TSG-RAN Working Group 1 meeting #9 Dresden, Germany November 30 – December 3, 1999

# TSGR1#9(99)i61

## Agenda item:

Source:	Ericsson
Title:	CR 25.213-007: Update of uplink spreading sections
Document for:	Decision

The current structure of the uplink parts of TS 25.213 V3.0.0 is a bit "un-structured":

- There is no clear distinction between "uplink scrambling codes" and those codes use for DPCCH/DPDCH and PRACH and PCPCH.
- The scrambling codes are described in several different places.
- The channelization code allocation is described in several different places.
- The sub-clause structure for PCPCH preamble code is not aligned with the PRACH preamble code structure.
- Although the spreading for preamble and message parts are very different, this is not reflected in the structure.

Moreover, there is room for improvement of the description in the following areas:

- Explicit information about how the scrambling code is applied to a radio frame is missing.
- The use of both C<sub>ch,n</sub> and C<sub>ch,SF,n</sub> to denote channelization codes is a bit confusing.
- The description of the HPSK transformation is not very clear.
- The distinction between 0/1 and 1/-1 sequences is not very clear.
- There is an error in the terminology used for the allocation of channelization code used for PRACH message data part. The names of the codes are old and should be updated.
- The list with symbol definitions is not consistent and not complete
- The short scrambling code section is a bit unclear:
  - The connection between code number and initialisation of registers is not clear.
  - How the sequence is used together with HPSK is not clearly defined.
  - The letter "c" is used for both the final codes and one of the component codes.
  - "Shift suspend" is not very clear in the figure of the generator.

To fix all this, a CR has been generated that updates the structure and makes the other clarifying changes. It should be noted that no technical change is done, i.e. the chips that come out from the spreading/scrambling is identical to what was assumed before. Also note that this CR only deals with the uplink parts.

It is acknowledged that it might have been better to address smaller issues in separate CRs, however, it was not clear how this would be accomplished together with an structural change in another CR. Hence, only one big CR has been produced.

e.g. for 3GPP use the format TP-99xxx or for SMG, use the format P-99-xxx

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<u>Reason for</u> <u>change:</u>	<ul> <li>Improves the readability of TS 25.213:</li> <li>Makes a clear distinction between "uplink scrambling codes" and those codes use for DPCCH/DPDCH and PRACH and PCPCH.</li> <li>The scrambling codes to use and allocation of channelization codes are now described in a more consistent way (not spread over different sections).</li> <li>The structure of PCPCH preamble code definition is now aligned with the PRACH preamble code definition.</li> <li>Spreading for preamble and message parts are very different. This is now reflected in the structure.</li> <li>Explicit information how scrambling code is applied to radio frame has been added</li> <li>Channelization codes now have consistent names in Figure 1.</li> <li>The description of the HPSK transformation is now clear.</li> <li>The terminology used for the allocation of channelization code used for PRACH message data part has been updated to the new code names.</li> <li>The list with symbol definitions is now consistent and more complete</li> <li>The connection between code number and initialisation of registers for short scrambling codes is now clearly defined.</li> <li>Figure of short sequence generator has been simplified.</li> <li>Codes have been renamed to not interfere with each other.</li> </ul>								
Clauses affec	ted: 3.2, a	Il sub-clauses to cl	ause 4						
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The present document describes spreading and modulation for UTRA Physical Layer FDD mode.

# 2 References

The following documents contain provisions which, through reference in this text, constitute provisions of the present document.

- References are either specific (identified by date of publication, edition number, version number, etc.) or non-specific.
- For a specific reference, subsequent revisions do not apply.
- For a non-specific reference, the latest version applies.

[1] TS 25.201: "Physical layer - general description".

# 3 Definitions, symbols and abbreviations

# 3.1 Definitions

For the purposes of the present document, the following terms and definitions apply.

# 3.2 Symbols

For the purposes of the present document, the following symbols apply:

C <sub>ch,SF,n</sub> :	<i>n</i> :th channelisation code with spreading factor SF
C <sub>scramb</sub> :	scrambling code for uplink
Cpre,n,s:	PRACH preamble code for <i>n</i> :th preamble scrambling code and signature s
C <sub>c-acc,n,s</sub> :	PCPCH access preamble code for <i>n</i> :th preamble scrambling code and signature <i>s</i>
$\underline{C}_{c-cd,n,s}$ :	PCPCH CD preamble code for <i>n</i> :th preamble scrambling code and signature s
C <sub>sig,s</sub> :	PRACH/PCPCH signature code- for signature s
S <sub>longul,n</sub> :	n:th DPCCH/DPDCH long uplinkUL scrambling code-for desicated channels
S <sub>short,n</sub> :	n:th DPCCH/DPDCH short uplink scrambling code
S <sub>r-pre,n</sub> :	<u><i>n</i>:th PRACH preamble scrambling codedd</u>
S <sub>r-msg,n</sub> :	<u>n:th PRACH message scrambling code</u>
$S_{c-acc,n}$ :	<u><i>n</i>:th PCPCH access preamble scrambling code</u>
$S_{c-cd.n}$ :	<u>n:th PCPCH CD preamble scrambling code</u>
$S_{c-msg,n}$ :	<u><i>n</i>:th PCPCH message scrambling code</u>
S <sub>dl.n</sub> :	DL scrambling code
C <sub>sch,n</sub> :	n:th SCH code (primary or secondary)
C <sub>psc</sub> :	PSC code
$C_{ssc,n}$ :	n:th SSC code

# 3.3 Abbreviations

For the purposes of the present document, the following abbreviations apply:

AICH	Acquisition Indicator Channel
AP	Access Preamble

BCH	Broadcast Control Channel
CCPCH	Common Control Physical Channel
CD	Collision Detection
CPCH	Common Packet Channel
DCH	Dedicated Channel
DPCH	Dedicated Physical Channel
DPCCH	Dedicated Physical Control Channel
DPDCH	Dedicated Physical Data Channel
FDD	Frequency Division Duplex
Mcps	Mega Chip Per Second
OVSF	Orthogonal Variable Spreading Factor (codes)
PDSCH	Physical Dedicated Shared Channel
PICH	Page Indication Channel
PRACH	Physical Random Access Channel
PSC	Primary Synchronisation Code
RACH	Random Access Channel
SCH	Synchronisation Channel
SSC	Secondary Synchronisation Code
SF	Spreading Factor
UE	User Equipment

# 4 Uplink spreading and modulation

# 4.1 Overview

Spreading is applied to the physical channels. It consists of two operations. The first is the channelization operation, which transforms every data symbol into a number of chips, thus increasing the bandwidth of the signal. The number of chips per data symbol is called the Spreading Factor (SF). The second operation is the scrambling operation, where a scrambling code is applied to the spread signal.

With the channelization, data symbols on so-called I- and Q-branches are independently multiplied with an OVSF code. With the scrambling operation, the resultant signals on the I- and Q-branches are further multiplied by complex-valued scrambling code, where I and Q denote real and imaginary parts, respectively.

# 4.2 Spreading

# 4.2.1 Uplink Dedicated Physical Channels (uplink DPCDCH/DPDCCH)

Figure 1 illustrates the principle of the uplink spreading of DPCCH and DPDCHs. The binary DPCCH and DPDCHs to be spread are represented by real-valued sequences, i.e. the binary value "0" is mapped to the real value +1, while the binary value "1" is mapped to the real value -1. The DPCCH is spread to the chip rate by the channelization code  $\underline{c}_{\underline{c}} \underline{C}_{\underline{ch,0}}$ , while the *n*:th DPDCH called DPDCH<sub>n</sub> is spread to the chip rate by the channelization code  $\underline{c}_{\underline{d,n}} \underline{C}_{\underline{ch,n}}$ . One DPCCH and up to six parallel DPDCHs can be transmitted simultaneously, i.e.  $0 \le n \le 6$ .



DPDCH

DPDCH

DPDCH



9



Figure 1: Spreading/modulation for uplink DPCCH and DPDCHs

After channelization, the real-valued spread signals are weighted by gain factors,  $\beta_c$  for DPCCH and  $\beta_d$  for all DPDCHs.

At every instant in time, at least one of the values  $\beta_c$  and  $\beta_d$  has the amplitude 1.0. The  $\beta$ -values are quantized into 4 bit words. The quantization steps are given in table 1.

Signalling values for $eta_c$ and $eta_d$	Quantized amplitude ratios $\beta_c \ \text{ and } \beta_d$
15	1.0
14	0.9333
13	0.8666
12	0.8000
11	0.7333
10	0.6667
9	0.6000
8	0.5333
7	0.4667
6	0.4000
5	0.3333
4	0.2667
3	0.2000
2	0.1333
1	0.0667
0	Switch off

#### Table 1: The quantization of the gain parameters

After the weighting, the stream of real-valued chips on the I- and Q-branches are then summed and treated as a complex-valued stream of chips. This complex-valued signal is then scrambled by the complex-valued scrambling code  $\underline{S}_{\text{long,n}}$  or  $\underline{S}_{\text{short,n}}$ , depending on if long or short scrambling codes are used  $\underline{C}_{\text{scramb}}$ . After pulse-shaping (described in [1]), QPSK modulation is performed. The scrambling code is applied aligned with the radio frames, i.e. the first scrambling chip corresponds to the beginning of a radio frame.

# 4.2.2 PRACH

## 4.2.2.1 PRACH preamble part

The PRACH preamble part consist of a complex-valued  $code_3$ , that after pulse shaping is transmitted using QPSK. The preamble code in described in section 4.3.3.1.

## 4.2.2.2 PRACH message part

The spreading and modulation of the message part of the PRACH message part is basically the same as for the uplink dedicated physical channels.

Figure 2 illustrates the principle of the spreading and scrambling of the PRACH message part, consisting of data and control parts. The binary control and data parts to be spread are represented by real-valued sequences, i.e. the binary value "0" is mapped to the real value +1, while the binary value "1" is mapped to the real value -1. The control part is spread to the chip rate by the channelization code  $c_c$ , while the data part is spread to the chip rate by the channelization code  $c_d$ .



#### Figure 2: Spreading and scrambling of PRACH message part

After channelization, the real-valued spread signals are weighted by gain factors,  $\beta_c$  for the control part and  $\beta_d$  for the data part. At every instant in time, at least one of the values  $\beta_c$  and  $\beta_d$  has the amplitude 1.0. The  $\beta$ -values are quantized into 4 bit words. The quantization steps are given in section 4.2.1.

After the weighting, the stream of real-valued chips on the I- and Q-branches are treated as a complex-valued stream of chips. This complex-valued signal is then scrambled by the complex-valued scrambling code  $S_{r-msg,n}$ . After pulse-shaping (described in [1]), QPSK modulation is performed. The scrambling code is applied aligned with the radio frames, i.e. the first scrambling chip corresponds to the beginning of a radio frame.

# 4.2.3 PCPCH

## 4.2.3.1 PCPCH preamble part

The PCPCH preamble part consists of a complex-valued  $code_a$ , that after pulse shaping is transmitted using QPSK. The preamble code in described in section 4.3.4.3.

## 4.2.3.2 PCPCH message part

Figure 3 illustrates the principle of the spreading and scrambling of the PCPCH message part, consisting of data and control parts. The binary control and data parts to be spread are represented by real-valued sequences, i.e. the binary value "0" is mapped to the real value +1, while the binary value "1" is mapped to the real value -1. The control part is spread to the chip rate by the channelization code  $c_c$ , while the data part is spread to the chip rate by the channelization code  $c_d$ .



Figure 3: Spreading and scrambling of PCPCH message part

After channelization, the real-valued spread signals are weighted by gain factors,  $\beta_c$  for the control part and  $\beta_d$  for the data part. At every instant in time, at least one of the values  $\beta_c$  and  $\beta_d$  has the amplitude 1.0. The  $\beta$ -values are quantized into 4 bit words. The quantization steps are given in section 4.2.1.

After the weighting, the stream of real-valued chips on the I- and Q-branches are treated as a complex-valued stream of chips. This complex-valued signal is then scrambled by the complex-valued scrambling code  $S_{c-msg,n}$ . After pulse-shaping (described in [1]), QPSK modulation is performed. The scrambling code is applied aligned with the radio frames, i.e. the first scrambling chip corresponds to the beginning of a radio frame.

# 4.3 Code generation and allocation

# 4.3.1 Channelization codes

# 4.3.1.1 Code definition

The channelization codes of figure 1 are Orthogonal Variable Spreading Factor (OVSF) codes that preserve the orthogonality between a user's different physical channels. The OVSF codes can be defined using the code tree of figure 4.



#### Figure 4: Code-tree for generation of Orthogonal Variable Spreading Factor (OVSF) codes

In figure 4, the channelization codes are uniquely described as  $C_{ch,SF,k}$ , where SF is the spreading factor of the code and *k* is the code number,  $0 \le k \le SF-1$ .

Each level in the code tree defines channelization codes of length SF, corresponding to a spreading factor of SF in figure 4.

The generation method for the channelization code is defined as:

$$\begin{split} \mathbf{C}_{ch,1,0} &= \mathbf{1}, \\ \begin{bmatrix} C_{ch,2,0} \\ C_{ch,2,1} \end{bmatrix} = \begin{bmatrix} C_{ch,1,0} & C_{ch,1,0} \\ C_{ch,1,0} & -C_{ch,1,0} \end{bmatrix} = \begin{bmatrix} \mathbf{1} & \mathbf{1} \\ \mathbf{1} & -\mathbf{1} \end{bmatrix} \\ \begin{bmatrix} C_{ch,2}^{(n+1),0} \\ C_{ch,2}^{(n+1),1} \\ C_{ch,2}^{(n+1),2} \\ C_{ch,2}^{(n+1),2} \\ \vdots \\ C_{ch,2}^{(n+1),2(n+1),2} \\ C_{ch,2}^{(n+1),2(n+1),2} \end{bmatrix} = \begin{bmatrix} C_{ch,2^{n},0} & C_{ch,2^{n},0} \\ C_{ch,2^{n},0} & -C_{ch,2^{n},0} \\ C_{ch,2^{n},1} & C_{ch,2^{n},1} \\ \vdots & \vdots \\ C_{ch,2^{n},2^{n},1} & -C_{ch,2^{n},1} \\ \vdots & \vdots \\ C_{ch,2^{n},2^{n},1} & -C_{ch,2^{n},2^{n},1} \\ C_{ch,2^{n},2^{n},1} & -C_{ch,2^{n},2^{n},1} \\ \end{bmatrix}$$

The leftmost value in each channelization code word corresponds to the chip transmitted first in time.

## 4.3.1.2 Code allocation for DPCCH/DPDCH

For the DPCCH and DPDCHs the following applies:

- The DPCCH is always spread by code  $\underline{c}_{c} \underline{C}_{eh,0} = C_{ch,256,0.}$
- When only one DPDCH is to be transmitted, DPDCH<sub>1</sub> is spread by code  $\underline{c_{d,1}} = C_{ch,SF,k}$  where SF is the spreading factor of DPDCH<sub>1</sub> and  $k_{-} = SF_{d,+} / 4$
- When more than one DPDCH is to be transmitted, all DPDCHs have spreading factors equal to 4. DPDCH<sub>n</sub> is spread by the the code  $\underline{c_{d,n}} = C_{ch,4,k}$ , where k = 1 if  $n \in \{1, 2\}$ , k = 3 if  $n \in \{3, 4\}$ , and k = 2 if  $n \in \{5, 6\}$ .

# 4.3.1.3 Code allocation for PRACH message part

The preamble signature  $s, 1 \le s \le 16$ , points to one of the 16 nodes in the code-tree that corresponds to channelization codes of length 16. The sub-tree below the specified node is used for spreading of the message part. The control part is spread with the channelization code  $c_c$  (as shown in section 4.2.2.2) of spreading factor 256 in the lowest branch of the sub-tree, i.e.  $c_c = C_{ch,256,m}$  where m = 16(s - 1) + 15. The data part uses any of the channelization codes from spreading factor 32 to 256 in the upper-most branch of the sub-tree. To be exact, the data part is spread by channelization code  $c_d = C_{ch,SF,m}$  and SF is the spreading factor used for the data part and  $m = SF \times (s - 1)/16$ .

# 4.3.1.4 Code allocation for PCPCH message part

The signature in the preamble specifies one of the 16 nodes in the code-tree that corresponds to channelization codes of length 16. The sub-tree below the specified node is used for spreading of the message part. The control part is always spread with a channelization code of spreading factor 256. The code is chosen from the lowest branch of the sub-tree. The data part may use channelization codes from spreading factor 4 to 64. A UE is allowed to increase its spreading factor during the message transmission by choosing any channelization code from the uppermost branch of the sub-tree code. For channelization codes with spreading factors less that 16, the node is located on the same sub-tree as the channelization code of the access preamble.

# 4.3.2 Scrambling codes

## 4.3.2.1 General

All uplink physical channels are subjected to scrambling with a complex-valued scrambling code. The DPCCH/DPDCH may be scrambled by either long or short scrambling codes, defined in section 4.3.2.4. The PRACH message part is scrambled with a long scrambling code, defined in section 4.3.2.5. Also the PCPCH message part is scrambled with a long scrambling code, defined in section 4.3.2.6.

There are  $2^{24}$  long and  $2^{24}$  short uplink scrambling codes. Uplink scrambling codes are assigned by higher layers.

The long scrambling code is built from constituent long sequences defined in section 4.3.2.2, while the constituent short sequences used to build the short scrambling code are defined in section 4.3.2.3.

There are  $2^{24}$ -uplink scrambling codes. All uplink channels shall use either short or long scrambling codes, except for the PRACH, for which only the long scrambling code is used. Both short and long scrambling codes are represented with complex-value.

The uplink scrambling generator (either short or long) shall be initialised by a 24 bit value.

Both short and long uplink scrambling codes are formed as follows:

 $-S_{ul,n} = C_{scramb,n}$ 

where

 $--C_{\text{scramb,n}} = c_1(w_0 + jc_2'w_1)$ 

where  $w_0$  and  $w_1$  are chip rate sequences defined as repetitions of:

 $w_0 = \{1 \ 1\}$ 

 $-w_{+} = \{1, \dots, 1\}$ 

Also,  $c_1$  is a real chip rate code, and  $c_2$ ' is a decimated version of the real chip rate code  $c_2$ -

With a decimation factor 2,  $c_2$ ' is given as:

 $c_2'(2k) = c_2'(2k+1) = c_2(2k), k=0,1,2...$ 

The constituent codes  $c_1$  and  $c_2$  are formed differently for the short and long scrambling codes as described in sections 4.3.2.2 and 4.3.2.3.

### 4.3.2.2 Long scrambling sequencecode

The long scrambling sequences  $c_{long,1,n}$  and  $c_{long,2,n}$  eodes are formed as described in section 4.3.2, where  $e_1$ -and  $e_2$ -are constructed from the position wise modulo 2 sum of 38400 chip segments of two binary *m*-sequences generated by means of two generator polynomials of degree 25. Let *x*, and *y* be the two *m*-sequences respectively. The *x* sequence is constructed using the primitive (over GF(2)) polynomial  $X^{25}+X^3+I$ . The *y* sequence is constructed using the polynomial  $X^{25}+X^3+X^2+X+I$ . The resulting sequences thus constitute segments of a set of Gold sequences.

The code, sequence  $c_{long,2,n}$ , used in generating the quadrature component of the complex spreading code is a 16,777,232 chip shifted version of the sequence code,  $c_{long,1,n}$ , used in generating the in phase component.

The uplink scrambling code word has a period of one radio frame.

Let  $n_{23} \dots n_0$  be the 24 bit binary representation of the scrambling <u>sequencecode</u> number *n*-(decimal) with  $n_0$  being the least significant bit. The *x* sequence depends on the chosen scrambling <u>sequencecode</u> number *n* and is denoted  $x_n$ , in the sequel. Furthermore, let  $x_n(i)$  and y(i) denote the *i*:th symbol of the sequence  $x_n$  and *y*, respectively

The *m*-sequences  $x_n$  and *y* are constructed as:

Initial conditions:

 $x_n(0)=n_0$ ,  $x_n(1)=n_1$ , ... = $x_n(22)=n_{22}$ ,  $x_n(23)=n_{23}$ ,  $x_n(24)=1$ 

y(0)=y(1)=...=y(23)=y(24)=1

Recursive definition of subsequent symbols:

 $x_n(i+25) = x_n(i+3) + x_n(i) \text{ modulo } 2, i=0,..., 2^{25}-27,$ 

 $y(i+25) = y(i+3)+y(i+2) + y(i+1) + y(i) \text{ modulo } 2, i=0,..., 2^{25}-27.$ 

The definition of the *n*:th scrambling code word for the in phase and quadrature components follows as (the left most index correspond to the chip scrambled first in each radio frame):

Define the binary Gold sequence  $z_n$  by

 $z_n(i) = x_n(i) + y(i) \mod 2, i = 0, 1, 2, \dots, 2^{25}-2,$ 

 $e_{i_{1,n}} = < x_n(0) + y(0), \ x_n(1) + y(1), \ \dots, x_n(N-1) + y(N-1) >,$ 

 $e_{2,n} = < x_n(M) + y(M), \ x_n(M+1) + y(M+1), \ \dots, \ x_n(M+N-1) + y(M+N-1) >,$ 

again all sums being modulo 2 additions.

Where N is the period in chips and M = 16,777,232.

These binary code words are converted to real valued sequences by the transformation  $0^{\circ} > +1^{\circ}, 1^{\circ}$ 

The real valued Gold sequence  $Z_n$  is defined by

$$Z_n(i) = \begin{cases} +1 & \text{if } z_n(i) = 0\\ -1 & \text{if } z_n(i) = 1 \end{cases} \quad \text{for } i = 0, 1, \dots, 2^{25} - 2.$$

Now, the real-valued long scrambling sequences clong,1,n and clong,2,n are defined as follows:

 $\underline{c}_{\text{long},1,n}(i) = \underline{Z}_n(i), \ i = 0, 1, 2, \dots, 2^{25} - 2 \text{ and}$ 

 $\underline{c}_{long,2,n}(i) = Z_n((i + 16777232) \text{ modulo } (2^{25} - 1)), i = 0, 1, 2, ..., 2^{25} - 2.$ 

Finally, the complex-valued long scrambling sequence Clong, n, is defined as

$$C_{long,n}(i) = \begin{cases} c_{long,1,n}(i)(1+jc_{long,2,n}(i)) & \text{if } i \text{ is even} \\ c_{long,1,n}(i)(1-jc_{long,2,n}(i-1)) & \text{if } i \text{ is odd} \end{cases}$$

where  $i = 0, 1, ..., 2^{25} - 2$ .



#### Figure 5: Configuration of uplink scrambling sequencecode generator

## 4.3.2.3 Short scrambling <u>sequencecode</u>

The short scrambling sequences  $c_{\text{short},1,n}(i)$  and  $c_{\text{short},2,n}(i)$  are defined from a sequence from the family of periodically extended S(2) codes.

Let  $n_{23}n_{22}...n_0$  be the 24 bit binary representation of the code number *n*.

The *n*:th quaternary S(2) sequence  $z_n(i)$ ,  $0 \le n \le 16777215$ , is obtained by modulo 4 addition of three sequences, a quaternary sequence a(i) and two binary sequences b(i) and d(i), where the initial loading of the three sequences is determined from the code number *n*. The sequence  $z_n(i)$  of length 255 is generated according to the following relation:

 $\underline{z_n(i)} = a(i) + 2b(i) + 2d(i) \mod 4, i = 0, 1, \dots, 254,$ 

where the quaternary sequence a(i) is generated recursively by the polynomial  $g_0(x) = x^8 + x^5 + 3x^3 + x^2 + 2x + 1$  as

 $\underline{a(0)} = 2n_0 + 1 \text{ modulo } 4,$ 

 $a(i) = 2n_i \mod 4, i = 1, 2, ..., 7,$ 

 $\underline{a(i)} = 3a(i-3) + a(i-5) + 3a(i-6) + 2a(i-7) + 3a(i-8) \mod 4, i = 8, 9, \dots, 254,$ 

and the binary sequence b(i) is generated recursively by the polynomial  $g_1(x) = x^8 + x^7 + x^5 + x + 1$  as

<u> $b(i) = n_{8+i} \mod 2, i = 0, 1, ..., 7,$ </u>

 $\underline{b(i)} = \underline{b(i-1)} + \underline{b(i-3)} + \underline{b(i-7)} + \underline{b(i-8)} \mod 2, i = 8, 9, \dots, 254,$ 

and the binary sequence c(i) is generated recursively by the polynomial  $g_2(x) = x^8 + x^7 + x^5 + x^4 + 1$  as

 $\underline{d(i)} = \underline{n_{16+i}} \mod 2, i = 0, 1, \dots, 7,$ 

 $d(i) = d(i-1) + d(i-3) + d(i-4) + d(i-8) \mod 2, i = 8, 9, \dots, 254.$ 

The sequence  $z_n(i)$  is extended to length 256 chips by setting  $z_n(255) = z_n(0)$ .

The mapping from  $z_n(i)$  to the real-valued binary sequences  $c_{\text{short},1,n}(i)$  and  $c_{\text{short},2,n}(i)$ , i = 0, 1, ..., 255 is defined in Table 2.

<u>z<sub>n</sub>(i)</u>	<u>C<sub>short,1,n</sub>(1)</u>	<u>Cshort,2,n(1)</u>
<u>0</u>	<u>+1</u>	<u>+1</u>
<u>1</u>	<u>-1</u>	<u>+1</u>
2	-1	<u>-1</u>
3	+1	<u>-1</u>

Table 2. Mapping from  $z_n(i)$  to  $c_{short,1,n}(i)$  and  $c_{short,2,n}(i)$ ,  $i = 0, 1, \dots, 255$ .

Finally, the complex-valued short scrambling sequence C<sub>short, n2</sub> is defined as

$$C_{short,n}(i) = \begin{cases} c_{short,1,n}(i \mod 256)(1 + jc_{short,2,n}(i \mod 256)) & \text{if } i \text{ is even} \\ c_{short,1,n}(i \mod 256)(1 - jc_{short,2,n}((i-1) \mod 256)) & \text{if } i \text{ is odd} \end{cases}$$

where i = 0, 1, 2, ...

An implementation of the short scrambling sequence generator for the 255 chip sequence to be extended by one chip is shown in Figure 6.



#### Figure 6. Uplink short scrambling sequence generator for 255 chip sequence.

The short scrambling codes are formed as described in section 4.3.2.1, where c1 and c2 are the real and imaginary components of a complex sequence from the family of periodically extended S(2) codes.

The uplink short codes  $S_*(n)$ , n=0,1,...255, of length 256 chips are obtained by one chip periodic extension of S(2) sequences of length 255. It means that the first chip ( $S_*(0)$ ) and the last chip ( $S_*(255)$ ) of any uplink short scrambling code are the same.

The quaternary S(2) sequence  $z_{+}(n)$ ,  $0 \le v \le 16,777,215$ , of length 255 is obtained by modulo 4 addition of three sequences, a quaternary sequence  $a_{+}(n)$  and two binary sequences  $b_{+}(n)$  and  $c_{+}(n)$ , according to the following relation:

 $z_{t}(n) = a_{t}(n) + 2^{t}b_{t}(n) + 2^{t}e_{t}(n) \pmod{4}, \quad n = 0, 1, \dots, 254.$ 

The user index v determines the indexes r, s, and t of the constituent sequences in the following way:

 $\frac{v = t^{-2^{16}} + s^{-2^8} + r}{r^{-2^8} + r}$ 

 $r=0, 1, 2, \dots, 255,$ 

 $-s = 0, 1, 2, \dots, 255,$ 

 $t = 0, 1, 2, \dots, 255.$ 

The quaternary sequence  $a_{\ell}(n)$  is generated by the recursive generator  $G_0$  defined by the polynomial

 $--g_0(x) = x^8 + x^5 + 3x^3 + x^2 + 2x + 1$  as

 $--a_{r}(n)=3.a_{r}(n-3)+1.a_{r}(n-5)+3.a_{r}(n-6)+2.a_{r}(n-7)+3.a_{r}(n-8) \pmod{4}.$ 

$$n = 8...254$$

The binary sequence  $b_*(n)$  is generated by the recursive generator  $G_1$ -defined by the polynomial

$$-g_{1}(x) = x^{8} + x^{7} + x^{5} + x + 1 - as$$

 $-b_s(n)=b_s(n-1)+b_s(n-3)+b_s(n-7)+b_s(n-8)\pmod{2}$ .

The binary sequence  $c_t(n)$  is generated by the recursive generator  $G_2$  defined by the polynomial

- $-g_2(x) = x^8 + x^7 + x^5 + x^4 + 1 as$
- $c_t(n) = c_t(n-1) + c_t(n-3) + c_t(n-4) + c_t(n-8) \pmod{2}.$

An implementation of the short scrambling code generator is shown in figure 6. The initial states for the binary generators  $G_1$  and  $G_2$  are the two 8 bit words representing the indexes *s* and *t* in the 24 bit binary representation of the user index  $\nu$ , as it is shown in figure 7.

The initial state for the quaternary generator  $G_{\theta}$  is according to figure 7 obtained after the transformation of 8-bit word representing the index *r*. This transformation is given by

 $a_r(0) = 2v(0)+1 \pmod{4}, \quad a_r(n) = 2v(n) \pmod{4}, \quad n = 1,...,7.$ 

The complex quadriphase sequence  $S_{+}(n)$  is obtained from quaternary sequence  $z_{+}(n)$  by the mapping function given in table 2.

The Re{Sv(n)} and Im{Sv(n)} of the S(2) code are the pair of two binary sequences corresponding to input binary sequences  $c_1$  and  $c_2$  respectively described in 4.3.2.

#### Table 2: Mapping between S<sub>4</sub>(n) and z<sub>4</sub>(n)

<del>zv(n)</del>	Sv(n)
θ	<del>+1 + j1</del>
4	<del>-1 + j1</del>
2	<del>-1 - j1</del>
3	<del>+1 - j1</del>



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#### Figure 6: Uplink short scrambling code generator



#### Figure 7: Uplink short scrambling code generator state initialisation

### 4.3.2.4 DPCCH/DPDCH scrambling code

The code used for scrambling of the uplink DPCCH/DPDCH may be of either long or short type. When the scrambling code is formed, different consituent codes are used for the long and short type as defined below.

The *n*:th long uplink scrambling code for DPCCH/DPDCH, denoted S<sub>long, n</sub>, is defined as

<u> $S_{long,n}(i) = C_{long,n}(i), i = 0, 1, ..., 38399,</u>$ </u>

where the lowest index corresponds to the chip transmitted first in time and Clong,n is defined in section 4.3.2.2.

The n:th short uplink scrambling code for DPCCH/DPDCH, denoted S<sub>short. n</sub>, is defined as

 $\underline{S}_{\text{short,n}}(i) = \underline{C}_{\text{short,n}}(i), i = 0, 1, ..., 38399,$ 

where the lowest index corresponds to the chip transmitted first in time and C<sub>short,n</sub> is defined in section 4.3.2.3.

### 4.3.2.5 PRACH message part scrambling code

The scrambling code used for the PRACH message part is 10 ms long, cell-specific and has a one-to-one correspondence to the scrambling code used for the preamble part.

The *n*:th PRACH message part scrambling code, denoted  $S_{r-msg,n}$  is based on the long scrambling sequence and is defined as

 $\underline{\mathbf{S}}_{\text{r-msg,n}}(i) = \underline{\mathbf{C}}_{\text{long,n}}(i + 4096), \ i = 0, 1, ..., 38399$ 

where the lowest index corresponds to the chip transmitted first in time and C<sub>long,n</sub> is defined in section 4.3.2.2.

#### 4.3.2.6 PCPCH message part scrambling code

The scrambling code used for the PCPCH message part is 10 ms long, cell-specific and has a one-to-one correspondence to the scrambling code used for the preamble part.

The *n*:th PCPCH message part scrambling code, denoted  $S_{c-msg,n}$  is based on the long scrambling sequence and is defined as

 $\underline{\mathbf{S}}_{\text{r-msg,n}}(i) = \underline{\mathbf{C}}_{\text{long,n}}(i+8192), \ i = 0, 1, \dots, 38399$ 

where the lowest index corresponds to the chip transmitted first in time and Clong, is defined in section 4.3.2.2.

In the case when the access resources are shared between the RACH and CPCH, then S<sub>c-msg,n</sub> is defined as

<u> $S_{r-msg,n}(i) = C_{long,n}(i + 4096), i = 0, 1, ..., 38399</u>$ </u>

where the lowest index corresponds to the chip transmitted first in time and Clong, is defined in section 4.3.2.2.

NOTE: Use of short scrambling code for CPCH message part is ffs.

# 4.3.3 Random access PRACH preamble codes

#### 4.3.3.1 Preamble Codes

#### 4.3.3.1.1 Preamble code construction

The random access preamble code  $C_{pre,n,\underline{s}}$ , is a complex valued sequence. It is built from a preamble scrambling code  $S_{r-pre,n}$  and a preamble signature  $C_{sig,s}$  as follows:

$$C_{\text{pre,n,s}}(k) = S_{\text{r-pre,n}}(k) \times C_{\text{sig,s}}(k) \times e^{j(\frac{\pi}{4} + \frac{\pi}{2}k)}, k = 0, 1, 2, 3, ..., 4095,$$

where k=0 corresponds to the chip transmitted first in time and  $S_{r-pre,n}$  and  $C_{sig,s}$  are defined in 4.3.3.4.2 and 4.3.3.22 below respectively.

### 4.3.3.1.2 Preamble scrambling code

The scrambling code for the <u>PRACH</u> preamble part is <u>constructed from the long scrambling sequences</u> as follows.

The code generating method is the same as for the real part of the uplink long scrambling codes on dedicated channels, see 4.3.2.1 and 4.3.2.2. Only the first 4096 chips of the code are used for preamble scrambling.

The definition of the *n*:th <u>preamble scrambling</u> code sequence is defined asfollows (the left most index correspond to the chip transmitted first in each slot):

 $\mathbf{S}_{r-pre,n}(i) = \underline{c}_{long,1,n}(i), i = 0, 1, \dots, 4095, \mathbf{Re}\left\{C_{scramb,n}\right\}$ , for chip indexes 0...4095 of  $C_{scramb,n}$ 

where the sequence  $c_{long,1,n}$  is defined in section 4.3.2.2.

### 4.3.3.<u>3</u>2 Preamble signature

The preamble signature corresponding to a signature s consists of 256 repetitions of a length 16 signature  $P_s(n)$ , n=0...15. This is defined as follows:

 $C_{sig,s}(i) = P_s(i \text{ modulo } 16), i = 0, 1, ..., 4095.$ 

The signature  $P_s(n)$  is from the set of 16 Hadamard codes of length 16. These are listed in table 3.

Preamble	Value of <i>n</i>															
signature	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
P <sub>0</sub> (n)	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
P₁(n)	1	-1	1	-1	1	-1	1	-1	1	-1	1	-1	1	-1	1	-1
P <sub>2</sub> (n)	1	1	-1	-1	1	1	-1	-1	1	1	-1	-1	1	1	-1	-1
P₃(n)	1	-1	-1	1	1	-1	-1	1	1	-1	-1	1	1	-1	-1	1
P₄(n)	1	1	1	1	-1	-1	-1	-1	1	1	1	1	-1	-1	-1	-1
P₅(n)	1	-1	1	-1	-1	1	-1	1	1	-1	1	-1	-1	1	-1	1
P <sub>6</sub> (n)	1	1	-1	-1	-1	-1	1	1	1	1	-1	-1	-1	-1	1	1
P <sub>7</sub> (n)	1	-1	-1	1	-1	1	1	-1	1	-1	-1	1	-1	1	1	-1
P <sub>8</sub> (n)	1	1	1	1	1	1	1	1	-1	-1	-1	-1	-1	-1	-1	-1
P₀(n)	1	-1	1	-1	1	-1	1	-1	-1	1	-1	1	-1	1	-1	1
P <sub>10</sub> (n)	1	1	-1	-1	1	1	-1	-1	-1	-1	1	1	-1	-1	1	1
P <sub>11</sub> (n)	1	-1	-1	1	1	-1	-1	1	-1	1	1	-1	-1	1	1	-1
P <sub>12</sub> (n)	1	1	1	1	-1	-1	-1	-1	-1	-1	-1	-1	1	1	1	1
P <sub>13</sub> (n)	1	-1	1	-1	-1	1	-1	1	-1	1	-1	1	1	-1	1	-1
P <sub>14</sub> (n)	1	1	-1	-1	-1	-1	1	1	-1	-1	1	1	1	1	-1	-1
P <sub>15</sub> (n)	1	-1	-1	1	-1	1	1	-1	-1	1	1	-1	1	-1	-1	1

#### Table 3: Preamble signatures

#### 4.3.3.3 Channelization codes for the message part

The preamble signature  $s, 1 \le s \le 16$ , points to one of the 16 nodes in the code-tree that corresponds to channelization codes of length 16. The sub-tree below the specified node is used for spreading of the message part. The control part is spread with the channelization code  $c_e$  (as shown in section 4.2.2) of spreading factor 256 in the lowest branch of the sub-tree, i.e.  $e_e = C_{eh,256,m}$  where m = 16(s - 1) + 15. The data part uses any of the channelization codes from spreading factor 32 to 256 in the upper-most branch of the sub-tree. To be exact, the data part is spread by channelization code  $C_{eh,d}$ , where  $C_{eh,d} = c_{SF,m}$  and SF is the spreading factor used for the data part and  $m = SF \times (s - 1)/16$ .

#### 4.3.3.4 Scrambling code for the message part

In addition to spreading, the message part is also subject to scrambling with a 10 ms complex code. The scrambling code is cell-specific and has a one-to-one correspondence to the scrambling code used for the preamble part.

 $-S_{r-msg,n} = C_{seramb,n}$ , for chip indexes 4096...42495 of  $C_{seramb,n}$ 

The generation of these codes is explained in 4.3.2.2. The mapping of these codes to provide a complex scrambling code is also the same as for the dedicated uplink channels and is described in 4.3.2.1.

# 4.3.4 Common packet channel PCPCH preamble codes

#### 4.3.4.1 Access preamble

#### 4.3.4.1.1 <u>Access</u> Ppreamble code construction

Similar to PRACH access preamble codes, the PCPCH access -preamble codes  $C_{c-acc,n,s}$ , are complex valued sequences. The PCPCH access preamble codes are built from the preamble scrambling codes  $S_{c-acc,n}$  and a preamble signature  $C_{sig,s}$  as follows:

$$C_{\text{c-acc,n,s}}(k) = S_{\text{c-acc,n}}(k) \times C_{\text{sig,s}}(k) \times e^{j(\frac{\pi}{4} + \frac{\pi}{2}k)}, \ k = 0, 1, 2, 3, ..., 4095,$$

----where S<sub>c-acc,n</sub> and C<sub>sig,s</sub> are is defined in sections 4.3.4.1.2 and 4.3.4.1.3, below respectively.

### 4.3.4.1.2 Access preamble scrambling code

The access preamble scrambling code generation is done in a way similar to that of PRACH-with a difference of the initialisation of the x m-sequence in section 4.3.2.2. The long code  $C_{scramb,n}$  (as described in sections 4.3.2.1 and 4.3.2.2) for the in-phase component is used directly on both in phase and quadrature branches without offset between branches. Only the first 4096 chips of the code are used for preamble scrambling. In the case when the access resources are shared between the RACH and CPCH, the scrambling codes used in the RACH preamble will be used for the CPCH preamble as well.

The <u>*n*:th</u> definition of the <u>P</u>CPCH access preamble scrambling code <u>is defined as</u>sequence follows (the left most index correspond to the chip transmitted first in each slot):

 $\mathbf{S}_{c_{\underline{-}acc,n}(i)} = \underline{c}_{\underline{long,1,n}(i)}, i = 0, 1, \dots, 4095, \overline{\text{Re}\{C_{\underline{scramb,n}}\}}, \text{for chip indexes } 0 \dots 4095 \text{ of } C_{\underline{scramb,n}}$ 

where the sequence  $c_{long,1,n}$  is defined in section 4.3.2.2.

In the case when the access resources are shared between the PRACH and PCPCH, the scrambling codes used in the PRACH preamble are used for the PCPCH preamble as well.

#### 4.3.4.1.3 Access preamble signature

The access preamble part of the CPCH-access burst carries one of the sixteen different orthogonal complex signatures identical to the ones used by the preamble part of the random-access burst.

# 4.3.4.2 CD Ppreamble

#### 4.3.4.2.1 CD <u>p</u>Preamble code construction

Similar to PRACH access preamble codes, the PCPCH CD preamble codes  $C_{c-cd,n,s}$  are complex valued sequences. The PCPCH CD preamble codes are built from the preamble scrambling codes  $S_{c-cd,n}$  and a preamble signature  $C_{sig,s}$  as follows:

$$C_{c-cd,n,s}(k) = S_{c-cd,n}(k) \times C_{sig,s}(k) \times e^{j(\frac{\pi}{4} + \frac{\pi}{2}k)}, k = 0, 1, 2, 3, ..., 4095,$$

----where S<sub>c-cd,n</sub> and C<sub>sig,s</sub> is defined in sections 4.3.4.2.2 and 4.3.4.2.3 below respectively.

#### 4.3.4.2.2 CD preamble scrambling code

The <u>PCPCH CD</u> preamble scrambling code is derived from the same scrambling code used in the <u>PCPCH</u> access preamble. The long code  $C_{\text{scramb,n}}$  (as described in sections 4.3.2.1 and 4.3.2.2) for the in-phase component is used directly on both in phase and quadrature branches without offset between branches. The 4096 chips of the code from 4096 to 8191 are used for CPCH CD preamble scrambling.

The <u>*n*:th</u> definition of the CPCH CD access preamble scrambling code is defined as sequence follows (the left most index correspond to the chip transmitted first in each slot):

 $S_{c-cd,n}(i) = c_{long,1,n}(i + 4096), i = 0, 1, ..., 4095, Re{C_{scramb,n}}, for chip indexes 4096...8191 of C_{scramb,n}$ 

where the sequence  $c_{long,1,n}$  is defined in section 4.3.2.2.

In the case when the access resources are shared between the RACH and CPCH, the scrambling codes used in the RACH preamble will be used for the CPCH CD preamble as well.

#### 4.3.4.2.3 CD preamble signature

The CD-preamble part of the CPCH-access burst carries one of sixteen different orthogonal complex signatures identical to the ones used by the preamble part of the random-access burst.

#### 4.3.4.3 CPCH preamble signatures

#### 4.3.4.3.1 Access preamble signature

The access preamble part of the CPCH-access burst carries one of the sixteen different orthogonal complex signatures identical to the ones used by the preamble part of the random-access burst.

#### 4.3.4.2.2 CD preamble signature

The CD preamble part of the CPCH access burst carries one of sixteen different orthogonal complex signatures identical to the ones used by the preamble part of the random access burst.

### 4.3.4.3 Channelization codes for the CPCH message part

The signature in the preamble specifies one of the 16 nodes in the code tree that corresponds to channelization codes of length 16. The sub-tree below the specified node is used for spreading of the message part. The control part is always spread with a channelization code of spreading factor 256. The code is chosen from the lowest branch of the sub-tree. The data part may use channelization codes from spreading factor 4 to 64. A UE is allowed to increase its spreading factor during the message transmission by choosing any channelization code from the uppermost branch of the sub-tree code. For channelization codes with spreading factors less that 16, the node is located on the same sub-tree as the channelization code of the access preamble.

#### 4.3.4.4 Scrambling code for the CPCH message part

In addition to spreading, the message part is also subject to scrambling with a 10 ms complex code. The scrambling code is cell-specific and has a one-to-one correspondence to the scrambling code used for the preamble part.

 $-S_{e-msg,n} = C_{seramb,n}$ , for chip indexes 8192...46591 of  $C_{seramb,n}$ .

In the case when the access resources are shared between the RACH and CPCH,

- S<sub>e-msg,n</sub> = C<sub>scramb,n</sub>, for chip indexes 4096...42495 of C<sub>scramb,n</sub>.

The generation of these codes is explained in 4.3.2.2. The mapping of these codes to provide a complex scrambling code is also the same as for the dedicated uplink channels and is described in 4.3.2.1.

NOTE: Use of short scrambling code for CPCH message part is ffs.

# 4.4 Modulation

# 4.4.1 Modulating chip rate

The modulating chip rate is 3.84 Mcps.

## 4.4.2 Modulation

In the uplink, the modulation of both DPCCH and DPDCH is BPSK.