Agenda Item : adhoc 16 (Measurements)

Source : Siemens

Title : Update of TS 25.225 concerning measurement definitions, ranges

and mappings (revised version)

Document for : approval

1. Introduction

This Tdoc proposes changes to the current TDD specification TS 25.225 [1] concerning measurement definitions, ranges and mappings of the measurements.

It is based on Tdoc RAN WG1 (99) i82 [16] and includes changes agreed in a drafting session (Ericsson, Nortel, Siemens) during the first day of the meeting which had the intention to harmonise the proposals for the ranges and mappings for FDD (Tdocs RAN WG1 (99) i75 [17], i69 & i73 & j41, i74 & j42) and for TDD (Tdoc RAN WG1 (99) i82) where it is reasonable.

Concerning the ranges & mappings the following assumptions were made:

- The *proposed ranges* are based on the maximum range that is assumed to occur considering all possible cases (and some margin if it was found to be useful). For the TX power ranges this does not mean that the entire range has to be measured by each UE or Node B, but only within its dynamic range (see UE power class and base station maximum output power) the measurements are needed. Therefore also the accuracies specified by RAN WG4 should take this into account.
- The ranges often include cases ,greater than upper limit' or ,lower than lower limit' cases. This was done for two reasons: On the one hand to have a predictable behaviour if a measurement value occurs which is not in the range and on the other hand if the intended range with the chosen resolution was (slightly) reduced to fit into an efficient bit resolution.
- Concerning the *upper and the lower limit of the range* as well as for the *resolution* it was found to be advantageous to consider same values for FDD and TDD if it is applicable. ,Applicable' means that the measurement is used in both modes, the ranges are similar and there is no effort (i.e. additional bits) for the extension for the smaller range.
- The *mapping of the ranges* to a certain number of bits was done in that way that an efficient use of the number of bits is achieved (e.g. it is not efficient if the most significant bit is only needed for less than 10% of the range; in this case a reduction of the range was considered to save a bit).

A CR proposal introducing these changes into the specification is attached.

2. Measurement definitions

This section has the intention to clarify the details of the measurements in a TDD cell.

2.1 UE: SFN-SFN observed time difference

Comparable to the FDD specification 25.215 [8] it is the proposal for the TDD specification 25.225 [1] to distinguish also 2 types of the measurement ,SFN-SFN observed time difference (the exact definition can be found in the CR proposal):

- Type 1 (when serving and target cell do not have the same frame timing or the SFN numbering is not synchronous):
 - Difference between the start of the received frame SFN(serving) of the serving TDD cell and the start of the received frame SFN(target) in the target UTRA cell which is following next after the frame start of the serving cell. Time difference in chips.
 - Analogous to FDD this time difference is defined modulo 256 frames (the 8 least significant bit of the SFN, which means up to 2.56sec –1 chip).
- Type 2 (when serving and target cell have the same frame timing and SFN numbering): Difference between the start of a received timeslot of the serving TDD cell and the start of a received timeslot of the target UTRA cell that is closest in time to the start of the timeslot of the serving TDD cell. Time difference in chips.
 - It allows the determination of propagation delays from the UE to different UTRA cells.

2.2 UE: Observed time difference to GSM cell

For the TDD specification 25.225 [1] the following definition is proposed:

,Observed time difference to GSM cell' is the time difference in ms between the start of the received frame SFN=0 of the serving TDD cell and the start of the received 51-GSM-multiframe of the considered target GSM frequency which is following next after the start of frame SFN=0 of the serving cell. (The exact definition can be found in the CR proposal).

The purpose of this measurement is to ease the monitoring and handover to the GSM cell from the TDD cell by some knowledge about the relative timing.

2.3 UTRAN: RX Timing Deviation

This measurement (TDD only) may be used for timing advance calculations and location services.

For the TDD specification 25.225 [1] the following definition is proposed (the exact definition can be found in the CR proposal):

'RX Timing Deviation' is the time difference in chips between the reception of the first significant uplink path to be used in the detection process and the beginning of the respective slot according to the Node B internal timing.

2.4 UTRAN and UE: RSCP, ISCP and SIR

Following the discussions on the RAN WG1 email reflector on dependence of RSCP and ISCP and SIR on the spreading factor SF we thought some clarifications are necessary:

- The received signal code power (RSCP) and the interference on signal code power (ISCP) can only be determined *after* despreading. This was intended to be expressed in the definition.
- The 'reference point antenna connector' is mentioned to enable a testing of power measurement accuracies since a UE or base station internal reference point would make it difficult/impossible to interpret the measurement result. The relating to this reference point can be considered as a normalization of the measurement.
- Considering the RSCP of a code with spreading factor SF the energy per chip Ec divided by the chip duration Tc before despreading (at the antenna connector) will be equal to the energy collected for one bit SF*Ec divided by the duration of one bit SF*Tc after despreading. Therefore both power values remain the same and independent of SF.
- Nevertheless, since the transmit power is controlled in that way that it is proportional to 1/SF. This means that both values of RSCP before and after despreading are increasing in the same way when the spreading factor is reduced.
 - This was the reason why in RAN WG1 Tdoc (99) i72 it is proposed to multiply the SIR with SF (or SF/2). It is just to get a more uniform exploitation of the measurement range for all the different SF. However this is more critical for FDD (SFmax=256) than for TDD (SFmax=16).
- The multiplication of SIR with SF (or SF/2) assumes that ISCP is independent of SF!

 If it is assumed that all codes (with SFmax) after despreading experience the same non-orthogonal interference ISCP, it is questionable whether ISCP is independent of SF. It is more likely that a code with SF=8 is interfered by two times of the interference for a code with SF=16 (i.e. ISCP after despreading ~ 1/SF).

Our proposal is therefore to leave the RSCP and ISCP definition untouched (we can add a note to differentiate between the measurement point and the reference point to which the measurement is normalized if this is really necessary). Furthermore, since in the current SIR definition RSCP/ISCP is independent of SF (see explanations above) the proposal of a multiplication of RSCP/ISCP with SF is not proposed because it is based on the questionable assumptions 'ISCP independent of SF' and would therefore lead to the opposite of what was intended.

3. Measurement Ranges and Mappings

This section is divided into 4 sections: RX power measurements (including SIR and Ec/No), TX power measurements, time measurements and error rate measurements.

3.1 RX Power Measurements

3.1.1 UE: GSM carrier RSSI

Instead of citing the GSM 05.08 specification [5] it is proposed to include the detailed mapping:

GSM_carrier_RSSI_LEV 0: less than -110dBm GSM_carrier_RSSI_LEV 1: -110dBm to -109dBm GSM_carrier_RSSI_LEV 2: -109dBm to -108dBm ...

GSM_carrier_RSSI_LEV 62: -49dBm to -48dBm GSM_carrier_RSSI_LEV 63: greater than -48dBm That means GSM_carrier_RSSI_LEV is mapped on 6 bits.

3.1.2 UE: UTRA carrier RSSI

The noise floor level at the antenna connector reference point is -174.4 dBm + 10*Log10(B/Hz) = -108.55 dBm for the minimum UE operating temperature T=-10°C [25.102, see 2] and a signal bandwidth: B=3.84MHz.

The maximum input level, respectively received power, at the UE is specified with –25dBm for this bandwidth [25.102, see 2].

This would lead to a range -109 dBm to -25 dBm. However, in pratice neither the upper limit (high interference/load situation) nor the lower limit (no signal apart from thermic noise) will be reached.

For FDD Tdoc i75 [17] formerly proposed a range of –95dBm to –30dBm with 1dB which can not be mapped on 6 bits. So the harmonised proposal is slighly modified.

According to [6] an implementation with an accuracy down to $\pm 2.5 dB$ seems to be feasible for the whole range. The accuracies proposed by RAN WG4 in [7] consider the values $\pm 4 dB$ below -70 dBm and $\pm 6 dB$ over the full range (as they are also applied for GSM [5]). We therefore assume that a 1dB resolution is sufficient.

So for TDD (and also for FDD) the proposal is:

3.1.3 UTRAN: RSSI

The noise floor level at the antenna connector reference point is about –109dBm (compare 3.1.2) and the dynamic range of the receiver input power at the antenna connector is specified to 30dB [25.105, see 3].

Therefore the maximum received power at the antenna connector of the base station would be -79dBm.

Taking also into account that the thermic noise level will not be reached it seems to be useful to take a lower limit of more than -109dBm, e.g. -105dBm.

For FDD the formerly [17] proposed range was -105dBm to -70dBm in 0.5dB steps which required 7bits. In the harmonised proposal the range was slightly modified to fit into 6 bits. (According to [25.104, see 10] the receiver's dynamic range is also 30dB for FDD.)

According to [6] an implementation with an accuracy down to ±1.5dB seems to be feasible for the whole range. The

accuracy proposed by RAN WG4 in [7] considers ±4dB over that range.

So due to the smaller dynamic range in the UTRAN compared to the UE we propose a finer resolution of 0.5dB.

```
For TDD (and also for FDD) the proposal is:
```

```
RSSI LEV 0:
                          RSSI < -105.0dBm
RSSI\_LEV 1:-105.0dBm \le RSSI < -104.5dBm
RSSI LEV 2:-104.5dBm \leq RSSI < -103.0dBm
RSSI\_LEV~62:~-74.5dBm \le ~RSSI < -74.0dBm
```

RSSI_LEV 63: -74.0dBm≤ RSSI

which means RSSI_LEV is mapped on 6 bits.

3.1.4 UE: RSCP and Primary CCPCH RSCP

A maximum received power at the UE antenna port of -25dBm is specified in 25.102 [2] for TDD (same value for FDD see 25.104 [9]).

Considering the noise floor of -108.55dBm in 3.1.2 and taking a maximum spreading gain of 12dB (TDD: SFmax=16) into account we don't expect values below -120.55dBm for RSCP.

The TDD UE reference sensitivity level is specified in 25.102 [2] to -105dBm (for 12.2kbps & BER<=0.001, signal bandwidth B=243kHz and noise figure NF(UE)=9dB, see also [13]). (For FDD the reference sensitivity of the UE receiver is calculated to -117dBm, see [14]).

These values consider an AWGN channel and a UE noise figure NF=9dB (note: this value is not standardized) to read a dedicated channnel with a BER of 0.1%.

For FDD formerly [17] a range of -115dBm ... -40dBm with steps of 1dB was proposed which required 7 bits. The lower limit was estimated using the reference sensitivity level -117dBm and taking +5dB for a fading channel and further -3dB (to allow measuring the channel before reading is possible) into account. For TDD the same consideration would lead to a lower limit of -103dBm.

For the reason of compatibility with FDD and also as a further margin (future services, perhaps a little lower NF etc.) we would also use -115dBm as the lower limit of the range.

Concerning the upper limit we agree that for cell selection/reselection or handover decisions it might not be necessary to distinguish between high levels of -40dBm ... -25dBm at the UE. However since a measurement on the P-CCPCH (or another channel with reference TX power) can be used for pathloss calculations and power control it would be useful to resolve also the measurement results above -40dBm. This seems applicable since only a part of the 128 steps of the 7bit resolution are used in the FDD proposal. (Even for FDD an upper limit of -25dBm will be proposed to harmonise the measurement range & mapping with TDD since it proposes no further effort.)

The accuracy currently discussed in RAN WG4 [7] for these RSCP measurements is ±6dB over the full range.

For the range and mapping for the RSCP and the P-CCPCH RSCP measurement of TDD (and FDD) the proposal is:

```
RSCP < -115dBm
RSCP 000:
RSCP\_001: -115dBm \le RSCP < -114dBm
RSCP\_002: -114dBm \le RSCP < -113dBm
RSCP 101:
            -24dBm \le RSCP < -25dBm
            -25dBm \le RSCP
RSCP_102:
(RSCP_103 to RSCP_127: not used)
```

3.1.5 UE: CPICH RSCP

For TDD this measurement is used for monitoring FDD cells in preparation of cell selection/reselection or handover. For FDD the former proposal [17] was: range of -115dBm ... -40dBm in steps of 1dB.

For TDD and FDD it is now proposed to use the same range & mapping as proposed for RSCP and P-CCPCH RSCP in 3.1.4.

3.1.6 UE: Timeslot ISCP

This measurement may be used for DCA, HO evaluation and for SIR calculations.

The accuracy currently discussed in RAN WG4 [7] for this ISCP measurement is ±6dB over the full range.

The considerations are similar to the RSCP of 3.1.4. Therefore, we propose to use the RSCP range and mapping of 3.1.4 also for the TDD UE measurement Timeslot ISCP.

3.1.7 UE: SIR

For FDD the proposed [17] range range is –10dB ... 20dB in steps of 0.5dB (60steps or 6bit). Since in our opinion the definition RSCP/ISCP is already independent of SF (see 2.4) we propose for TDD:

```
\begin{array}{lll} \mbox{UE\_SIR\_00:} & \mbox{SIR} < -11.0dB \\ \mbox{UE\_SIR\_01:} & -11.0dB \leq & \mbox{SIR} < -10.5dB \\ \mbox{UE\_SIR\_02:} & -10.5dB \leq & \mbox{SIR} < -10.0dB \\ \mbox{....} & \mbox{UE\_SIR\_61:} & 19.0dB \leq & \mbox{SIR} < & 19.5dB \\ \mbox{UE\_SIR\_62:} & 19.5dB \leq & \mbox{SIR} < & 20.0dB \\ \mbox{UE\_SIR\_63:} & 20.0dB \leq & \mbox{SIR} & \mbox{SIR} & \mbox{SIR} & \mbox{SIR} \\ \end{array}
```

3.1.8 UE: CPICH Ec/No

For TDD this measurement is used for monitoring FDD cells in preparation of cell selection/reselection or handover. For FDD the former proposal [17] was a range of -20dB ... 0dB ('as it is used for IS-95') in steps of 1dB. Due to the maximum spreading gain 24dB for FDD (Sfmax=256) it is now proposed to extend the lower limit to -24dB

So the common proposal for FDD and TDD is

```
\begin{array}{lll} {\rm CPICH\_Ec/No\_00:} & {\rm Ec/No} < -24 {\rm dB} \\ {\rm CPICH\_Ec/No\_01:} & -24 {\rm dB} \le & {\rm Ec/No} < -23 {\rm dB} \\ {\rm CPICH\_Ec/No\_02:} & -23 {\rm dB} \le & {\rm Ec/No} < -22 {\rm dB} \\ {\rm ...} \\ {\rm CPICH\_Ec/No\_23:} & -2 {\rm dB} \le & {\rm Ec/No} < & -1 {\rm dB} \\ {\rm CPICH\_Ec/No\_24:} & -1 {\rm dB} \le & {\rm Ec/No} < & 0 {\rm dB} \\ {\rm CPICH\_Ec/No\_25:} & 0 {\rm dB} \le & {\rm Ec/No} \\ {\rm (CPICH\_Ec/No\_26 \ to \ CPICH\_Ec/No\_31: \ not \ used)} \end{array}
```

3.1.9 UTRAN: RSCP

This measurement is only used for TDD.

Considering the noise floor of about -109dBm in 3.1.3 and taking a maximum spreading gain of 12dB (TDD: SFmax=16) into account we don't expect values below -121dBm for RSCP measured in the UTRAN. The TDD base station reference sensitivity level is specified in 25.105 [3]: -110dBm (for 12.2kbps & BER<=0.001, signal bandwidth B=243kHz and NF(BS)=5dB, see also [13]). (For FDD a sensitivity level of -122dBm is specified in 25.104 [10], see also [15]).

The dynamic range for the base station receiver input is 30dB, specified for TDD in 25.105 [3] and for FDD in 25.104 [10]. This results in a maximum received power at the base station of -80dBm for TDD.

Analogous to the RSSI measured by the UTRAN we propose a step size of 0.5dB.

For this step size and an upper limit of -80dBm this would result in a lower limit of -111dBm with a 6 bit resolution which might be to high. To have some margin we propose a 7 bit resolution for this RSCP measurement:

```
\begin{array}{llll} UTRAN\_RSCP\_00: & UTRAN\_RSCP < -120.0dBm \\ UTRAN\_RSCP\_01: -120.0dBm & \leq UTRAN\_RSCP < -119.5dBm \\ UTRAN\_RSCP\_02: -119.5dBm & \leq UTRAN\_RSCP < -119.0dBm \\ ... & UTRAN\_RSCP\_81: -79.5dBm & \leq UTRAN\_RSCP < -80.0dBm \\ UTRAN\_RSCP\_82: -80.0dBm & \leq UTRAN\_RSCP \\ (UTRAN\_RSCP\_83 \ to \ UTRAN\_RSCP\_127: \ not \ used) \end{array}
```

3.1.10 UTRAN: Timeslot ISCP

This measurement may be used for DCA, HO evaluation and for SIR calculations.

The accuracy currently discussed in RAN WG4 [7] for this ISCP measurement is ±6dB over the full range.

The considerations are similar to the RSCP of 3.1.9 therefore we propose to use the same range & mapping for this TDD UTRAN measurement 'Timeslot ISCP'.

3.1.11 UTRAN: SIR

For FDD the former proposal was [17] –10dB ... 20dB in steps of 0.5dB with a working point between –3dB and +4dB.

For TDD we propose the same range and mapping as for the TDD SIR measurement in the UE, see 3.1.7.

3.2 TX Power Measurements

3.2.1 UE: UE transmitted power

In 25.102 [2] the minimum transmit power –44dBm for the UE is specified. A maximum UE ouput power of 30dBm is also specified there for TDD power class 1.

For FDD a maximum UE output power of 33dBm for power class 1 and also a minimum UE transmit power of –44dBm is specified in 25.101 [9]. For FDD a range of –50dBm ... 33dBm is proposed in [17] with steps of 1dB. The lower limit is proposed to include cases where the UE is very close to the base station.

So for the UE transmitted power a measurement range of -44dBm to 33dBm (to harmonise the range with FDD) seems to be reasonable. However, there are also discussions to reduce the lower limit -44dB (e.g. to -50dB) which may be useful if the UE is very close to the base station.

According to [6] an implementation limit for the absolute measurement of $\pm 4dB$ over the whole range and $\pm 2dB$ over the range from 15dBm to 30dBm is assumed. The minimum power control step size tolerance of $\pm 0.5dB$ for the 1dB power step size is specified for TDD in 25.102 [2]. However this does not mean that the UE transmitted power measurement must have a resolution of equal or less than 0.5dB.

So to harmonise the range and mapping with the FDD proposal we also propose a range from -50dBm to 33dBm with a stepsize of 1dB.

(UE_TX_POWER_000 to UE_TX_POWER_020: currently not used, but reserved for further extension to lower values)

```
UE_TX_POWER_021: -50dBm ≤ UE_TX_POWER < -49dBm
UE_TX_POWER_022: -49dBm ≤ UE_TX_POWER < -48dBm
UE_TX_POWER_023: -48dBm ≤ UE_TX_POWER < -47dBm
...
UE_TX_POWER_104: 31dBm ≤ UE_TX_POWER < 32dBm
UE_TX_POWER_105: 32dBm ≤ UE_TX_POWER < 33dBm
UE_TX_POWER_106: 33dBm ≤ UE_TX_POWER < 34dBm
```

(UE_TX_POWER_107 to UE_TX_POWER_127: currently not used, but reserved for further extension to higher values)

3.2.2 UTRAN: Transmitted carrier power

As specified in 25.105 [3] for TDD and 25.104 [10] for FDD the maximum base station power has to be declared by the manufacturer but no limit is specified. Nevertheless, the maximum base station transmit power values used in RF system scenario simulations in 25.924 [12] are 43dBm (=20W) for macro cells and 33 dBm (2W) for micro cells.

Although 20W (=43dBm) are a realistic assumption now, a higher upper limit should be chosen to take care of possible future extensions, e.g. 50dBm. Also a margin for the lower limit should be taken into account (e.g. for pico cells).

In 25.105 [3] for TDD a downlink total power dynamic range of 30dB is specified (for FDD in 25.104 [10] '18dB or greater') which would imply 3dBm for the micro base station based on the simulation assumption.

In RAN WG4 values for the accuracy of this measurement between ±6dB or ±3dB are discussed for TDD (see [7]). For FDD formerly a range of 10dBm to 50dBm with steps of 0.5dB was proposed [17].

Now the proposed range and mapping for TDD (and also for FDD) is

(UTRAN_TX_POWER_000 to UTRAN_TX_POWER_015: currently not used, but reserved for further extension to lower values)

```
UTRAN_TX_POWER_016:
                          0.0dBm \le UTRAN\_TX\_POWER < 0.5dBm
UTRAN_TX_POWER_017:
                          0.5dBm \le UTRAN\_TX\_POWER < 1.0dBm
UTRAN_TX_POWER_018:
                          1.0dBm \le UTRAN_TX_POWER < 1.5dBm
UTRAN_TX_POWER_115:
                          49.0dBm \le UTRAN_TX_POWER < 49.5dBm
                          49.5 dBm \ \leq \ UTRAN\_TX\_POWER < 50.0 dBm
UTRAN_TX_POWER_116:
```

 $50.0dBm \le UTRAN_TX_POWER < 50.5dBm$ (UTRAN_TX_POWER_118 to UTRAN_TX_POWER_127: currently not used, but reserved for further extension to higher values)

UTRAN: Transmitted code power

UTRAN_TX_POWER_117:

For FDD formerly proposed [17] a range from 0dBm to 46dBm with steps of 0.5dB.

For TDD and FDD it is now proposed (to have some margin below the UE TX carrier power):

(UTRAN_TX_CODE_POWER_000 to UTRAN_TX_ CODE_POWER_009: currently not used, but reserved for further extension to lower values)

```
UTRAN_TX_ CODE_POWER_010:
                             -10.0dBm \le UTRAN_TX_CODE_POWER < -9.5dBm
UTRAN TX CODE POWER 011:
                              -9.5dBm≤ UTRAN TX CODE POWER < -8.5dBm
UTRAN_TX_ CODE_POWER_012:
                              -8.5dBm \le UTRAN_TX_CODE_POWER < -7.5dBm
UTRAN TX CODE POWER 121:
                             45.0dBm ≤ UTRAN TX CODE POWER < 45.5dBm
UTRAN_TX_ CODE_POWER_122:
                             45.5dBm ≤ UTRAN_TX_ CODE_POWER < 46.0dBm
UTRAN_TX_ CODE_POWER_123:
                             46.0dBm \le UTRAN_TX_CODE_POWER < 46.5dBm
```

(UTRAN_TX_CODE_POWER_124 to UTRAN_TX_POWER_127: currently not used, but reserved for further extension to higher values)

3.3 Time Measurements

UE: SFN-SFN observed time difference

This measurement may be used for cell reselection, handover and location services.

The range for type 1 of this measurement described in 2.1 is the same as for this measurement type 1 of FDD [25.215, see8] that means 0 chips ... (255 * 38400 + 38399 = 9830399 chips) in steps of 1 chip which requires 24 bits.

(Annotation to 25.215 definition of type 1: There seems to a writing error in the definition of Tm since as it is defined now Tm would always be negative contradicting to the mentioned range.)

For measurements between synchronized cells the type 2 of 2.1 applies. The range of this measurement is -1279 chips ... 1280 chips in steps of 1 chip which requires 12 bits.

For both types of this measurement an accuracy of ± 0.5 chips was proposed in [7].

3.3.2 UE: Observed time difference to GSM cell

This measurement may be used for handover to GSM.

With the duration of a 51GSM-multiframe of 51*60ms/13 = 3060ms/13 = about 235.3846ms the range of this measurement can be 0ms ... 3060ms/13 * (1-1/2^N) with N bits resolution.

With 10 bits the resolution would be 3060ms/(13*1024)=229.87µs which is about 0.4 GSM slot.

With 12 bits the resolution would be 3060ms/(13*4096)=57.47µs which is about 0.1 GSM slot.

The latter resolution is also proposed for FDD and is therefore also accepted for TDD.

For this measurement an accuracy of ± 0.5 chips was proposed in [7].

3.3.3 UTRAN: RX Timing Deviation

Since according to [25.224, see 11] the ,timing advance will be represented as an 8 bit number (0-255) being the multiple of 4 chips' we propose a range of 0 chips ... 255*4chips = 1020 chips.

Since this measurement may also be used for location services we propose a resolution of 1 chip (propagation distance for one chip: 78.125m) which results in a 10 bit resolution.

For this measurement an accuracy of ± 0.5 chips was proposed in [7].

3.4 Error Rate Measurements

3.4.1 UE and UTRAN: Physical channel BER

For GSM in [05.08, see 5] 3 bits are specified for a bit error rate before channel decoding:

```
RXQUAL 0:
                    BER < 0.2\%
RXQUAL 1:
             0.2\% < BER < 0.4\%
RXQUAL 2:
             0.4\% < BER < 0.8\%
RXQUAL 3:
             0.8\% < BER < 1.6\%
RXQUAL 4:
             1.6\% < BER < 3.2\%
RXQUAL_5:
             3.2\% < BER < 6.4\%
RXQUAL_6:
             6.4% < BER < 12.8%
RXQUAL 7:
             12.8% < BER
```

that means the main range considers rates of 0.002 to 0.128.

For FDD it is discussed to consider a raw BER range from -5.08 < Log10(raw BER) < 0 in 0.02 steps and a further value raw BER=0 (that means an 8 bit resolution). On the other hand there is a proposal to delete this measurement due to the .flat behaviour in some environments'.

From our point of view: On the one hand, as mentioned in the definition the raw BER measurement is just an estimation. Due to the different coding possibilities it also may be complex to estimate from the raw BER the consequences for the user BER (<< raw BER). On the other hand the BER measurement may be a means for first quick and dirty guess for the signal quality which may be improved by averaging.

Therefore it is our proposal to relax the resolution to 4 bit (in contrast to FDD proposal of 8 bit).

From simulations it can be noticed that the raw BER usually has values in the range from 1% to 20%. Proposal for TDD:

```
Raw BER 00:
                        BER < 0.1 \%
Raw BER 01:
                0.1\% < BER < 0.3\%
Raw BER 02:
                0.3\% < BER < 0.5\%
                0.5\,\%\ < BER <\ 1\quad\%
Raw_BER_03:
Raw_BER_04:
                1 % < BER < 2 %
Raw BER 05:
                2 % < BER < 4 %
Raw_BER_06:
                4 % < BER < 6 %
                6 % < BER < 8 %
Raw BER 07:
                8 % < BER < 10 %
Raw BER 08:
Raw BER 09:
                10 % < BER < 12 %
Raw_BER_10:
                12 % < BER < 14 %
Raw BER 11:
                14 % < BER < 16 %
Raw BER 12:
                16 % < BER < 18 %
Raw_BER_13:
                18 % < BER < 20 %
Raw_BER_14:
                20 % < BER < 30 %
Raw_BER_15:
                30 % < BER
with quantisation steps of 0.2%, 0.5%, 1%, 2%, 10%
```

3.4.2 UE and UTRAN: Transport channel BLER

For the ,Transport channel block error rate' there is a proposal to use the same range and mapping as proposed for the raw BER: -5.08 < Log10(BLER) < 0 in 0.02 steps and a further value BLER=0 (that means an 8 bit resolution) for FDD

Since the BLER of a transport channel is considered after channel decoding it gives a more realistic impression of the signal quality than the raw BER. Furthermore, it requires a larger time period to evaluate the BLER. Therefore we propose for TDD to modify the FDD proposal:

• For the lower limit we assume that BLER about 10^-4 since for a lower BLER the required time period is not practicable (several hours, depending on interleaving and accuracy).

Also we would increase the logarithmic step size from 0.02 (see RAN WG1 Tdoc (99)j42) to 0.15 since we assume that for the lowest considered BLERs the difference would be too small compared to the accuracy (for reasonable measurement times).

```
BLER = 0
BLER_00:
                                            (for cases with a prescibed time period where BLER is evaluated)
BLER 01:
               -4.05
                         \leq Log10(BLER) < -3.90
                                                      (that means 0.0089% to 0.0126%)
BLER_02:
               -3.90
                         \leq Log10(BLER) < -3.75
                                                      (that means 0.0125% to 0.025%)
BLER_03:
               -3.60
                         \leq Log10(BLER) < -3.45
                                                      (that means 0.025% to 0.035%)
BLER 22:
               -0.90
                         \leq Log10(BLER) < -0.75
                                                      (that means 12.59% to 17.78%)
               -0.75
                         \leq Log10(BLER) < -0.60
BLER_23:
                                                      (that means 17.78% to 25.12%)
BLER_24:
               -0.60
                         \leq Log10(BLER) < -0.45
                                                      (that means 25.12% to 35.48%)
BLER 25:
               -0.45
                         \leq Log10(BLER) < -0.30
                                                      (that means 35.48% to 50.12%)
               -0.30
BLER_26:
                         \leq Log10(BLER) < -0.15
                                                      (that means 50.12% to 70.79%)
BLER 27:
               -0.15
                         \leq Log10(BLER) \leq
                                             0.00
                                                      (that means 70.79% to 100%)
(BLER_28 to BLER_31: not used)
```

4. References

- [1] RAN WG1 TS 25.225 v.3.0.0 (1999-10) Physical Layer Measurements (TDD), source: RAN #5
- [2] RAN WG4 TS 25.102 v.3.0.0 (1999-10) UTRA (UE) TDD; Radio Transmission and Reception, source: RAN #5
- [3] RAN WG4 TS 25.105 v.3.0.0 (1999-10) UTRA (BS) TDD; Radio Transmission and Reception, source: RAN #5
- [4] GSM 05.05 (Phase 2+), Radio Transmission and Reception, source: ETSI
- [5] GSM 05.08 (Phase 2+), Radio Subsystem Link Control, source: ETSI
- [6] RAN WG4 AH02 Tdoc (99)R4rrm05, RRM Measurements performance requirements for TDD, source: Siemens
- [7] RAN WG4 Tdoc (99)785 Annex B Measurement Requirements, Report of AH02: TS 25.123 and TS 25.133 drafting session (Helsinki, Nov. 18-19, 1999), source: RAN WG4 AH02
- [8] RAN WG1 TS 25.215 v.3.0.0 (1999-10) Physical Layer Measurements (FDD), source: RAN #5
- [9] RAN WG4 TS 25.101 v.3.0.0 (1999-10) UTRA (UE) FDD; Radio Transmission and Reception, source: RAN #5
- [10] RAN WG4 TS 25.104 v.3.0.0 (1999-10) UTRA (BS) FDD; Radio Transmission and Reception, source: RAN #5
- [11] RAN WG1 TS 25.224 v.3.0.0 (1999-10) Physical Layer Procedures (TDD), source: RAN #5
- [12] RAN WG4 TR 25.924 v2.0.0 (1999-10) RF System Scenarios, source: RAN #5
- [13] RAN WG4 Tdoc (99)549 Proposal for UE and BS reference sensitivity in the TDD mode (Makuhari, Japan), source: Siemens
- [14] RAN WG4 Tdoc (99)012 MS receiver sensitivity in UTRA FDD mode (Espoo, Finland), source: Nokia
- [15] RAN WG4 Tdoc (99)373 Required Eb/No for Voice channel (12.2kbps, BS for FDD), source: NEC, NTT DoCoMo, Panasonic, Fujitsu
- [16] RAN WG1 Tdoc (99)i82 Update of TS 25.225 concerning measurement definitions, ranges and mappings (Dresden, Germany), source: Siemens
- [17] RAN WG1 Tdoc (99)i75 CR 25.215-009: Range and resolution for RF related measurements (Dresden, Germany), source: Ericsson

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| Reason for change: | | urement definition he measurements | | | e more precise | and range & ma | pping |
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| Other comments: | Draft CR pro | posed in RAN WO | G1 #9 Td | oc (99) i82 | | | |
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monitor is available, the UE may perform the measurements on the PCCPCH directly without prior SCH synchronisation.

4.1 4.4 Measurements for DCA

DCA is used to optimise the resource allocation by means of a channel quality criteria or traffic parameters. The DCA measurements are configured by the UTRAN. The UE reports the measurements to the UTRAN.

For DCA no measurements are performed in idle mode in the serving TDD cell.

When connecting with the initial access the UE immediately starts measuring the ISCP of time slots which are communicated on the BCH. The measurements and the preprocessing are done while the UTRAN assigns an UL channel for the UE for signalling and measurement reporting.

In connected mode the UE performs measurements according to a measurement control message from the UTRAN.

4.2 4.5 Measurements for timing advance

To update timing advance of a moving UE the UTRAN measures 'Received Timing Deviation', i.e. the time difference of the received UL transmission (PRACH, DPCH, PUSCH) in relation to its timeslot structure that means in relation to the ideal case where an UL transmission would have zero propagation delay. The measurements are reported to higher layers, where timing advance values are calculated and signalled to the UE.

5 Measurement abilities for UTRA TDD

In this chapter the physical layer measurements reported to higher layers. (this may also include UE internal measurements not reported over the air-interface) are defined.

4.3 5.1 UE measurement abilities

- NOTE 1: Measurements for TDD which are carried out on Primary CCPCH (PCCPCH) can also be carried out on another CCPCH if it has the same constant power level as the PCCPCH and no beamforming is used.
- NOTE 2: The UTRAN has to take into account the UE capabilities when specifying the timeslots to be measured in the measurement control message.
- NOTE 3: The RSCP can either be measured on the data part or the midamble of a burst, since there is no power offset between both. However, in order to have a common reference, the measurement on the midamble is assumed.
- NOTE 4: The line 'applicable for' indicates whether the measurement is applicable for inter-frequency and/or intra-frequency and furthermore for idle and/or connected mode.

4.3.1 5.1.1 PCCPCH RSCP

| Definition | Received Signal Code Power, the received power on PCCPCH of own or neighbour cell after despreading. The reference point for the RSCP is the antenna connector at the UE. | | | |
|----------------|--|--|--|--|
| Applicable for | idle mode, connected mode (intra-frequency & inter-frequency) | | | |
| Range/mapping | P-CCPCH_RSCP_000: P-CCPCH_RSCP < -115dBm P-CCPCH_RSCP_001: -115dBm ≤ P-CCPCH_RSCP < -114dBm P-CCPCH_RSCP_002: -114dBm ≤ P-CCPCH_RSCP < -113dBm P-CCPCH_RSCP_101: -24dBm ≤ P-CCPCH_RSCP < -25dBm P-CCPCH_RSCP_102: -25dBm ≤ P-CCPCH_RSCP | | | |

4.3.2 5.1.2 CPICH RSCP

| Definition | Received Signal Code Power, the received power on the CPICH code after despreading. The reference point for the RSCP is the antenna connector at the UE. | | | | |
|----------------|--|--|--|--|--|
| Applicable for | idle mode, connected mode (inter-frequency) | | | | |
| Range/mapping | CPICH_RSCP_000: CPICH_RSCP < -115dBm | | | | |

4.3.3 5.1.3 RSCP

| Definition | Received Signal Code Power, the received power on the code of a specified DPCH or PDSCH after despreading. The reference point for the RSCP is the antenna connector at the UE. | | | | |
|----------------|---|--|--|--|--|
| Applicable for | connected mode (intra-frequency) | | | | |
| Range/mapping | $ \begin{array}{lll} & & & & & & & & & \\ & & & & & & & \\ & & & & & & \\ & & & & & \\ & & & & & \\ & & & & \\ & & & & \\ & & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & \\ & & & \\$ | | | | |

4.3.4 5.1.4 Timeslot ISCP

| Definition | Interference Signal Code Power, the interference on the received signal in a specified timeslot after despreading. Only the non-orthogonal part of the interference is included in the measurement. The reference point for the ISCP is the antenna connector at the UE. | | | |
|----------------|--|--|--|--|
| Applicable for | connected mode (intra-frequency) | | | |
| Range/mapping | $\begin{array}{c ccccccccccccccccccccccccccccccccccc$ | | | |

4.3.5 5.1.5 UTRA carrier RSSI

| Definition | Received Signal Strength Indicator, the wide-band received power within the relevant channel | | | | |
|------------|--|--|--|--|--|
| | bandwidth in a specified timeslot. Measurement shall be performed on a UTRAN DL carrier. | | | | |
| | The reference point for the RSSI is the antenna connector at the UE. | | | | |
| | | | | | |

| Applicable for | idle mode, connected mode (intra- & inter-frequency) | | | | | |
|----------------|--|--|--|--|--|--|
| Range/mapping | UTRA carrier RSSI LEV 0: RSSI < -94dBm | | | | | |

4.3.6 5.1.6 GSM carrier RSSI

| Definition | Received Signal Strength Indicator, the wide-band received power within the relevant channel bandwidth in a specified timeslot. Measurement shall be performed on a GSM BCCH carrier. The reference point for the RSSI is the antenna connector at the UE. | | | |
|----------------|--|--|--|--|
| Applicable for | idle mode, connected mode (inter-frequency) | | | |
| Range/mapping | GSM_carrier_RSSI_LEV 0: less than -110dBm GSM_carrier_RSSI_LEV 1: -110dBm to -109dBm GSM_carrier_RSSI_LEV 2: -109dBm to -108dBm GSM_carrier_RSSI_LEV 62: -49dBm to -48dBm GSM_carrier_RSSI_LEV 63: greater than -48dBm For GSM: according to the definition of RXLEV in GSM 05.08. | | | |

4.3.7 5.1.7 SIR

| Definition | Signal to Interference Ratio, defined as the RSCP of a DPCH or PDSCH divided by ISCP of the same timeslot. The reference point for the SIR is the antenna connector of the UE. | | | | |
|----------------|--|--|--|--|--|
| Applicable for | connected mode (intra-frequency) | | | | |
| Range/mapping | $\begin{array}{llllllllllllllllllllllllllllllllllll$ | | | | |

4.3.8 5.1.8 CPICH Ec/No

| Definition | The received energy per chip divided by the power density in the band. The Ec/No is identical to RSCP/RSSI. The reference point for Ec/No is the antenna connector at the UE. |
|----------------|---|
| Applicable for | idle mode, connected mode (inter-frequency) |
| Range/mapping | |

4.3.9 5.1.9 Physical channel BER

| Definition | The physical channel BER is an estimation of the average bit error rate (BER) before channel | | | | | | |
|----------------|--|--|--|--|--|--|--|
| | decoding of the data. | | | | | | |
| Applicable for | connected mode (intra-frequency) | | | | | | |
| Range/mapping | Raw BER 00: BER < 0.1 % | | | | | | |

4.3.10 5.1.10 Transport channel BLER

| Definition | Estimation of the transport channel block error rate (BLER). The BLER estimation shall be | | | | | | |
|----------------|---|-------|--------------------------------|--|--|--|--|
| | based on evaluating the CRC on each transport block. | | | | | | |
| Applicable for | connected mode (intra-frequency) | | | | | | |
| Range/mapping | BLER_00: BLER = 0 | | | | | | |
| | BLER_01: | -4.05 | \leq Log10(BLER) < -3.90 | | | | |
| | BLER_02: | -3.90 | \leq Log10(BLER) < -3.75 | | | | |
| | BLER_03: | -3.60 | \leq Log10(BLER) < -3.45 | | | | |
| | <u></u> | | | | | | |
| | BLER 22: | -0.90 | \leq Log10(BLER) < -0.75 | | | | |
| | BLER_23: | -0.75 | \leq Log10(BLER) < -0.60 | | | | |
| | BLER_24: | -0.60 | \leq Log10(BLER) < -0.45 | | | | |
| | BLER_25: | -0.45 | \leq Log10(BLER) < -0.30 | | | | |
| | BLER_26: | -0.30 | \leq Log10(BLER) < -0.15 | | | | |
| | BLER_27: | -0.15 | \leq Log10(BLER) \leq 0.00 | | | | |
| | (BLER_28 to BLER_31: not used) | | | | | | |

4.3.11 5.1.11 UE transmitted power

| Definition | The total UE transmitted power on one carrier measured in a timeslot. The reference point for the UE transmitted power shall be the UE antenna connector. |
|----------------|--|
| Applicable for | connected mode (intra-frequency). |
| Range/mapping | -(UE_TX_POWER_000 to UE_TX_POWER_020: not used, but reserved for further extension to lower values) UE_TX_POWER_021: -50dBm ≤ UE_TX_POWER < -49dBm UE_TX_POWER_022: -49dBm ≤ UE_TX_POWER < -48dBm UE_TX_POWER_023: -48dBm ≤ UE_TX_POWER < -47dBm UE_TX_POWER_104: 31dBm ≤ UE_TX_POWER < 32dBm UE_TX_POWER_105: 32dBm ≤ UE_TX_POWER < 33dBm UE_TX_POWER_106: 33dBm ≤ UE_TX_POWER < 34dBm (UE_TX_POWER_106: 33dBm ≤ UE_TX_POWER < 34dBm (UE_TX_POWER_107 to UE_TX_POWER_127: not used, but reserved for further extension to higher values) |

4.3.12 5.1.12 SFN-SFN observed time difference

| Definition | Time difference in the frame timing between the serving TDD cell and the frame timing of the |
|----------------|--|
| | target UTRA cell measured by means of PCCPCH for a TDD cell and by means of CPICH for |
| | an FDD cell. |
| | SFN-SFN observed time difference is the time difference of the reception times of frames from two cells (serving and target) measured in the UE and expressed in chips. It is distinguished in two types: Type 2 applies if the serving and the target cell have the same frame timing and |
| | SFN numbering. Type 1 applies in all other cases. |
| | Type 1: SFN-SFN observed time difference = OFF \times 38400+ T_m in chips, where: |
| | $\underline{T}_{m} = \underline{T}_{RxSFNk} - \underline{T}_{RxSFNi}$, given in chip units with the range [0, 1,, 38399] chips |
| | T _{RxSFNi} : time of start of the received frame SFN _i of the serving TDD cell i. |
| | T_{RxSFNk} : time of start of the received frame SFN_k of the target UTRA cell k after the time |
| | instant T _{RxSFNi} in the UE. If the next frame of the target UTRA cell is received exactly at |
| | T_{RxSFNi} then $T_{RxSFNk} = T_{RxSFNi}$ (which leads to $T_m = 0$). |
| | OFF=(SFN _k - SFN _i) mod 256, given in number of frames with the range [0, 1,, 255] frames SFNi: system frame number for downlink frame from serving TDD cell i in the UE at the |
| | time T _{RxSFNi} . |
| | SFNk: system frame number for downlink frame from target UTRA cell k received in the |
| | UE at the time T _{RxSFNk} .(for FDD: the P-CCPCH frame) |
| | T 2. |
| | Type 2: SFN-SFN observed time difference = T_{RxTSk} - T_{RxTSi} , in chips, where |
| | T_{RxTSi} : time of start of a timeslot received of the serving TDD cell i. |
| | T_{RxTSk} : time of start of a timeslot received from the target UTRA cell k that is closest in |
| | time to the start of the timeslot of the serving TDD cell i. |
| Applicable for | idle mode, connected mode (intra-frequency) |
| Range/mapping | Type 1: 0 chips 9830399 chips with a resolution of 1 chip (24 bit) |
| | Type 2: -1279 chips 1280 chips with a resolution of 1 chip (12 bit) |

4.3.13 5.1.13 Observed time difference to GSM cell

| Definition | Time difference between the Primary CCPCH of the current cell and the timing of the GSM cell Observed time difference to GSM cell is the time difference T _m in ms, where T _m = T _{RxGSMk} - T _{RxSFN0i} T _{RxSFN0i} : time of start of the received frame SFN=0 of the serving TDD cell i T _{RxGSMk} : time of start of the received 51-GSM-multiframe of the considered target GSM beacon frequency k which is following next after the start of frame SFN=0 of the serving TDD cell. |
|----------------|--|
| Applicable for | Idle mode, connected mode (inter-frequency) |
| Range/mapping | 0ms 3060ms/13 – 3060ms/(13*4096) in steps of 3060ms/(13*4096) (12 bit) |

4.4 5.2 UTRAN measurement abilities

NOTE 1: If the UTRAN supports multiple frequency bands then the measurements apply for each frequency band individually.

NOTE 2: The RSCP can either be measured on the data part or the midamble of a burst, since there is no power offset between both. However, in order to have a common reference, the measurement on the midamble is assumed.

4.4.1 5.2.1 RSCP

| Definition | Received Signal Code Power, the received power on one DPCH, PRACH or PUSCH code after despreading. The reference point for the RSCP shall be the antenna connector. | | | | |
|---------------|---|--|--|--|--|
| Range/mapping | UTRAN_RSCP_00: UTRAN_RSCP < -120.0dBm | | | | |

4.4.2 5.2.2 Timeslot ISCP

| Definition | Interference Signal Code Power, the interference on the received signal in a specified timeslot after despreading. Only the non-orthogonal part of the interference is included in the measurement. The reference point for the ISCP shall be the antenna connector. | | | |
|---------------|--|--|--|--|
| Range/mapping | $\begin{array}{lll} \underline{UTRAN_ISCP_00:} & \underline{UTRAN_ISCP} < -120.0 \mathrm{dBm} \\ \underline{UTRAN_ISCP_01:} & -120.0 \mathrm{dBm} & \leq & \underline{UTRAN_ISCP} < -119.5 \mathrm{dBm} \\ \underline{UTRAN_ISCP_02:} & -119.5 \mathrm{dBm} & \leq & \underline{UTRAN_ISCP} < -119.0 \mathrm{dBm} \\ \end{array}$ | | | |
| | UTRAN_ISCP_81: -79.5dBm ≤ UTRAN_ISCP < -80.0dBm UTRAN_ISCP_82: -80.0dBm ≤ UTRAN_ISCP | | | |

4.4.3 5.2.3 RSSI

| Definition | Received Signal Strength Indicator, the wide-band received power within the UTRAN UL channel bandwidth in a specified timeslot. The reference point for the RSSI shall be the antenna connector. | | |
|---------------|--|--|--|
| Range/mapping | RSSI LEV 0: RSSI < -105.0dBm | | |
| | $\underline{RSSI_LEV 1: -105.0dBm} \leq \underline{RSSI} < -104.5dBm$ | | |
| | $RSSI_LEV 2: -104.5dBm \le RSSI < -103.0dBm$ | | |
| | <u></u> | | |
| | RSSI_LEV 62: -74.5 dBm \leq RSSI < -74.0 dBm | | |
| | RSSI_LEV 63: -74.0dBm≤ RSSI | | |

4.4.4 5.2.4 SIR

| Definition | Signal to Interference Ratio, defined as the RSCP of the DPCH or PUSCH divided by ISCP of the same timeslot. The reference point for the SIR shall be the antenna connector. | | | | |
|---------------|--|--|--|--|--|
| Range/mapping | $UTRAN_SIR_00: SIR < -11.0dB$ | | | | |
| | $UTRAN_SIR_01: -11.0dB \le SIR < -10.5dB$ | | | | |
| | $\overline{UTRAN_SIR_02: -10.5dB} \le SIR < -10.0dB$ | | | | |
| | <u></u> | | | | |
| | $\underline{UTRAN_SIR_61:} 19.0dB \le \qquad \underline{SIR} < 19.5dB$ | | | | |
| | $UTRAN_SIR_62: 19.5dB \le SIR < 20.0dB$ | | | | |
| | $\underline{UTRAN_SIR_63}: 20.0dB \le \qquad \underline{SIR}$ | | | | |

4.4.5 5.2.5 Physical channel BER

| Definition | The physical channel BER is an estimation of the average bit error rate (BER) of a DPCH or | | |
|---------------|--|---------------------------|--|
| | PUSCH before chann | nel decoding of the data. | |
| Range/mapping | Raw_BER_00: | BER < 0.1% | |
| | Raw_BER_01: | 0.1% < BER < 0.3% | |
| | Raw_BER_02: | 0.3% < BER < 0.5% | |
| | Raw_BER_03: | 0.5 % < BER < 1 % | |
| | Raw_BER_04: | 1 % < BER < 2 % | |
| | Raw_BER_05: | 2 % < BER < 4 % | |
| | Raw_BER_06: | 4 % < BER < 6 % | |
| | Raw_BER_07: | 6 % < BER < 8 % | |
| | Raw_BER_08: | 8 % < BER < 10 % | |
| | Raw_BER_09: | 10 % < BER < 12 % | |
| | Raw_BER_10: | 12 % < BER < 14 % | |
| | Raw_BER_11: | 14 % < BER < 16 % | |
| | Raw_BER_12: | 16 % < BER < 18 % | |
| | Raw_BER_13: | 18 % < BER < 20 % | |
| | Raw_BER_14: | 20 % < BER < 30 % | |
| | Raw_BER_15: | 30 % < BER | |

4.4.6 5.2.6 Transport channel BLER

| Definition | | | on evaluating the CRC on each transport block. |
|---------------|---------------|----------|--|
| Range/mapping | BLER_00: | BLER = | = 0 |
| | BLER_01: | -4.05 | \leq Log10(BLER) < -3.90 |
| | BLER_02: | -3.90 | \leq Log10(BLER) < -3.75 |
| | BLER_03: | -3.60 | \leq Log10(BLER) < -3.45 |
| | <u></u> | | |
| | BLER_22: | -0.90 | \leq Log10(BLER) < -0.75 |
| | BLER_23: | -0.75 | \leq Log10(BLER) < -0.60 |
| | BLER 24: | -0.60 | $\leq \text{Log}10(\text{BLER}) < -0.45$ |
| | BLER_25: | -0.45 | \leq Log10(BLER) < -0.30 |
| | BLER_26: | -0.30 | \leq Log10(BLER) < -0.15 |
| | BLER_27: | -0.15 | $\leq \text{Log}10(\text{BLER}) \leq 0.00$ |
| | (BLER_28 to I | BLER_31: | not used) |

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4.4.7 5.2.7 Transmitted carrier power

| Definition | Transmitted carrier power, is the total transmitted power on one DL carrier from one UTRAN access point measured in a timeslot. The reference point for the UTRAN total transmitted power measurement shall be the antenna connector. |
|---------------|---|
| Range/mapping | (UTRAN_TX_POWER_000 to UTRAN_TX_POWER_015: not used, but reserved for further |
| | extension to lower values) |
| | $UTRAN_TX_POWER_016: 0.0dBm \le UTRAN_TX_POWER < 0.5dBm$ |
| | UTRAN_TX_POWER_017: 0.5dBm ≤ UTRAN_TX_POWER < 1.0dBm |
| | UTRAN_TX_POWER_018: 1.0dBm ≤ UTRAN_TX_POWER < 1.5dBm |
| | <u></u> |
| | UTRAN TX POWER 115: 49.0dBm ≤ UTRAN TX POWER < 49.5dBm |
| | UTRAN_TX_POWER_116: 49.5dBm ≤ UTRAN_TX_POWER < 50.0dBm |
| | $\underline{\text{UTRAN_TX_POWER_117:}} 50.0 \text{dBm} \leq \underline{\text{UTRAN_TX_POWER}} < 50.5 \text{dBm}$ |
| | (UTRAN_TX_POWER_118 to UTRAN_TX_POWER_127: not used, but reserved for further |
| | extension to higher values) |
| | |

4.4.8 5.2.8 Transmitted code power

| Definition | Transmitted Code Power, is the transmitted power on one carrier and one channelisation code in one timeslot. The reference point for the transmitted code power measurement shall be the antenna connector at the UTRAN access point cabinet. |
|---------------|---|
| Range/mapping | (UTRAN TX CODE POWER 000 to UTRAN TX CODE POWER 009: not used, butreserved for further extension to lower values)UTRAN TX CODE POWER 010: -10.0dBm≤ CODE POWER < -9.5dBmUTRAN TX CODE POWER 011: -9.5dBm≤ CODE POWER < -8.5dBmUTRAN TX CODE POWER 012: -8.5dBm≤ CODE POWER < -7.5dBmUTRAN TX CODE POWER 121: 45.0dBm ≤ CODE POWER < 45.5dBmUTRAN TX CODE POWER 122: 45.5dBm ≤ CODE POWER < 46.0dBmUTRAN TX CODE POWER 123: 46.0dBm ≤ CODE POWER < 46.5dBm(UTRAN TX CODE POWER 124 to UTRAN TX POWER 127: not used, but reserved for |
| | further extension to higher values) |

4.4.9 5.2.9 RX Timing Deviation

| Definition | The difference of the time of arrival of the UL transmissions in relation to the arrival time of a signal with zero propagation delay. |
|---------------|--|
| | 'RX Timing Deviation' is the time difference TRXdev = TTS - TRXpath in chips, with |
| | TRXpath: time of the reception in the Node B of the first significant uplink path to be used in the detection process |
| | TTS: time of the beginning of the respective slot according to the Node B internal timing |
| Range/mapping | 0 chips 1023 chips in steps of 1 chip (10 bit) |

NOTE: This measurement can be used for timing advance calculation or location services.