N_{ij}$. The bits after separation are denoted by $x_{bi1}, x_{bi2}, x_{bi3}, \dots, x_{biX_i}$. For turbo encoded TrCHs with puncturing, *b* indicates systematic, first parity, or second parity bit. the three sequences defined in section 4.2.7.3. The sequence denoted as b=1 contains all of the systematic bits and some systematic, first and second parity trellis termination bits; the sequence denoted as b=2 contains all of the first parity bits and some systematic, first and second parity trellis termination bits. For all other cases *b* is defined to be 1. X_i is the number of bits in each separated bit sequence. The relation between e_{ik} and x_{bik} is given below.

$x_{1,i,k} = e_{i,3(k-1)+1+(\alpha_1+\beta_{n_i}) \mod 3}$	$K = 1, 2, 3,, X_i$	$X_i \equiv$
$\begin{bmatrix} N_i / 3 \end{bmatrix}$ $x_{1,i \mid N / 3 \mid k} = e_{i,3 \mid N / 3 \mid k}$	$k = 1,, N_i \mod 3$	Note: When $(N_i \mod$
3) = 0 this row is not needed.		
$x_{2,i,k} = e_{i,3(k-1)+1+(\alpha_2+\beta_{n_i}) \mod 3}$	$k = 1, 2, 3,, X_i$	$X_i = \lfloor N_i / 3 \rfloor$
$x_{3,i,k} = e_{i,3(k-1)+1+(\alpha_3+\beta_{n_i}) \mod 3}$	$k = 1, 2, 3,, X_i$	$X_i = \lfloor N_i / 3 \rfloor$

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

 $x_{1,i,k} = e_{i,k}$ $k = 1, 2, 3, ..., X_i$ $X_i = N_i$

4.2.7.3.2 Bit collection

The bits x_{bik} are input to the rate matching algorithm described in section 4.2.7.5. The bits output from the rate matching algorithm are denoted $y_{bi1}, y_{bi2}, y_{bi3}, \dots, y_{biY_i}$.

Bit collection is the inverse function of the separation. The bits after collection are denoted by $z_{bi1}, z_{bi2}, z_{bi3}, ..., z_{biY_i}$. After bit collection, the bits indicated as punctured are removed and the bits are then denoted by $f_{i1}, f_{i2}, f_{i3}, ..., f_{iV_i}$, where *i* is the TrCH number and $V_i = N_{ij} + \Delta N_{ij}$. The relations between y_{bik}, z_{bik} , and f_{ik} are given below. For turbo encoded TrCHs with puncturing $(Y_i = X_i)$:

$z_{i,3(k-1)+1+(\alpha,+\beta) \mod 3} = y_{1,i,k}$	$k = 1, 2, 3,, Y_i$	
$z_{i,3 N_i/3 +k} = y_{1,i,N_i/3 +k}$	$k = 1,, N_i \mod 3$	Note: When $(N_i \mod$
(3) = 0 this row is not needed.		
$z_{i,3(k-1)+1+(\alpha_2+\beta_{n_i}) \mod 3} = y_{2,i,k}$	$k = 1, 2, 3,, Y_i$	
$z_{i,3(k-1)+1+(\alpha_3+\beta_{n_i}) \mod 3} = y_{3,i,k}$	$k = 1, 2, 3,, Y_i$	

After the bit collection, bits $z_{i,k}$ with value δ , where $\delta \underline{\boldsymbol{e}} \{0, 1\}$, are removed from the bit sequence. Bit $f_{i,1}$ corresponds to the bit $z_{i,k}$ with smallest index k after puncturing, bit $f_{i,2}$ corresponds to the bit $z_{i,k}$ with second smallest index k after puncturing, and so on.

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

 $z_{i,k} = y_{1,i,k}$ $k = 1, 2, 3, ..., Y_i$

When repetition is used, $f_{i,k}=z_{i,k}$ and $Y_i=V_i$.

When puncturing is used, $Y_i=X_i$ and bits $z_{i,k}$ with value δ , where $\delta \notin \{0, 1\}$, are removed from the bit sequence. Bit $f_{i,1}$ corresponds to the bit $z_{i,k}$ with smallest index *k* after puncturing, bit $f_{i,2}$ corresponds to the bit $z_{i,k}$ with second smallest index *k* after puncturing, and so on.

4.2.7.4 Bit separation and collection in downlink

The systematic bits (excluding bits for trellis termination) of turbo encoded TrCHs shall not be punctured. The systematic bit, first parity bit, and second parity bit in the bit sequence input to the rate matching block are therefore separated from each other. Puncturing is only applied to the parity bits and systematic bits used for trellis termination.

The systematic bits of turbo encoded TrCHs shall not be punctured, however, systematic bits for trellis termination may be punctured. The systematic bits, first parity bits and second parity bits in the bit sequence input to the rate matching block are therefore separated into three sequences, one sequence containing all of the systematic bits and some systematic, first and second parity trellis termination bits; the second sequence containing all of the first parity bits and some systematic, first and second parity trellis termination bits; the second parity trellis termination bits and the third sequence containing all of the second parity bits and some systematic, first and second parity trellis termination bits. Puncturing is only applied to the second and third sequences.

The bit separation function is transparent for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition. The bit separation and bit collection are illustrated in figures 8 and 9.

Figure 7: Puncturing of turbo encoded TrCHs in downlink.

Figure 8: Rate matching for uncoded TrCHs, convolutionally encoded TrCHs, and for turbo encoded TrCHs with repetition in downlink.

4.2.7.4.1 Bit separation

The bits input to the rate matching are denoted by $c_{i1}, c_{i2}, c_{i3}, \dots, c_{iE_i}$, where *i* is the TrCH number and E_i is the number of bits input to the rate matching block. Note that E_i is a multiple of 3 for turbo encoded TrCHs and that the transport format combination number *j* for simplicity has been left out in the bit numbering, i.e. $E_i = N_{ij}$. The bits after separation are denoted by $x_{bi1}, x_{bi2}, x_{bi3}, \dots, x_{biX_i}$. For turbo encoded TrCHs with puncturing, *b*

indicates systematic, first parity, or second parity bit the three sequences defined in section 4.2.7.4. For all other cases *b* is defined to be 1. X_i is the number of bits in each separated bit sequence. The relation between c_{ik} and x_{bik} is given below. For turbo encoded TrCHs with puncturing:

$x_{1,i,k} = c_{i,3(k-1)+1}$	$k = 1, 2, 3,, X_i$	$X_i =$
$E_i/3$	k = 1, 2, 3, X	$Y_{.} - F_{.}/3$
$x_{2,i,k} = c_{i,3(k-1)+2}$ $x_{2,i,k} = c_{i,3(k-1)+2}$	$k = 1, 2, 3, \dots, A_l$ k = 1, 2, 3, X	$X_i = E_i / 3$ $X_i = F_i / 3$
$x_{3,i,k} - c_{i,3(k-1)+3}$	$\kappa = 1, 2, 3,, n_l$	$A_l = L_l / S$

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

$$x_{1,i,k} = c_{i,k}$$
 $k = 1, 2, 3, ..., X_i$ $X_i = E_i$

e.g. for 3GPP use the format TP-99xxx or for SMG, use the format P-99-xxx

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4.2.2.2 Code block segmentation

Segmentation of the bit sequence from transport block concatenation is performed if $X_i > Z$. The code blocks after segmentation are of the same size. The number of code blocks on TrCH *i* is denoted by C_i . If the number of bits input to the segmentation, X_i , is not a multiple of C_i , filler bits are added to the beginning of the first block. If turbo coding is selected and $X_i < 40$, filler bits are added to the beginning of the code block. The filler bits are transmitted and they are always set to 0. The maximum code block sizes are:

- convolutional coding: Z = 504;
- turbo coding: Z = 5114;
- no channel coding: *Z* = *unlimited*.

The bits output from code block segmentation are denoted by $o_{ir1}, o_{ir2}, o_{ir3}, \dots, o_{irK_i}$, where *i* is the TrCH number, *r* is the code block number, and K_i is the number of bits.

Number of code blocks: $C_i = [X_i / Z]$

Number of bits in each code block:

if $X_i < 40$ and Turbo coding is used, then

$$K_i = 40$$

else

$$K_i = \left[X_i / C_i \right]$$

end if

Number of filler bits: $Y_i = C_i K_i - X_i$

 $\underline{\text{If } X_i \leq Z, \text{ then}}$

 $o_{i1k} = 0 \quad k = 1, 2, ..., Y_i$

for k = 1 to Y_i -- Insertion of filler bits

 $_{i1k} = 0$

end for

 $o_{i1k} = x_{i,(k-Y_i)}$, $k = Y_i + 1, Y_i + 2, ..., K_i$ for $k = Y_i + 1$ to K_i

 $_o_{i1k} = x_{i,(k-Y_i)}$

end for

end if

```
If X_t > Z, then
```

 $\theta_{i1k} = 0 - k = 1, 2, ..., Y_i$

 $o_{i1k} = x_{i,(k-Y_i)} - k = Y_i + 1, Y_i + 2, \dots, K_i$

<u>r = 2</u> -- Segmentation

while $r \leq C_i$

for k = 1 to K_i

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4.3.3 Coding of Transport-Format-Combination Indicator (TFCI)

The TFCI bits are is encoded using a (32, 10) sub-code of the second order Reed-Muller code. The coding procedure is as shown in figure 9.

Figure 9: Channel coding of TFCI information bits

If the TFCI consist of less than 10 bits, it is padded with zeros to 10 bits, by setting the most significant bits to zero. The length of the TFCI code word is 32 bits.

The code words of the (32,10) sub-code of second order Reed-Muller code are linear combination of 10 basis sequences. The basis sequences are as in the following table 7.

i	M i,0	M i,1	M _{i,2}	M _{i,3}	M i,4	M _{i,5}	M i,6	M _{i,7}	M _{i,8}	M _{i,9}
0	1	0	0	0	0	1	0	0	0	0
1	0	1	0	0	0	1	1	0	0	0
2	1	1	0	0	0	1	0	0	0	1
3	0	0	1	0	0	1	1	0	1	1
4	1	0	1	0	0	1	0	0	0	1
5	0	1	1	0	0	1	0	0	1	0
6	1	1	1	0	0	1	0	1	0	0
7	0	0	0	1	0	1	0	1	1	0
8	1	0	0	1	0	1	1	1	1	0
9	0	1	0	1	0	1	1	0	1	1
10	1	1	0	1	0	1	0	0	1	1
11	0	0	1	1	0	1	0	1	1	0
12	1	0	1	1	0	1	0	1	0	1
13	0	1	1	1	0	1	1	0	0	1
14	1	1	1	1	0	1	1	1	1	1
15	1	0	0	0	1	1	1	1	0	0
16	0	1	0	0	1	1	1	1	0	1
17	1	1	0	0	1	1	1	0	1	0
18	0	0	1	0	1	1	0	1	1	1
19	1	0	1	0	1	1	0	1	0	1
20	0	1	1	0	1	1	0	0	1	1
21	1	1	1	0	1	1	0	1	1	1
22	0	0	0	1	1	1	0	1	0	0
23	1	0	0	1	1	1	1	1	0	1
24	0	1	0	1	1	1	1	0	1	0
25	1	1	0	1	1	1	1	0	0	1
26	0	0	1	1	1	1	0	0	1	0
27	1	0	1	1	1	1	1	1	0	0
28	0	1	1	1	1	1	1	1	1	0
29	1	1	1	1	1	1	1	1	1	1
30	0	0	0	0	0	1	0	0	0	0
31	0	0	0	0	1	1	1	0	0	0

Table 7: Basis sequences for (32,10) TFCI code

Let's define the TFCI information bits as a_0 , a_1 , a_2 , a_3 , a_4 , a_5 , a_6 , a_7 , a_8 , a_9 (a_0 is LSB and a_9 is MSB). The TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the CCTrCH in the associated DPCH radio frame.

The output code word bits b_i are given by:

$$b_i = \sum_{n=0}^{9} (a_n \times M_{i,n}) \mod 2$$

where $\frac{1}{10} = 0, \dots, 31$.

The output bits are denoted by b_k , k = 0, 1, 2, ..., 31.

In downlink, when the SF < 128 the encoded TFCI code words are repeated yielding 8 encoded TFCI bits per slot in normal mode and 16 encoded TFCI bits per slot in compressed mode. Mapping of repeated bits to slots is explained in subclause 4.3.5.

4.3.4 Operation of Transport-Format-Combination Indicator (TFCI) in Split Mode

If one of the DCH is associated with a DSCH, the TFCI code word may be split in such a way that the code word relevant for TFCI activity indication is not transmitted from every cell. The use of such a functionality shall be indicated by higher layer signalling.

The TFCI bits are is encoded using a (16, 5) bi-orthogonal (or first order Reed-Muller) code. The coding procedure is as shown in figure 10.

Figure 10: Channel coding of split mode TFCI information bits

The code words of the (16,5) bi-orthogonal code are linear combinations of 5 basis sequences as defined in table 8.

i	M i,0	M i,1	M i,2	M i,3	M i,4
0	1	0	0	0	1
1	0	1	0	0	1
2	1	1	0	0	1
3	0	0	1	0	1
4	1	0	1	0	1
5	0	1	1	0	1
6	1	1	1	0	1
7	0	0	0	1	1
8	1	0	0	1	1
9	0	1	0	1	1
10	1	1	0	1	1
11	0	0	1	1	1
12	1	0	1	1	1
13	0	1	1	1	1
14	1	1	1	1	1
15	0	0	0	0	1

Table 8: Basis sequences for (16,5) TFCI code

Let's define a first set of -TFCI information bits as $a_{1,0}$, $a_{1,1}$, $a_{1,2}$, $a_{1,3}$, $a_{1,4}$ ($a_{1,0}$ is LSB and $a_{1,4}$ is MSB). This set of TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the DCH CCTrCH in the associated DPCH radio frame.

Let's define a second set of -TFCI information bits as $a_{2,0}$, $a_{2,1}$, $a_{2,2}$, $a_{2,3}$, $a_{2,4}$ ($a_{2,0}$ is LSB and $a_{2,4}$ is MSB). This set of TFCI information bits shall correspond to the TFC index (expressed in unsigned binary form) defined by the RRC layer to reference the TFC of the associated DSCH CCTrCH in the corresponding PDSCH radio frame.

The output code word bits b_k are given by:

$$b_{2i} = \sum_{n=0}^{4} (a_{1,n} \times M_{i,n}) \mod 2;$$

$$---b_{2i+1} = \sum_{n=0}^{4} (a_{2,n} \times M_{i,n}) \mod 2$$

where $\frac{i}{i} = 0, \dots, 15, \frac{j}{j} = 0, 1$.

The output bits are denoted by b_k , k = 0, 1, 2, ..., 31.

4.3.5 Mapping of TFCI words

4.3.5.1 Mapping of TFCI word in non compressed<u>normal</u> mode

The bits of the code word are directly mapped to the slots of the radio frame. Within a slot the bit with lower index -is transmitted before the bit with higher index. The coded bits b_k , are mapped to the transmitted TFCI bits d_k , according to the following formula:

 $\mathbf{d}_{\mathbf{k}}\underline{d}_{\underline{k}} = \mathbf{b}_{\mathbf{k}}\underline{b}_{\underline{k}} \mod 32$

For uplink physical channels regardless of the SF and downlink physical channels, if $-SF \ge 128$, k = 0, 1, 2, ..., 29.

NOTE: <u>Note that t</u>This means that bits b_{30} and b_{31} are not transmitted.

For downlink physical channels whose SF_<_128, k = 0, 1, 2, ..., 119. Note that this means that bits b_0 to b_{23} are transmitted four times and bits b_{24} to b_{31} are transmitted three times.

4.3.5.2 Mapping of TFCI word in compressed mode

The mapping of the TFCI bits in compressed mode is different for uplink, downlink with $SF \ge 128$ and downlink with $SF \le 128$.

4.3.5.2.1 Uplink compressed mode

For uplink compressed mode, the slot format is changed so that no TFCI <u>coded</u> bits are lost. The different slot formats in compressed mode do not match the exact number of TFCI <u>coded</u> bits for all possible TGLs. Repetition of the TFCI bits is therefore used.

Denote the number of bits available in the TFCI fields of one compressed radio frame by D and the number of bits in the TFCI field in a slot by N_{TFCI} . Denote by E the first bit to be repeated, The parameter E is used to determine the number of the first TFCI bit to be repeated.

 $-E_{=}N_{first}N_{TFCI}N_{first}N_{TFCI}$, if the start of the transmission gap is allocated to the current frame. E = 0, if the start of the transmission gap is allocated to the previous frame and the end of the transmission gap is allocated to the current frame.

: If $N_{tast} \neq 14$, then *E* corresponds to the number of the first TFCI bit in the slot directly after the TG. The TFCI coded bits <u> b_k are mapped to the bits in the TFCI fields d_k </u>. The following relations then define the mapping for each compressed frame.

 $\mathbf{d}_{\mathbf{k}}\underline{d}_{\underline{k}} = \mathbf{b}_{\mathbf{k}}\underline{b}_{\underline{k} \mod 32}$

where $k = 0, 1, 2, ..., \min(31, \frac{DD}{2}-1)$.

If $\frac{D}{D} > 32$, the remaining positions are filled by repetition (in reversed order):

 $\mathbf{d}_{\mathbf{D}-\mathbf{k}-1}\underline{d}_{\underline{D}-\underline{k}-1} = \mathbf{b}_{(\underline{\mathbf{E}}+\underline{\mathbf{k}})}\underline{b}_{(\underline{E}+\underline{k})} \mod 32$

where $k = 0, \dots, \frac{\mathbf{D}\mathbf{D}}{\mathbf{D}}$ -33.

4.3.5.2.2 Downlink compressed mode

For downlink compressed mode, the slot format is changed so that no TFCI <u>coded</u> bits are lost. The different slot formats in compressed mode do not match the exact number of TFCI bits for all possible TGLs. DTX is therefore used if the number of <u>bits available in the</u> TFCI fields <u>in one compressed frame</u> exceeds the number of TFCI bits <u>given from</u> the slot format. The block of <u>bits in the TFCI</u> fields, where DTX is used, starts on the first <u>TFCI</u> field after the transmission gap. If there are fewermore bits available in the TFCI fields afterbefore the transmission gap than DTXTFCI bits, DTX is also used on the bits in the last <u>TFCI</u> fields before of the transmission gap are also filled with DTX.

Denote the number of bits available in the TFCI fields of one compressed radio frame by D and the number of bits in the TFCI field in a slot by N_{TFCI} . Denote by E the first bit to be repeated The parameter E is used to determine the position of the first bit in the TFCI field on which DTX is used.

 $E = N_{first}N_{TFCI}$, if $N_{first} + TGL \le 15$, else E = 0

<u> $E = N_{\text{first}} N_{\text{TFCI}}$ </u>, if the start of the transmission gap is allocated to the current frame. <u>E = 0</u>, if the start of the transmission gap is allocated to the previous frame and the end of the transmission gap is allocated to the current frame.

If the transmission gap does not extend to the end of the frame then *E* corresponds to the number of the first TFCI bit in the slot directly after the TG. Denote the total number of TFCI bits to be transmitted by $N_{tot}F$. If SF \ge 128 then $N_{tot}F$ = 32 for slot formats *n*A or *n*B, where *n* = 0, 1, ..., 11 (see table 11 in [2]). Otherwise, else $N_{tot}F$ = 128. The TFCI coded bits b_k are mapped to the bits in the TFCI fields d_k . The following relations then define the mapping for each compressed frame. \ddagger

If E > 0,

 $\mathbf{d}_{\mathbf{k}}\underline{d}_{\underline{k}} = \mathbf{b}_{(\mathbf{k}}\underline{b}_{\underline{k} \mod 32)}$

where $k = 0, 1, 2, ..., \min(\underline{EE}, N_{tot}F) - 1_{\underline{.}}$ -and,

 $\underline{\text{Iif }} \underline{E} \underline{E} < N_{tot} \underline{F},$

 $\mathbf{d}_{\mathbf{k}+\mathbf{D}-\mathbf{Ntot}}\underline{d}_{k+\mathbf{D}-F} = \mathbf{b}_{(\mathbf{k}}\underline{b}_{k \mod 32)}$

where $k = \underbrace{\mathbf{E}}_{\underline{E}}, ..., \underbrace{\mathbf{N}_{tot}}_{\underline{F}} - 1$.

DTX bits are sent is used on d_k where $k = \min(\underbrace{\mathbb{E}E}, \underbrace{N_{tot}E}), ..., \min(\underbrace{\mathbb{E}E}, \underbrace{N_{tot}E}) + \underbrace{\mathbb{D}D} - \underbrace{N_{tot}E} - 1$.

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4.2.7 Rate matching

Rate matching means that bits on a transport channel are repeated or punctured. Higher layers assign a rate-matching attribute for each transport channel. This attribute is semi-static and can only be changed through higher layer signalling. The rate-matching attribute is used when the number of bits to be repeated or punctured is calculated.

The number of bits on a transport channel can vary between different transmission time intervals. In the downlink the transmission is interrupted if the number of bits is lower than maximum. When the number of bits between different transmission time intervals in uplink is changed, bits are repeated or punctured to ensure that the total bit rate after TrCH multiplexing is identical to the total channel bit rate of the allocated dedicated physical channels.

If no bits are input to the rate matching for all TrCHs within a CCTrCH, the rate matching shall output no bits for all TrCHs within the CCTrCH and no uplink DPDCH will be selected in the case of uplink rate matching.

Notation used in subcaluse 4.2.7 and subclauses:

 N_{ij} : For uplink: Number of bits in a radio frame before rate matching on TrCH *i* with transport format combination *j*.

For downlink: An intermediate calculation variable (not an integer but a multiple of 1/8).

- N_{il}^{TTI} : Number of bits in a transmission time interval before rate matching on TrCH *i* with transport format *l*. Used in downlink only.
- ΔN_{ij} : For uplink: If positive number of bits that should be repeated in each radio frame on TrCH *i* with transport format combination *j*.

If negative - number of bits that should be punctured in each radio frame on TrCH i with transport format combination j.

For downlink : An intermediate calculation variable (not an integer but a multiple of 1/8).

 ΔN_{il}^{TTI} : If positive - number of bits to be repeated in each transmission time interval on TrCH *i* with transport format *j*.

If negative - number of bits to be punctured in each transmission time interval on TrCH i with transport format j.

Used in downlink only.

 $Np^{TTI, m}_{i,b}$ m=0 to $F_{max}/F_i - 1$: Positive or null: number of bits to be removed in TTI number *m* within the largest TTI, to create the required gaps in the compressed radio frames of this TTI, in case of compressed mode by puncturing, for TrCh *i* with transport format *l*. In case of fixed positions and compressed mode by puncturing, this value is noted $Np^{TTI, m}_{i,max}$ since it is calculated for all TrCh with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

 $Np^{n}_{i,l}$ n=0 to F_{max} -1:Positive or null: number of bits, in radio frame number n within the largest TTI, corresponding to the gap for compressed mode in this radio frame, for TrCH i with transport format l. The value will be null for the un-compressed radio frames. In case of fixed positions and compressed mode by puncturing, this value is noted $Np^{n}_{i,max}$ since it is calculated for all TrChs with their maximum number of bits; thus it is the same for all TFCs

Used in downlink only.

- $N_{TGL}[k]$, k=0 to $F_i 1$: Positive or null: number of bits in each radio frame corresponding to the gap for compressed mode for the CCTrCh.
- RM_i : Semi-static rate matching attribute for transport channel *i*. Signalled from RM_i is provided by higher layers or takes a value as indicated in section 4.2.13.

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- *PL:* Puncturing limit for uplink. This value limits the amount of puncturing that can be applied in order to avoid multicode or to enable the use of a higher spreading factor. Signalled from higher layers.
- $N_{data,j}$: Total number of bits that are available for the CCTrCH in a radio frame with transport format combination *j*.
- *I:* Number of TrCHs in the CCTrCH.
- Z_{ij} : Intermediate calculation variable.
- F_i : Number of radio frames in the transmission time interval of TrCH *i*.
- F_{max} Maximum number of radio frames in a transmission time interval used in the CCTrCH :

$$F_{\max} = \max_{1 \le i \le I} F_i$$

- n_i : Radio frame number in the transmission time interval of TrCH i ($0 \le n_i < F_i$).
- *q:* Average puncturing or repetition distance (normalised to only show the remaining rate matching on top of an integer number of repetitions). Used in uplink only.
- $I_F(n_i)$: The inverse interleaving function of the 1st interleaver (note that the inverse interleaving function is identical to the interleaving function itself for the 1st interleaver). Used in uplink only.
- $S(n_i)$: The shift of the puncturing or repetition pattern for radio frame n_i . Used in uplink only.
- $TF_i(j)$: Transport format of TrCH *i* for the transport format combination *j*.
- TFS(i) The set of transport format indexes *l* for TrCH *i*.
- *TFCS* The set of transport format combination indexes *j*.
- *e*_{ini} Initial value of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- e_{plus} Increment of variable *e* in the rate matching pattern determination algorithm of subclause4.2.7.5.
- e_{minus} Decrement of variable *e* in the rate matching pattern determination algorithm of subclause 4.2.7.5.
- *b:* Indicates systematic and parity bits

b=1: Systematic bit. X(t) in subclause 4.2.3.2.1.

b=2: 1st parity bit (from the upper Turbo constituent encoder). Y(t) in subcaluse 4.2.3.2.1.

b=3: 2^{nd} parity bit (from the lower Turbo constituent encoder). *Y*'(*t*) in subclause 4.2.3.2.1.

The * (star) notation is used to replace an index x when the indexed variable X_x does not depend on the index x. In the left wing of an assignment the meaning is that " $X_* = Y$ " is equivalent to "**for all** <u>x</u> **do** $X_x = Y$ ". In the right wing of an assignment, the meaning is that " $Y = X_*$ " is equivalent to "**take any** <u>x</u> **and do** $Y = X_x$ ".

The following relations, defined for all TFC j, are used when calculating the rate matching parameters:

$$Z_{0,i} = 0$$

$$Z_{ij} = \left[\frac{\left\{ \left[\sum_{m=1}^{i} RM_{m} N_{mj} \right] N_{data, j} \right\}}{\sum_{m=1}^{l} RM_{m} N_{mj}} \right] \text{ for all } i = 1 \dots I$$
 (1)

$$\Delta N_{ij} = Z_{ij} - Z_{i-1,j} - N_{ij} \qquad \text{for all } i = 1 \dots I$$

4.2.13 Restrictions on different types of CCTrCHs

Restrictions on the different types of CCTrCHs are described in general terms in TS 25.302[11]. In this subclause those restrictions are given with layer 1 notation.

4.2.13.1 Uplink Dedicated channel (DCH)

The maximum value of the number of TrCHs I in a CCTrCH, the maximum value of the number of transport blocks M_i on each transport channel, and the maximum value of the number of DPDCHs P are given from the UE capability class.

4.2.13.2 Random Access Channel (RACH)

- There can only be one TrCH in each RACH CCTrCH, i.e. I=1, $s_k = f_{1k}$ and $S = V_1$.
- The maximum value of the number of transport blocks M_1 on the transport channel is given from the UE capability class.
- The transmission time interval is either 10 ms or 20 ms.
- Only one PRACH is used, i.e. P=1, $u_{1k} = s_k$, and U = S.
- The Static rate matching parameter RM₁ is not provided by higher layer signalling on the System information as the other transport channel parameters. Any value may be used as there is one transport channel in the CCTrCH, hence one transport channel per Transport Format Combination and no need to do any balancing between multiple transport channels.

4.2.13.3 Common Packet Channel (CPCH)

- The maximum value of the number of TrCHs I in a CCTrCH, the maximum value of the number of transport blocks M_i on each transport channel, and the maximum value of the number of DPDCHs P are given from the UE capability class.
- NOTE: Only the data part of the CPCH can be mapped on multiple physical channels (this note is taken from TS 25.302).

4.2.13.4 Downlink Dedicated Channel (DCH)

The maximum value of the number of TrCHs I in a CCTrCH, the maximum value of the number of transport blocks M_i on each transport channel, and the maximum value of the number of DPDCHs P are given from the UE capability class.

4.2.13.5 Downlink Shared Channel (DSCH) associated with a DCH

- The spreading factor is indicated with the TFCI or with higher layer signalling on DCH.
- The maximum value of the number of TrCHs I in a CCTrCH, the maximum value of the number of transport blocks M_I on the transport channel and the maximum value of the number of PDSCHs P are given from the UE capability class.

4.2.13.6 Broadcast channel (BCH)

- There can only be one TrCH in the BCH CCTrCH, i.e. I=1, $s_k = f_{1k}$, and $S = V_1$.
- There can only be one transport block in each transmission time interval, i.e. $M_1 = 1$.
- All transport format attributes have predefined values. which are provided in [11] apart from the rate matching RM₁.

- The Static rate matching parameter *RM*₁ is not provided by higher layer signalling neither fixed. Any value may be used as there is one transport channel in the CCTrCH, hence one transport channel per Transport Format Combination and no need to do any balancing between multiple transport channels.
- Only one primary CCPCH is used, i.e. *P*=1.

4.2.13.7 Forward access and paging channels (FACH and PCH)

- The maximum value of the number of TrCHs I in a CCTrCH and the maximum value of the number of transport blocks M_i on each transport channel are given from the UE capability class.
- The transmission time interval for TrCHs of PCH type is always 10 ms.
- Only one secondary CCPCH is used per CCTrCH, i.e. P=1.

4.2.14 Multiplexing of different transport channels into one CCTrCH, and mapping of one CCTrCH onto physical channels

The following rules shall apply to the different transport channels which are part of the same CCTrCH:

 Transport channels multiplexed into one CCTrCh shall have co-ordinated timings. When the TFCS of a CCTrCH is changed because one or more transport channels are added to the CCTrCH or reconfigured within the CCTrCH, or removed from the CCTrCH, the change may only be made at the start of a radio frame with CFN fulfilling the relation

CFN mod $F_{max} = 0$,

where F_{max} denotes the maximum number of radio frames within the transmission time intervals of all transport channels which are multiplexed into the same CCTrCH, including any transport channels *i* which are added, reconfigured or have been removed, and CFN denotes the connection frame number of the first radio frame of the changed CCTrCH.

After addition or reconfiguration of a transport channel *i* within a CCTrCH, the TTI of transport channel *i* may only start in radio frames with CFN fulfilling the relation:

 $CFN_i \mod F_i = 0.$

- 2) Only transport channels with the same active set can be mapped onto the same CCTrCH.
- 3) Different CCTrCHs cannot be mapped onto the same PhCH.
- 4) One CCTrCH shall be mapped onto one or several PhCHs. These physical channels shall all have the same SF.
- 5) Dedicated Transport channels and common transport channels cannot be multiplexed into the same CCTrCH.
- 6) For the common transport channels, only the FACH and PCH may belong to the same CCTrCH.

There are hence two types of CCTrCH:

- 1) CCTrCH of dedicated type, corresponding to the result of coding and multiplexing of one or several DCHs.
- 2) CCTrCH of common type, corresponding to the result of the coding and multiplexing of a common channel, RACH in the uplink, DSCH ,BCH, or FACH/PCH for the downlink.

4.2.14.1 Allowed CCTrCH combinations for one UE

4.2.14.1.1 Allowed CCTrCH combinations on the uplink

A maximum of one CCTrCH is allowed for one UE on the uplink. It can be either:

- 1) one CCTrCH of dedicated type;
- 2) one CCTrCH of common type.

4.2.14.1.2 Allowed CCTrCH combinations on the downlink

The following CCTrCH combinations for one UE are allowed:

- x CCTrCH of dedicated type + y CCTrCH of common typeThe allowed combination of CCTrCHs of dedicated and common type are given from UE radio access capabilities. There can be a maximum on one CCTrCH of common type for DSCH and a maximum of one CCTrCH of common type for FACH. With one CCTrCH of common type for DSCH, there shall be at least one CCTrCH of dedicated type.
- NOTE 1: There is only one DPCCH in the uplink, hence one TPC bits flow on the uplink to control possibly the different DPDCHs on the downlink, part of the same or several CCTrCHs.
- NOTE 2: There is only one DPCCH in the downlink, even with multiple CCTrCHs. With multiple CCTrCHs, the DPCCH is transmitted on one of the physical channels of that CCTrCH which has the smallest SF among the multiple CCTrCHs. Thus there is only one TPC command flow and only one TFCI word in downlink even with multiple CCTrCHs.

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4.3.1 Blind transport format detection

When no TFCI is available then explicit blind detection or guided detection shall be performed on all TrCHs within the CCTrCH that have more than one transport format. The UE shall only be required to support blind transport format detection if all of the following restrictions are fulfilled:

- 1. only one CCTrCH is received by the UE.
- 2. <u>1.</u> the number of CCTrCH bits received per radio frame is 600 or less;
- <u>3.</u> <u>2.</u> the number of transport format combinations of the CCTrCH is 64 or less;
- <u>4.</u> <u>3.</u> fixed positions of the transport channels is used on the CCTrCH to be detected;
- 5. 4. convolutional coding is used on all explicitly detected TrCHs;
- 6. <u>5.</u> CRC is appended to all transport blocks on all <u>explicitely</u> detected TrCHs;
- 7. 6.-the number of explicitly explicitly detected TrCHs is 3 or less;
- 8. 7. for all explicitly detected TrCHs *i*, the number of code blocks in one TTI (C_i) shall not exceed 1;
- 9. 8. the sum of the transport format set sizes of all <u>explicitly</u> detected TrCHs, is 16 or less. The transport format set size is defined as the number of transport formats within the transport format set;
- <u>10.</u> 9. there is at least one TrCH that can be used as the guiding transport channel for all transport channels using guided detection.

Examples of blind transport format detection methods are given in annex A.

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4.2.7.2.2 Determination of rate matching parameters for flexible positions of TrCHs

4.2.7.2.2.1 Calculations for normal mode, compressed mode by higher layer scheduling, and compressed mode by spreading factor reduction

First an intermediate calculation variable N_{ij} is calculated for all transport channels *i* and all transport format combinations *j* by the following formula:

$$N_{i,j} = \frac{1}{F_i} \cdot N_{i,TF_i(j)}^{TTI}$$

Then rate matching ratios RF_i are calculated for each the transport channel *i* in order to minimise the number of DTX bits when the bit rate of the CCTrCH is maximum. The RF_i ratios are defined by the following formula:

$$RF_{i} = \frac{N_{data,*}}{\max_{j \in TFCS} \sum_{i=1}^{i=I} (RM_{i} \cdot N_{i,j})} \cdot RM_{i}$$

The computation of $\Delta N_{i,l}^{TTI}$ parameters is then performed in two phases. In a first phase, tentative temporary values of $\Delta N_{i,l}^{TTI}$ are computed, and in the second phase they are checked and corrected. The first phase, by use of the *RF_i* ratios, ensures that the number of DTX indication bits inserted is minimum when the CCTrCH bit rate is maximum, but it does not ensure that the maximum CCTrCH bit rate is not greater than $N_{data,*}$. per 10ms. The latter condition is ensured through the checking and possible corrections carried out in the second phase.

At the end of the second phase, the latest value of $\Delta N_{i,l}^{TTI}$ is the definitive value.

The first phase defines the tentative temporary $\Delta N_{i,l}^{TTI}$ for all transport channel *i* and any of its transport format *l* by use of the following formula:

$$\Delta N_{i,l}^{TTI} = F_i \cdot \left[\frac{RF_i \cdot N_{i,l}^{TTI}}{F_i} \right] - N_{i,l}^{TTI} = F_i \cdot \left[\frac{N_{data,*} \cdot RM_i \cdot N_{i,l}^{TTI}}{F_i \cdot \max_{j \in TFCS} \sum_{i=1}^{l} (RM_i \cdot N_{i,j})} \right] - N_{i,l}^{TTI}$$

The second phase is defined by the following algorithm:

for all *j* in *TFCS* in ascending order of <u>TFCI</u> do- -- for all TFC

$$D = \sum_{i=1}^{i=1} \frac{N_{i,TF_i(j)}^{TTI} + \Delta N_{i,TF_i(j)}^{TTI}}{F_i} \qquad -- \text{ CCTrCH bit rate (bits per 10ms) for TFC } l$$

if $D > N_{data,*}$ then

for i = 1 to I do

-- for all TrCH

$$\Delta N = F_i \cdot \Delta N_{i,j} \qquad \qquad \text{--} \ \Delta N_{i,j} \text{ is derived from } N_{i,j} \text{ by the formula given at subclause 4.2.7}$$

if $\Delta N_{i,TF_i(j)}^{TTI} > \Delta N$ then

$$\Delta N_{i,TF_i(j)}^{TTI} = \Delta N$$

end-if

end-for
end-if

end-for

NOTE: The order in which the transport format combinations are checked does not change the final result.

If $\Delta N_{i,l}^{TTI} = 0$ then, for TrCH *i* at TF *l*, the output data of the rate matching is the same as the input data and the rate matching algorithm of subclause 4.2.7.5 does not need to be executed.

If $\Delta N_{i,l}^{TTI} \neq 0$ the parameters listed in subclauses 4.2.7.2.2.2 and 4.2.7.2.2.3 shall be used for determining e_{ini} , e_{plus} , and e_{minus} .

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4.2.3.2.3.2 Intra-row and inter-row permutations

After the bits-input to the $R \times C$ rectangular matrix, the intra-row and inter-row permutations for the $R \times C$ rectangular matrix are performed by using the following algorithm.

(1) Select a primitive root v from table 2.

(2) Construct the base sequence s(i) for intra-row permutation as: $s(i) = [v \times s(i-1)] \mod p$, i = 1, 2, ..., (p-2), and s(0) = 1.

(3) Let $q_0 = 1$ be the first prime integer in $\{q_j\}$, and select the consecutive minimum prime integers $\{q_j\}$ (j = 1, 2, ..., R- 1) such that:

g.c.d{ q_j , p - 1} = 1, $q_j > 6$, and $q_j > q_{(j-1)}$, where g.c.d. is greatest common divisor.

(4) Permute $\{q_j\}$ to make $\{r_j\}$ such that $r_{T(j)} = q_j, \ j = 0, 1, \ \dots, R - 1,$

where T(j) indicates the original row position of the *j* th permuted row, and T(j) (j = 0, 1, 2, ..., R - 1) is the inter-row permutation pattern defined as the one of the following four kind of patterns: Pat_1 , Pat_2 , Pat_3 and Pat_4 depending on the number of input bits *K*.

$$\frac{T(j)}{T(j)} \cdot \frac{\{T(0), T(1), T(2), \dots, T(R-1)\}}{\{T(0), T(1), T(2), \dots, T(R-1)\}} = \begin{cases} Pat_4 & \text{if } (40 \le K \le 159) \\ Pat_3 & \text{if } (160 \le K \le 200) \\ Pat_1 & \text{if } (201 \le K \le 480) \\ Pat_3 & \text{if } (481 \le K \le 530) \\ Pat_1 & \text{if } (531 \le K \le 2280) \\ Pat_2 & \text{if } (2281 \le K \le 2480) \\ Pat_2 & \text{if } (2281 \le K \le 2480) \\ Pat_2 & \text{if } (3161 \le K \le 3210) \\ Pat_1 & \text{if } (3211 \le K \le 5114) \end{cases}$$

where *Pat*₁, *Pat*₂, *Pat*₃ and *Pat*₄ have the following patterns respectively.

*Pat*₁: {19, 9, 14, 4, 0, 2, 5, 7, 12, 18, 10, 8, 13, 17, 3, 1, 16, 6, 15, 11} *Pat*₂: {19, 9, 14, 4, 0, 2, 5, 7, 12, 18, 16, 13, 17, 15, 3, 1, 6, 11, 8, 10}

*Pat*₃: {9, 8, 7, 6, 5, 4, 3, 2, 1, 0}

*Pat*₄: {4, 3, 2, 1, 0}

(5) Perform the *j*-th (j = 0, 1, 2, ..., R - 1) intra-row permutation as: if (C = p) then $U_i(i) = s([i \times r_i] \mod(p - 1)), i = 0, 1, 2, ..., (p - 2), and U_i(p - 1) = 0,$

where $U_i(i)$ is the input bit position of *i*-th output after the permutation of *j*-th row.

end if

if (C = p + 1) then

 $U_i(i) = s([i \times r_i] \mod(p-1)), i = 0, 1, 2, ..., (p-2), U_i(p-1) = 0, and U_i(p) = p,$

where $U_i(i)$ is the input bit position of *i*-th output after the permutation of *j*-th row, and

if $(K = C \times R)$ then

Exchange $U_{R-1}(p)$ with $U_{R-1}(0)$.

end if

end if

if (C = p - 1) then

 $U_i(i) = s([i \times r_i] \mod(p-1)) - 1, \quad i = 0, 1, 2, ..., (p-2),$

where $U_i(i)$ is the input bit position of *i*-th output after the permutation of *j*-th row.

end if

(6) Perform the inter-row permutation based on the pattern T(j) (j = 0, 1, 2, ..., R - 1), where T(j) is the original row position of the *j*-th permuted row.

р	v	р	v	р	v	р	V	Р	v
7	3	47	5	101	2	157	5	223	3
11	2	53	2	103	5	163	2	227	2
13	2	59	2	107	2	167	5	229	6
17	3	61	2	109	6	173	2	233	3
19	2	67	2	113	3	179	2	239	7
23	5	71	7	127	3	181	2	241	7
29	2	73	5	131	2	191	19	251	6
31	3	79	3	137	3	193	5	257	3
37	2	83	2	139	2	197	2		
41	6	89	3	149	2	199	3		
43	3	97	5	151	6	211	2		

Table 2: Table of prime *p* and associated primitive root *v*

4.2.3.2.3.3

Bits-output from rectangular matrix with pruning

After intra-row and inter-row permutations, the bits of the permuted rectangular matrix are denoted by y'k:

The output of the Turbo code internal interleaver is the bit sequence read out column by column from the intra-row and inter-row permuted $R \times C$ matrix starting with bit y'_1 in row 0 of column 0 and ending with bit y'_{CR} in row R - 1 of column C - 1. The output is pruned by deleting bits that were not present in the input bit sequence, i.e. bits y'_k that corresponds to bits x_k with k > K are removed from the output. The bits output from Turbo code internal interleaver are denoted by x'_1, x'_2, \ldots, x'_K , where x'_1 corresponds to the bit y'_k with smallest index k after pruning, x'_2 to the bit y'_k with second smallest index k after pruning, and so on. The number of bits output from Turbo code internal interleaver is K and the total number of pruned bits is:

 $R \times C - K$.

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4.2.7.4.1 Bit separation

The bits input to the rate matching are denoted by $c_{i1}, c_{i2}, c_{i3}, \dots, c_{iE_i}$, where *i* is the TrCH number and E_i is the number of bits input to the rate matching block. Note that E_i is a multiple of 3 for turbo encoded TrCHs and that the transport format combination number <u>jl</u> for simplicity has been left out in the bit numbering, i.e. $E_i = N_{ij} N_{il}^{TTI}$. The bits

after separation are denoted by $x_{bi1}, x_{bi2}, x_{bi3}, \dots, x_{biX_i}$. For turbo encoded TrCHs with puncturing, *b* indicates systematic, first parity, or second parity bit. For all other cases *b* is defined to be 1. X_i is the number of bits in each separated bit sequence. The relation between C_{ik} and x_{bik} is given below.

For turbo encoded TrCHs with puncturing:

$x_{1,i,k} = c_{i,3(k-1)+1}$	$k = 1, 2, 3,, X_i$	$X_i = E_i/3$
$x_{2,i,k} = c_{i,3(k-1)+2}$	$k = 1, 2, 3,, X_i$	$X_i = E_i/3$
$x_{3,i,k} = c_{i,3(k-1)+3}$	$k = 1, 2, 3,, X_i$	$X_i = E_i / 3$

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

 $x_{1,i,k} = c_{i,k}$ $k = 1, 2, 3, ..., X_i$ $X_i = E_i$

4.2.7.4.2 Bit collection

The bits x_{bik} are input to the rate matching algorithm described in section 4.2.7.5. The bits output from the rate matching algorithm are denoted $y_{bi1}, y_{bi2}, y_{bi3}, \dots, y_{biY_i}$.

Bit collection is the inverse function of the separation. The bits after collection are denoted by $z_{bi1}, z_{bi2}, z_{bi3}, \dots, z_{biY_i}$. After bit collection, the bits indicated as punctured are removed and the bits are then denoted by $g_{i1}, g_{i2}, g_{i3}, \dots, g_{iG_i}$, where *i* is the TrCH number and $G_i = N_{ij} N_{il}^{TTI} + \Delta N_{ij} \Delta N_{il}^{TTI}$. The relations between y_{bik}, z_{bik} , and g_{ik} are given below.

For turbo encoded TrCHs with puncturing $(Y_i = X_i)$:

$z_{i,3(k-1)+1} = y_{1,i,k}$	$k = 1, 2, 3,, Y_i$
$z_{i,3(k-1)+2} = y_{2,i,k}$	$k = 1, 2, 3,, Y_i$
$z_{i,3(k-1)+3} = y_{3,i,k}$	$k = 1, 2, 3,, Y_i$

After the bit collection, bits $z_{i,k}$ with value δ , where $\delta \notin \{0, 1\}$, are removed from the bit sequence. Bit $g_{i,1}$ corresponds to the bit $z_{i,k}$ with smallest index *k* after puncturing, bit $g_{i,2}$ corresponds to the bit $z_{i,k}$ with second smallest index *k* after puncturing, and so on.

For uncoded TrCHs, convolutionally encoded TrCHs, and turbo encoded TrCHs with repetition:

$$z_{i,k} = y_{1,i,k}$$
 $k = 1, 2, 3, ..., Y_i$

When repetition is used, $g_{i,k}=z_{i,k}$ and $Y_i=G_i$.

When puncturing is used, $Y_i = X_i$ and bits $z_{i,k}$ with value δ , where $\delta \notin \{0, 1\}$, are removed from the bit sequence. Bit $g_{i,1}$ corresponds to the bit $z_{i,k}$ with smallest index *k* after puncturing, bit $g_{i,2}$ corresponds to the bit $z_{i,k}$ with second smallest index *k* after puncturing, and so on.

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4.2.7.1 Determination of rate matching parameters in uplink

4.2.7.1.1 Determination of SF and number of PhCHs needed

In uplink, puncturing can be applied to match the CCTrCH bit rate to the PhCH bit rate. The bit rate of the PhCH(s) is limited by the UE capability and restrictions imposed by UTRAN, through limitations on the PhCH spreading factor. The maximum amount of puncturing that can be applied is signalled from higher layers and denoted by *PL*. The number of available bits in the radio frames of one PhCH for all possible spreading factors is given in [2]. Denote these values by N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , and N_4 , where the index refers to the spreading factor. The possible number of bits available to the CCTrCH on all PhCHs, N_{data} , then are { N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , $4N_4$, $5N_4$, $6N_4$ }.

For a RACH CCTrCH SET0 represents the set of N_{data} values allowed by the UTRAN, as set by the minimum SF provided by higher layers. SET0 may be a sub-set of { N_{256} , N_{128} , N_{64} , N_{32} }. SET0 does not take into account the UE's capability.

For other CCTrCHs, SET0 denotes the set of N_{data} values allowed by the UTRAN and supported by the UE, as part of the UE's capability. Depending on the UE capability and the restrictions from UTRAN, the allowed set of N_{data} , denoted SET0₇ can be a subset of { N_{256} , N_{128} , N_{64} , N_{32} , N_{16} , N_8 , N_4 , $2N_4$, $3N_4$, $4N_4$, $5N_4$, $6N_4$ }.

 $N_{data, i}$ for the transport format combination j is determined by executing the following algorithm:

SET1 = {
$$N_{data}$$
 in SET0 such that $\min_{1 \le y \le I} \{ RM_y \} \cdot N_{data} - \sum_{x=1}^{I} RMx \cdot N_{x,j}$ is non negative }

If SET1 is not empty and the smallest element of SET1 requires just one PhCH then

$$N_{data,i} = \min \text{SET1}$$

else

SET2 = {
$$N_{data}$$
 in SET0 such that $\min_{1 \le y \le I} \{RM_y\}$ $N_{data} - PL \cdot \sum_{x=1}^{I} RMx \cdot N_{x,j}$ is non negative }

Sort SET2 in ascending order

 $N_{data} = \min \text{SET2}$

While N_{data} is not the max of SET2 and the follower of N_{data} requires no additional PhCH do

$$N_{data}$$
 = follower of N_{data} in SET2

End while

 $N_{data,j} = N_{data}$

End if

For a RACH CCTrCH, if N_{data,i} is not part of the UE's capability then the TFC j cannot be used.